

TMUX611x $\pm 17V$ 低电容、低泄漏电流、 精密四通道 SPST 开关

1 特性

- 宽电源电压范围： $\pm 5V$ 至 $\pm 17V$ （双电源）， $10V$ 至 $17V$ （单电源）
- 所有引脚的闩锁性都能达到 $100mA$ ，符合 JESD78 II 类 A 级要求
- 低导通电容： $4.2pF$
- 低输入泄漏： $0.5pA$
- 低电荷注入： $0.6pC$
- 轨至轨运行
- 低导通电阻： 120Ω
- 快速开关开启时间： $66ns$
- 先断后合开关 (TMUX6113)
- EN 引脚可连接至 V_{DD}
- 低电源电流： $17\mu A$
- 人体放电模型 (HBM) ESD 保护：针对所有引脚提供 $\pm 2kV$ 保护
- 行业标准 TSSOP 封装和较小的 WQFN 封装

2 应用

- 工厂自动化和工业过程控制
- 可编程逻辑控制器 (PLC)
- 模拟输入模块
- 半导体测试设备
- 电池测试设备

3 说明

TMUX6111、TMUX6112 和 TMUX6113 器件是现代化的互补金属氧化物半导体 (CMOS) 器件，具有四个独立的可选单刀单掷 (SPST) 开关。这些器件在双电源 ($\pm 5V$ 至 $\pm 17V$)、单电源 ($10V$ 至 $17V$) 或非对称电源供电时均能正常运行。所有数字输入均具有兼容晶体管-晶体管逻辑 (TTL) 的阈值，从而确保 TTL/CMOS 逻辑兼容性。

逻辑 0 会打开 TMUX6111 中数字控制输入上的开关。要打开 TMUX6112 中的开关，则需要逻辑 1。

TMUX6113 有两个开关的数字控制逻辑与 TMUX6111 类似，而另外两个开关上的逻辑则与之相反。

TMUX6113 具有先断后合开关，因此可用于交叉点开关应用。

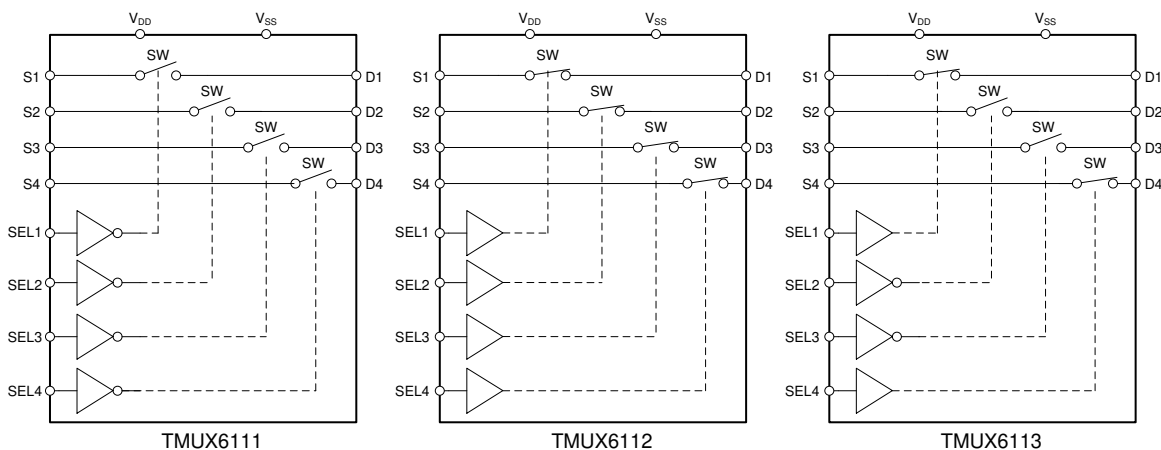
TMUX611x 器件是德州仪器 (TI) 精密开关和多路复用器系列的一部分。这些器件具有非常低的泄漏电流和电荷注入，因此可用于高精度测量应用中的数字输入 D 类音频放大器。这些器件的电源电流低至 $17\mu A$ ，因此可用于便携式应用供电的出色器件。

器件信息⁽¹⁾

器件型号	封装	封装尺寸 (标称值)
TMUX6111	TSSOP (16)	5.00mm x 4.40mm
TMUX6112	WQFN (16)	3.00mm x 3.00mm
TMUX6113		

