TLV5617A 2.7-V TO 5.5-V LOW-POWER DUAL 10-BIT DIGITAL-TO-ANALOG CONVERTER WITH POWER DOWN SLAS234F – JULY 1999 – REVISED JULY 2002

features

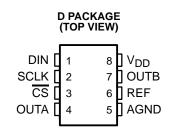
- Dual 10-Bit Voltage Output DAC
- Programmable Settling Time
 3 µs in Fast Mode
 40 µs in Slaw Mode
 - 10 μs in Slow Mode
- Compatible With TMS320 and SPI™ Serial Ports
- Differential Nonlinearity <0.1 LSB Typ
- Monotonic Over Temperature
- Direct Replacement for TLC5617A

description

The TLV5617A is a dual 10-bit voltage output DAC with a flexible 3-wire serial interface. The serial interface is compatible with TMS320, SPI[™], QSPI[™], and Microwire[™] serial ports. It is programmed with a 16-bit serial string containing 4 control bits and 10 data bits.

applications

- Digital Servo Control Loops
- Digital Offset and Gain Adjustment
- Industrial Process Control
- Machine and Motion Control Devices
- Mass Storage Devices



The resistor string output voltage is buffered by a x2 gain rail-to-rail output buffer. The buffer features a Class-AB output stage to improve stability and reduce settling time. The programmable settling time of the DAC allows the designer to optimize speed versus power dissipation.

Implemented with a CMOS process, the device is designed for single supply operation from 2.7 V to 5.5 V. It is available in an 8-pin SOIC package in standard commercial and industrial temperature ranges.

	PACKAGE						
TA	SOIC (D)						
0°C to 70°C	TLV5617ACD						
-40°C to 85°C	TLV5617AID						

AVAILABLE OPTIONS



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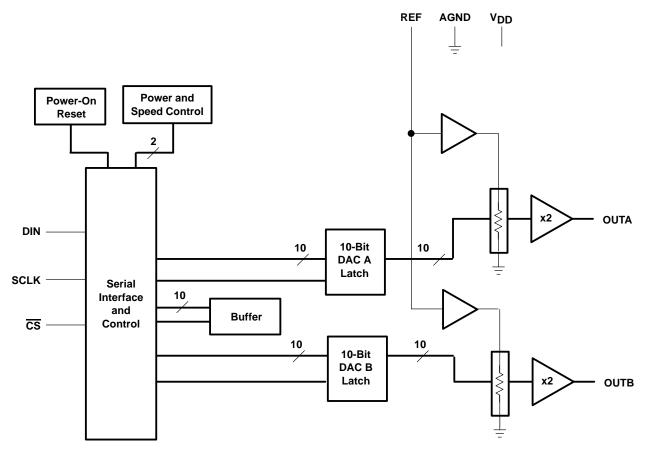
PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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functional block diagram



Terminal Functions

TERM	INAL		DEGODIDEION
NAME	NO.	I/O/P	DESCRIPTION
AGND	5	Р	Ground
CS	3	Ι	Chip select. Digital input active low, used to enable/disable inputs.
DIN	1	Ι	Digital serial data input
OUTA	4	0	DAC A analog voltage output
OUTB	7	0	DAC B analog voltage output
REF	6	Ι	Analog reference voltage input
SCLK	2	I	Digital serial clock input
V _{DD}	8	Р	Positive power supply



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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Supply voltage (V _{DD} to AGND)	
Reference input voltage range	
Digital input voltage range	$\dots \dots \dots \dots \dots - 0.3 \text{ V to V}_{\text{DD}}^{-1} + 0.3 \text{ V}$
Operating free-air temperature range, T _A : TLV5617AC	0°C to 70°C
	–40°C to 85°C
Storage temperature range, T _{stg} Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