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。

简化原理图



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4 修订历史记录

注：之前版本的页码可能与当前版本有所不同。

Changes from Revision D (January 2019) to Revision E	Page
• 将标题中的“TMUX611x ± 16.5 V”更改成了“TMUX611x ± 17 V”.....	1
• 将特性中的“宽电源电压范围： ± 5 V 至 ± 16.5 V（双电源），10V 至 16.5V（单电源）”更改成了“宽电源电压范围： ± 5 V 至 ± 17 V（双电源），10V 至 17V（单电源）”.....	1
• 将说明中的“双电源（ ± 5 V 至 ± 16.5 V）、单电源（10V 至 16.5V）”更改成了“双电源（ ± 5 V 至 ± 17 V）、单电源（10V 至 17V）”.....	1
• Changed ± 16.5 -V to ± 17.5 -V in the Description of the <i>Device Comparison Table</i>	4
• Changed recommended power supply voltage differential from 33 V to 34 V.....	5
• Changed recommended single supply voltage from 16.5 V to 17 V.....	5
• Changed positive and negative power supply voltage to +17 V and -17V.....	5
• The <i>Overview</i> From: dual supplies (± 5 V to ± 16.5 V) or single supply (10 V to 16.5 V) To: dual supplies (± 5 V to ± 17 V) or single supply (10 V to 17 V).....	13
• Changed the <i>Application Information</i> From: 16.5 V (single supply) To: 17 V (single supply).....	20
• The <i>Power Supply Recommendations</i> From: wide supply range of of ± 5 V to ± 16.5 V (10 V to 16.5 V in single-supply mode) To: wide supply range of of ± 5 V to ± 17 V (10 V to 17 V in single-supply mode).....	22

Changes from Revision C (December 2018) to Revision D	Page
• Changed descriptions in the <i>Device Comparison Table</i> to match the data sheet title.....	4
• 已更改 图 30 to correct Op-Amp terminal polarities.	20

Changes from Revision B (November 2018) to Revision C	Page
• Changed units for channel current and ambient temperature.....	6

Changes from Revision A (November 2018) to Revision B **Page**

- 针对 TMUX6111 和 TMUX6113 将文档状态从产品预览 更改成了生产数据..... 1
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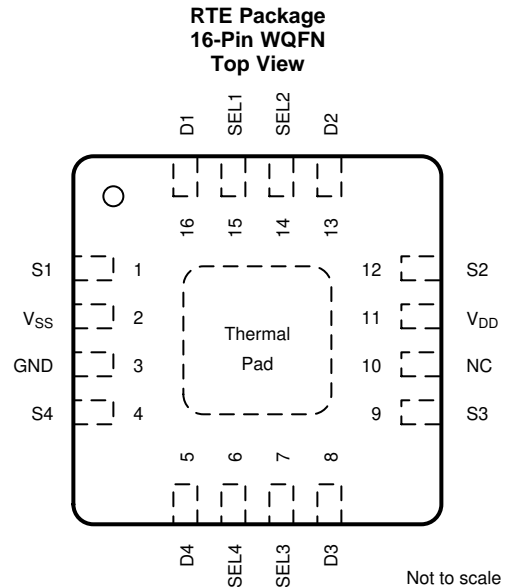
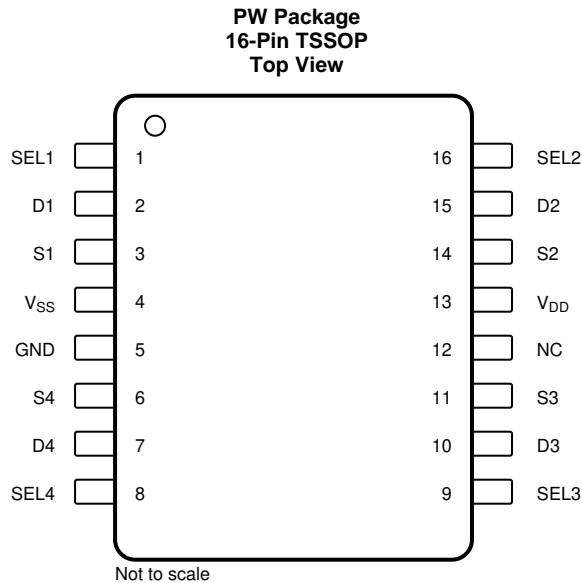
Changes from Original (August 2018) to Revision A **Page**

- 针对 TMUX6112 将文档状态从预告信息 更改成了生产数据 1
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5 Device Comparison Table

PRODUCT	DESCRIPTION
TMUX6111	±17-V, Low-Capacitance, Low-Leakage-Current, Precision, Quad SPST Switches (Normally Closed)
TMUX6112	±17-V, Low-Capacitance, Low-Leakage-Current, Precision, Quad SPST Switches (Normally Open)
TMUX6113	±17-V, Low-Capacitance, Low-Leakage-Current, Precision, Quad SPST Switches (Dual Open + Dual Closed)