		MIN	NOM	MAX	UNIT
	V _{DD} = 5 V	4.5	5	5.5	
Supply voltage, V _{DD}	$V_{DD} = 3 V$	2.7	3	3.3	V
Power on reset, POR		0.55		2	V
High-level digital input voltage, VIH	V _{DD} = 2.7 V	2			
	V _{DD} = 5.5 V	2.4			V
	V _{DD} = 2.7 V			0.6	
Low-level digital input voltage, V _{IL}	V _{DD} = 5.5 V			1	V
	V _{DD} = 5 V (see Note 1)	AGND	2.048	V _{DD} -1.5	
Reference voltage, V _{ref} to REF terminal	V _{DD} = 3 V (see Note 1)	AGND	1.024	V _{DD} -1.5	V
Load resistance, RL		2			kΩ
Load capacitance, CL				100	pF
Clock frequency, fCLK				20	MHz
Or another for a single mean time. T	TLV5617AC	0		70	
Operating free-air temperature, T _A	TLV5617AI	-40		85	°C

NOTE 1: Due to the x2 output buffer, a reference input voltage \geq (V_{DD}-0.4 V)/2 causes clipping of the transfer function.



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electrical characteristics over recommended operating conditions (unless otherwise noted)

power supply

PARAMETER TEST CONDITIONS			MIN	TYP	MAX	UNIT	
	Dower ownels oursent	No load, All inputs = AGND or V_{DD} ,	Fast		1.6	2.5	~^^
IDD Power supply current	DAC latch = 0x800			0.6	1	mA	
	Power down supply current				1		μA
0000	Developments and a string partie	Zero scale, See Note 2			-65		10
PSRR	Power supply rejection ratio	Full scale, See Note 3			-65		dB

NOTES: 2. Power supply rejection ratio at zero scale is measured by varying V_{DD} and is given by: $PSRR = 20 \log \left[(E_{ZS}(V_{DD}max) - E_{ZS}(V_{DD}min)/V_{DD}max) \right]$

3. Power supply rejection ratio at full scale is measured by varying V_{DD} and is given by: $PSRR = 20 \log [(E_G(V_{DD}max) - E_G(V_{DD}min)/V_{DD}max]]$

static DAC specifications

	PARAMETER	TES	T CONDITIONS	MIN	TYP	MAX	UNIT
	Resolution			10			bits
INL	Integral nonlinearity	See Note 4			±0.7	±1	LSB
DNL	Differential nonlinearity	See Note 5			±0.1	±0.5	LSB
EZS	Zero-scale error (offset error at zero scale)	See Note 6				±12	mV
E _{ZS} TC	Zero-scale-error temperature coefficient	See Note 7			3		ppm/°C
-		On a Nata O	V_{DD} = 2.7 V to 3.3 V			±0.6	% full
EG	Gain error	See Note 8	V _{DD} = 4.5 V to 5.5 V			±0.29	scale V
E _G T _C	Gain-error temperature coefficient	See Note 9			1		ppm/°C

NOTES: 4. The relative accuracy of integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale, excluding the effects of zero-code and full-scale errors.

5. The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1-LSB amplitude change of any two adjacent codes.

6. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.

7. Zero-scale-error temperature coefficient is given by: $E_{ZS} TC = [E_{ZS} (T_{max}) - E_{ZS} (T_{min})]/2V_{ref} \times 10^6/(T_{max} - T_{min})$. 8. Gain error is the deviation from the ideal output ($2V_{ref} - 1 LSB$) with an output load of 10 k Ω . 9. Gain temperature coefficient is given by: $E_G T_C = [E_G (T_{max}) - E_g (T_{min})]/2V_{ref} \times 10^6/(T_{max} - T_{min})$.

output specifications

	PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
VO	Output voltage range	$R_L = 10 k\Omega$		V	′DD-0.4	V
	Output load regulation accuracy	V_{O} = 4.096 V, 2.048 V, R_{L} = 2 k Ω to 10 k Ω		±0.1		% FS

reference input

	PARAMETER	TEST CONDITIONS		MIN	TYP MAX	UNIT
VI	Input voltage range				VDD-1.5	V
RI	Input resistance				10	MΩ
Cl	Input capacitance				5	pF
	Defense en innet here desidth		Fast		1.3	MHz
	Reference input bandwidth	$REF = 0.2 V_{pp} + 1.024 V dc$	Slow		525	kHz
	Reference feedthrough	REF = 1 V _{pp} at 1 kHz + 1.024 V dc (see	REF = 1 V _{pp} at 1 kHz + 1.024 V dc (see Note 10)			dB

NOTE 10: Reference feedthrough is measured at the DAC output with an input code = 0x000.