6 Pin Configuration and Functions



Pin Functions

PIN			TYPE ⁽¹⁾	DESCRIPTION
NAME	TSSOP	WQFN		
SEL1	1	15	I	Logic control input 1.
D1	2	16	I/O	Drain pin 1. Can be an input or output.
S1	3	1	I/O	Source pin 1. Can be an input or output.
V _{SS}	4	2	P	Negative power supply. This pin is the most negative power-supply potential. In single-supply applications, this pin can be connected to ground. For reliable operation, connect a decoupling capacitor ranging from 0.1 μF to 10 μF between V _{SS} and GND.
GND	5	3	P	Ground (0 V) reference
S4	6	4	I/O	Source pin 4. Can be an input or output.
D4	7	5	I/O	Drain pin 4. Can be an input or output.
SEL4	8	6	I	Logic control input 4.
SEL3	9	7	I	Logic control input 3.
D3	10	8	I/O	Drain pin 3. Can be an input or output.
S3	11	9	I/O	Source pin 3. Can be an input or output.
NC	12	10	–	No internal connection.
V _{DD}	13	11	P	Positive power supply. This pin is the most positive power-supply potential. For reliable operation, connect a decoupling capacitor ranging from 0.1 μF to 10 μF between V _{DD} and GND.
S2	14	12	I/O	Source pin 2. Can be an input or output.
D2	15	13	I/O	Drain pin 2. Can be an input or output.
SEL2	16	14	I	Logic control input 2.
–	–	EP	–	Exposed Pad. The exposed pad is electrically connected to V _{SS} internally. Connect EP to V _{SS} to achieve rated thermal and ESD performance.

(1) I = input, O = output, I/O = input and output, P = power

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
V _{DD} to V _{SS}	Supply voltage		36	V
V _{DD} to GND		-0.3	18	V
V _{SS} to GND		-18	0.3	V
V _{DIG}	Digital input pin (SEL1, SEL2, SEL3, SEL4) voltage	GND -0.3	V _{DD} +0.3	V
I _{DIG}	Digital input pin (SEL1, SEL2, SEL3, SEL4) current	-30	30	mA
V _{ANA_IN}	Analog input pin (Sx) voltage	V _{SS} -0.3	V _{DD} +0.3	V
I _{ANA_IN}	Analog input pin (Sx) current	-30	30	mA
V _{ANA_OUT}	Analog output pin (D) voltage	V _{SS} -0.3	V _{DD} +0.3	V
I _{ANA_OUT}	Analog output pin (D) current	-30	30	mA
T _A	Ambient temperature	-55	140	°C
T _J	Junction temperature		150	°C
T _{stg}	Storage temperature	-65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	±2000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±500	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Thermal Information

THERMAL METRIC		TMUX6111/ TMUX6112/ TMUX6113		UNIT
		PW (TSSOP)	RTE (QFN)	
		16 PINS	16 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	111.0	51.9	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	41.7	53.3	°C/W
R _{θJB}	Junction-to-board thermal resistance	57.2	26.6	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	4.1	1.7	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	56.6	26.6	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	N/A	11.6	°C/W

7.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V _{DD} to V _{SS} ⁽¹⁾	Power supply voltage differential	10		34	V
V _{DD} to GND	Positive power supply voltage (single supply, V _{SS} = 0 V)	10		17	V
V _{DD} to GND	Positive power supply voltage (dual supply)	5		17	V
V _{SS} to GND	Negative power supply voltage (dual supply)	-5		-17	V
V _S ⁽²⁾	Source pins voltage	V _{SS}		V _{DD}	V

(1) V_{DD} and V_{SS} can be any value as long as 10 V ≤ (V_{DD} - V_{SS}) ≤ 34 V.

(2) V_S is the voltage on all the S pins.

Recommended Operating Conditions (continued)

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_D	Drain pin voltage	V_{SS}		V_{DD}	V
V_{DIG}	Digital input pin (SEL1, SEL2, SEL3, SEL4) voltage	0		V_{DD}	V
I_{CH}	Channel current ($T_A = 25^\circ\text{C}$)	-25		25	mA
T_A	Ambient temperature	-40		125	$^\circ\text{C}$

7.5 Electrical Characteristics (Dual Supplies: $\pm 15\text{ V}$)

 at $T_A = 25^\circ\text{C}$, $V_{DD} = 15\text{ V}$, and $V_{SS} = -15\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
ANALOG SWITCH							
V_A	Analog signal range		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	V_{SS}		V_{DD}	V
R_{ON}	On-resistance	$V_S = 0\text{ V}$, $I_S = 1\text{ mA}$			120	135	Ω
					140	160	Ω
		$V_S = \pm 10\text{ V}$, $I_S = 1\text{ mA}$	$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			210	Ω
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			245	Ω
ΔR_{ON}	On-resistance mismatch between channels	$V_S = \pm 10\text{ V}$, $I_S = 1\text{ mA}$			2.5	6	Ω
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			9	Ω
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			11	Ω
R_{ON_FLAT}	On-resistance flatness	$V_S = -10\text{ V}$, 0 V , $+10\text{ V}$, $I_S = 1\text{ mA}$	$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$		23	33	Ω
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			37	Ω
					38	Ω	
R_{ON_DRIFT}	On-resistance drift	$V_S = 0\text{ V}$			0.52		$\%/^\circ\text{C}$
$I_{S(OFF)}$	Source off leakage current ⁽¹⁾	Switch state is off, $V_S = +10\text{ V}/-10\text{ V}$, $V_D = -10\text{ V}/+10\text{ V}$		-0.02	0.005	0.02	nA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$	-0.14		0.05	nA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	-1.3		0.25	nA
$I_{D(OFF)}$	Drain off leakage current ⁽¹⁾	Switch state is off, $V_S = +10\text{ V}/-10\text{ V}$, $V_D = -10\text{ V}/+10\text{ V}$		-0.02	0.005	0.02	nA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$	-0.14		0.05	nA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	-1.3		0.25	nA
$I_{D(ON)}$	Drain on leakage current	Switch state is on, $V_S = +10\text{ V}/-10\text{ V}$, $V_D = -10\text{ V}/+10\text{ V}$		-0.04	0.01	0.04	nA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$	-0.25		0.1	nA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	-1.8		0.5	nA
DIGITAL INPUT (SELx pins)							
V_{IH}	Logic voltage high			2			V
V_{IL}	Logic voltage low					0.8	V
$R_{PD(IN)}$	Pull-down resistance on SELx pins				6		M Ω
POWER SUPPLY							
I_{DD}	V_{DD} supply current	$V_A = 0\text{ V}$ or 3.3 V , $V_S = 0\text{ V}$			17	21	μA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			22	μA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			23	μA
I_{SS}	V_{SS} supply current	$V_A = 0\text{ V}$ or 3.3 V , $V_S = 0\text{ V}$			8	10	μA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			11	μA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			12	μA

 (1) When V_S is positive, V_D is negative, and vice versa.

7.6 Switching Characteristics (Dual Supplies: ±15 V)

 at $T_A = 25^\circ\text{C}$, $V_{DD} = 15\text{ V}$, and $V_{SS} = -15\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{ON}	Enable turn-on time	$V_S = \pm 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$		66	78	ns
		$V_S = \pm 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			107	ns
		$V_S = \pm 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			117	ns
t_{OFF}	Enable turn-off time	$V_S = \pm 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$		56	68	ns
		$V_S = \pm 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			77	ns
		$V_S = \pm 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			81	ns
t_{BBM}	Break-before-make time delay (TMUX6113 Only)	$V_S = 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	8	40		ns
Q_J	Charge injection	$V_S = 0\text{ V}$, $R_S = 0\ \Omega$, $C_L = 1\text{ nF}$		0.6		pC
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$		-85		dB
X_{TALK}	Channel-to-channel crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$, adjacent channel		-100		dB
		$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$, non-adjacent channel		-115		dB
I_L	Insertion loss	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$		-7.0		dB
ACPSRR	AC Power Supply Rejection Ratio	$R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $V_{PP} = 0.62\text{ V}$ on V_{DD} , $f = 1\text{ MHz}$		-59		dB
		$R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $V_{PP} = 0.62\text{ V}$ on V_{SS} , $f = 1\text{ MHz}$		-59		dB
BW	-3dB Bandwidth	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$		800		MHz
THD	Total harmonic distortion + noise	$R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $f = 20\text{ Hz}$ to 20 kHz		0.08		%
C_{IN}	Digital input capacitance	$V_{IN} = 0\text{ V}$ or V_{DD}		1.5		pF
$C_{S(OFF)}$	Source off-capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$ (PW package)		1.9	3.0	pF
		$V_S = 0\text{ V}$, $f = 1\text{ MHz}$ (RTE package)		2.5	3.6	pF
$C_{D(OFF)}$	Drain off-capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$		2.4	3.1	pF
$C_{S(ON)}$, $C_{D(ON)}$	Source and drain on-capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$		4.2	6.0	pF