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electrical characteristics over recommended operating conditions (unless otherwise noted) (Continued)

digital inputs

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Ι _Η	High-level digital input current	$V_{I} = V_{DD}$			1	μA
١ _١	Low-level digital input current	$V_{I} = 0 V$	-1			μA
Ci	Input capacitance			8		pF

analog output dynamic performance

	PARAMETER		TEST CONDITIONS			TYP	MAX	UNIT
4		D. 401-0	0. 400 aE Cas Nata 44	Fast		1	3	
^t s(FS)	Output settling time, full scale	$R_{L} = 10 \text{ k}\Omega,$	$C_L = 100 \text{ pF}$, See Note 11	Slow		3	10	μs
		D 4010	Fa			1		_
^t s(CC)	Output settling time, code to code	$R_{L} = 10 \text{ K}\Omega$	$R_L = 10 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Note 12 SI	Slow		2		μs
0.5		D 4010	0 400 x E 0 x Nixty 40	Fast		3		N// -
SR	Slew rate	R _L = 10 kΩ,	$C_L = 100 \text{ pF}$, See Note 13	Slow		0.5		V/µs
	Glitch energy	DIN = 0 to 1,	FCLK = 100 kHz, $\overline{CS} = V_{DD}$			5		nV–s
SNR	Signal-to-noise ratio					68		
SINAD	Signal-to-noise + distortion	f _S = 102 kSPS,	f _{out} = 1 kHz, R _L = 10 kΩ,			65		- ID
THD	Total harmonic distortion	C _L = 100 pF				-62		dB
SFDR	Spurious free dynamic range					64		

NOTES: 11. Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of 0x020 to 0xFDC and 0xFDC to 0x020 respectively. Not tested, assured by design.

12. Settling time is the time for the output signal to remain within ± 0.5 LSB of the final measured value for a digital input code change of one count. Not tested, assured by design.

13. Slew rate determines the time it takes for a change of the DAC output from 10% to 90% of full-scale voltage.



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digital input timing requirements

			MIN	NOM	MAX	UNIT
		V _{DD} = 2.7 V to 3.3 V	10			
^t su(CS–CK)	Setup time, \overline{CS} low before first negative SCLK edge	V _{DD} = 4.5 V to 5.5 V	5			ns
^t su(C16-CS)	Setup time, 16 th negative SCLK edge before CS rising ed	lge	10			ns
t _{wH}	SCLK pulse width high		25			ns
t _{wL}	SCLK pulse width low		25			ns
	Only the state of the second state of the seco	V _{DD} = 2.7 V to 3.3 V	10			
^t su(D)	Setup time, data ready before SCLK falling edge	V _{DD} = 4.5 V to 5.5 V	5			ns
	Held for a data hald with a free OOLK falls a side	V _{DD} = 2.7 V to 3.3 V	10			
^t h(D)	Hold time, data held valid after SCLK falling edge	V _{DD} = 4.5 V to 5.5 V	5			ns

timing requirements

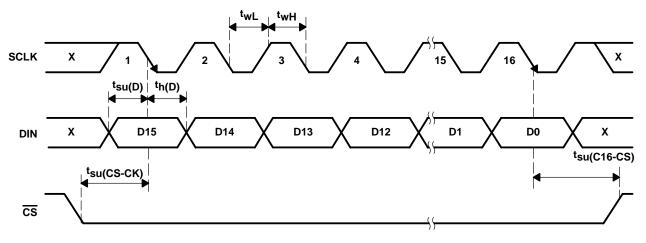
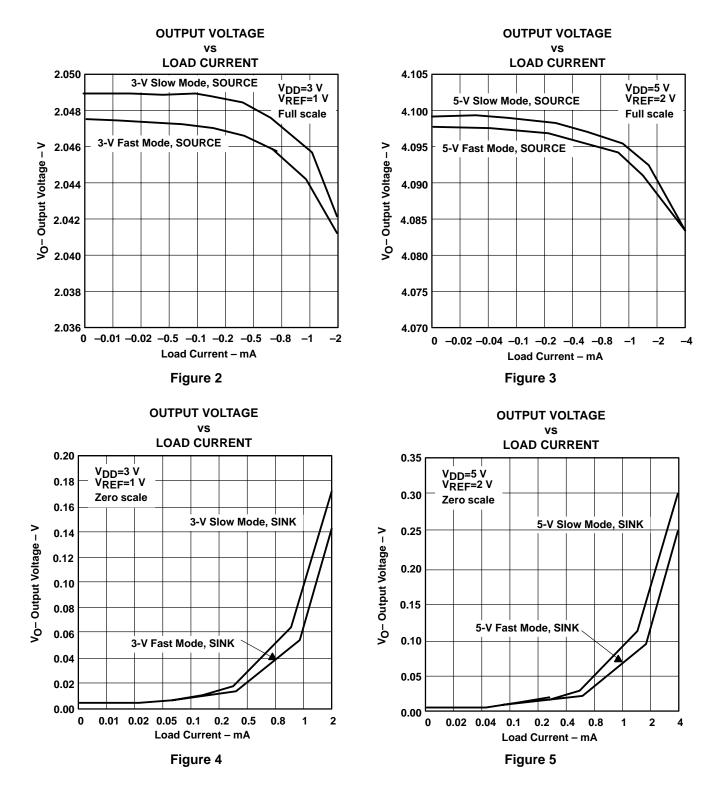


Figure 1. Timing Diagram



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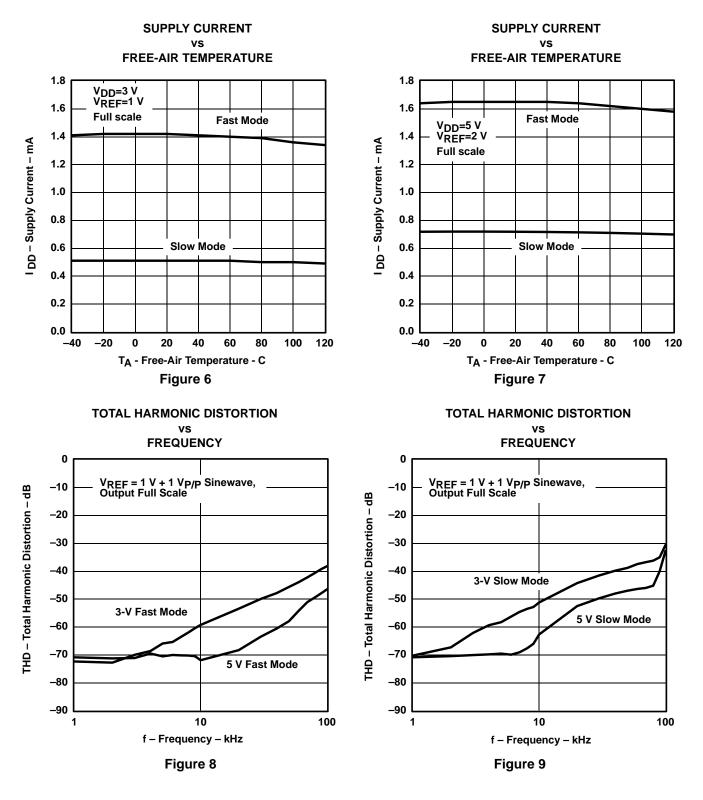
TYPICAL CHARACTERISTICS





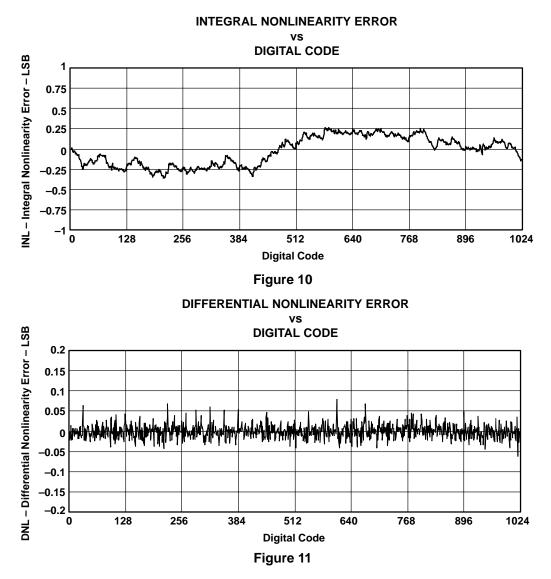
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TYPICAL CHARACTERISTICS





TYPICAL CHARACTERISTICS





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APPLICATION INFORMATION

general function

The TLV5617A is a dual 10-bit, single-supply DAC, based on a resistor-string architecture. It consists of a serial interface, speed and power-down control logic, a resistor string, and a rail-to-rail output buffer.

The output voltage (full scale determined by the reference) is given by:

$$2 \text{ REF } \frac{\text{CODE}}{2^n} [V]$$

Where REF is the reference voltage and CODE is the digital input value within the range of 010 to 2n-1, where n=10 (bits). The 16-bit data word, consisting of control bits and the new DAC value, is illustrated in the data format section. A power-on reset initially resets the internal latches to a defined state (all bits zero).

serial interface

A falling edge of CS starts shifting the data bit-per-bit (starting with the MSB) to the internal register on the falling edges of SCLK. After 16 bits have been transferred or CS rises, the content of the shift register is moved to the target latches (DAC A, DAC B, BUFFER, CONTROL), depending on the control bits within the data word.

Figure 12 shows examples of how to connect the TLV5617A to TMS320, SPI, and Microwire.

TMS320	TLV5617A	SPI	TLV5617A	Microwire	TLV5617A
DSP FSX	<u>CS</u>	1/0	<u>CS</u>	I/O	<u>cs</u>
DX	DIN	MOSI	DIN	SO	DIN
CLKX	SCLK	SCK	SCLK	SK	SCLK

Figure 12. Three-Wire Interface

Notes on SPI and Microwire: Before the controller starts the data transfer, the software has to generate a falling edge on the pin connected to \overline{CS} . If the word width is 8 bits (SPI and Microwire) two write operations must be performed to program the TLV5617A. After the write operation(s), the holding registers or the control register of the DAC update automatically on the rising \overline{CS} edge, ending the write cycle to the DAC. Note: After transfer of the LSB during a data or control write cycle, one additional rising edge on SCLK is required to reset the internal state machine. This edge can occur when \overline{CS} is high or low, but must occur before the next falling \overline{CS} edge that begins the following write cycle. Refer to the timing diagram for more information.

serial clock frequency and update rate

The maximum serial clock frequency is given by:

$$f_{sclkmax} = \frac{1}{t_{whmin} + t_{wlmin}} = 20 \text{ MHz}$$

The maximum update rate is:

$$f_{updatemax} = \frac{1}{16 (t_{whmin} + t_{wlmin})} = 1.25 \text{ MHz}$$

Note that the maximum update rate is just a theoretical value for the serial interface, as the settling time of the TLV5617A should also be considered.



2.7-V TO 5.5-V LOW-POWER DUAL 10-BIT DIGITAL-TO-ANALOG CONVERTER WITH POWER DOWN SLAS234F – JULY 1999 – REVISED JULY 2002

APPLICATION INFORMATION

data format

The 16-bit data word for the TLV5617A consists of two parts:

- Program bits (D15..D12)
- New data (D11..D0)

	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Γ	R1	SPD	PWR	R0	MSB				10 Da	ta bits				LSB	0	0

The following table lists all possible combinations of register-select bits:

register-select bits

R1	R0	REGISTER
0	0	Write data to DAC B and BUFFER
0	1	Write data to BUFFER
1	0	Write data to DAC A and update DAC B with BUFFER content
1	1	Reserved

The meaning of the 12 data bits depends on the register. If one of the DAC registers or the BUFFER is selected, then the 12 data bits determine the new DAC value:

examples of operation

• Set DAC A output, select fast mode:

Write new DAC A value and update DAC A output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	0				Nev	v DAC A	output va	alue				0	0

The DAC A output is updated on the rising clock edge after D0 is sampled.

• Set DAC B output, select fast mode:

Write new DAC B value to BUFFER and update DAC B output:

D1	5	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0		1	0	0				BUFFER	content a	nd DAC	B output	value			0	0

The DAC A output is updated on the rising clock edge after D0 is sampled.

- Set DAC A value, set DAC B value, update both simultaneously, select slow mode:
 - 1. Write data for DAC B to BUFFER:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	1					New DAC	B value					0	0

2. Write new DAC A value and update DAC A and B simultaneously:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	0					New DAC	CA value					0	0



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APPLICATION INFORMATION

examples of operation (continued)

Both outputs are updated on the rising clock edge after D0 from the DAC A data word is sampled.

Set powerdown mode:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Х	Х	1	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х

X = Don't care

linearity, offset, and gain error using single ended supplies

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset, the output voltage may not change with the first code, depending on the magnitude of the offset voltage.

The output amplifier attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground and clamps the output at 0 V.

The output voltage then remains at zero until the input code value produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in Figure 13.

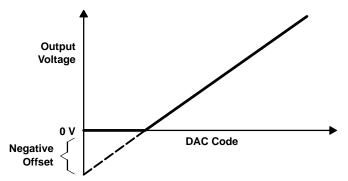


Figure 13. Effect of Negative Offset (Single Supply)

This offset error, not the linearity error, produces this breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below the ground rail.

For a DAC, linearity is measured between zero-input code (all inputs 0) and full-scale code (all inputs 1) after offset and full scale are adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity is measured between full-scale code and the lowest code that produces a positive output voltage.

definitions of specifications and terminology

integral nonlinearity (INL)

The relative accuracy or integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

differential nonlinearity (DNL)

The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.



definitions of specifications and terminology (continued)

zero-scale error (E_{ZS})

Zero-scale error is defined as the deviation of the output from 0 V at a digital input value of 0.

gain error (E_G)

Gain error is the error in slope of the DAC transfer function.

total harmonic distortion (THD)

THD is the ratio of the rms value of the first six harmonic components to the value of the fundamental signal. The value for THD is expressed in decibels.

signal-to-noise ratio + distortion (S/N+D)

S/N+D is the ratio of the rms value of the output signal to the rms sum of all other spectral components below the Nyquist frequency, including harmonics but excluding dc. The value for S/N+D is expressed in decibels.

spurious free dynamic range (SFDR)

Spurious free dynamic range is the difference between the rms value of the output signal and the rms value of the largest spurious signal within a specified bandwidth. The value for SFDR is expressed in decibels.





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PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TLV5617ACD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV5617ACDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV5617ACDR	ACTIVE	SOIC	D	8	2500	TBD	Call TI	Call TI	
TLV5617ACDRG4	ACTIVE	SOIC	D	8	2500	TBD	Call TI	Call TI	
TLV5617AID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV5617AIDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV5617AIDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV5617AIDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and pa

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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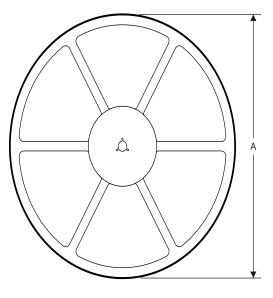
PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION

REEL DIMENSIONS

TEXAS INSTRUMENTS





TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV5617AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TEXAS INSTRUMENTS

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PACKAGE MATERIALS INFORMATION

14-Jul-2012



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV5617AIDR	SOIC	D	8	2500	367.0	367.0	35.0

D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.





NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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