7.7 Electrical Characteristics (Single Supply: 12 V)

 at $T_A = 25^\circ\text{C}$, $V_{DD} = 12\text{ V}$, and $V_{SS} = 0\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT		
ANALOG SWITCH								
V_A	Analog signal range	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	V_{SS}		V_{DD}	V		
R_{ON}	On-resistance	$V_S = 10\text{ V}$, $I_S = 1\text{ mA}$		230	265	Ω		
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			355	Ω	
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			405	Ω	
ΔR_{ON}	On-resistance mismatch between channels	$V_S = 10\text{ V}$, $I_S = 1\text{ mA}$		5	12	Ω		
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			19	Ω	
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			23	Ω	
R_{ON_DRIFT}	On-resistance drift	$V_S = 0\text{ V}$		0.5		%/ $^\circ\text{C}$		
$I_{S(OFF)}$	Source off leakage current ⁽¹⁾	Switch state is off, $V_S = 10\text{ V}/1\text{ V}$, $V_D = 1\text{ V}/10\text{ V}$		-0.02	0.005	0.02	nA	
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$		-0.1		0.04	nA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		-1		0.2	nA

 (1) When V_S is positive, V_D is negative, and vice versa.

Electrical Characteristics (Single Supply: 12 V) (continued)

 at $T_A = 25^\circ\text{C}$, $V_{DD} = 12\text{ V}$, and $V_{SS} = 0\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$I_{D(OFF)}$	Drain off leakage current ⁽¹⁾	Switch state is off, $V_S = 10\text{ V} / 1\text{ V}$, $V_D = 1\text{ V} / 10\text{ V}$		-0.02	0.005	0.02	nA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$	-0.1		0.04	nA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	-1		0.2	nA
$I_{D(ON)}$	Drain on leakage current	Switch state is on, $V_S =$ floating, $V_D = 1\text{ V} / 10\text{ V}$		-0.04	0.01	0.04	nA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$	-0.16		0.08	nA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	-1.4		0.4	nA
DIGITAL INPUT (SELx pins)							
V_{IH}	Logic voltage high			2			V
V_{IL}	Logic voltage low					0.8	V
$R_{PD(EN)}$	Pull-down resistance on SELx pins				6		M Ω
POWER SUPPLY							
I_{DD}	V_{DD} supply current	$V_A = 0\text{ V}$ or 3.3 V , $V_S = 0\text{ V}$			13	16	μA
			$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			17	μA
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			18	μA

7.8 Switching Characteristics (Single Supply: 12 V)

 at $T_A = 25^\circ\text{C}$, $V_{DD} = 12\text{ V}$, and $V_{SS} = 0\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t_{ON}	Enable turn-on time	$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$			72	84	ns
		$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$				117	ns
		$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$				128	ns
t_{OFF}	Enable turn-off time	$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$			57	66	ns
		$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$				78	ns
		$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$				84	ns
t_{BBM}	Break-before-make time delay (TMUX6113 only)	$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$, $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		17	47		ns
Q_J	Charge injection	$V_S = 0\text{ V}$ to 12 V , $R_S = 0\ \Omega$, $C_L = 1\text{ nF}$			0.6		pC
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$			-86		dB
X_{TALK}	Channel-to-channel crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$, adjacent channel			-98		dB
		$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$, non-adjacent channel			-117		dB
I_L	Insertion loss	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$			-14		dB
ACPSRR	AC Power Supply Rejection Ratio	$R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $V_{PP} = 0.62\text{ V}$, $f = 1\text{ MHz}$			-59		dB
BW	-3dB Bandwidth	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$			750		MHz
C_{IN}	Digital input capacitance	$V_{IN} = 0\text{ V}$ or V_{DD}			1.6		pF
$C_{S(OFF)}$	Source off-capacitance	$V_S = 6\text{ V}$, $f = 1\text{ MHz}$ (PW package)			2.2	3.1	pF
		$V_S = 6\text{ V}$, $f = 1\text{ MHz}$ (RTE package)			2.9	4.0	pF
$C_{D(OFF)}$	Drain off-capacitance	$V_S = 6\text{ V}$, $f = 1\text{ MHz}$			2.8	3.5	pF
$C_{S(ON)}$, $C_{D(ON)}$	Source and drain on-capacitance	$V_S = 6\text{ V}$, $f = 1\text{ MHz}$			4.6	6.3	pF

7.9 Typical Characteristics

at $T_A = 25^\circ\text{C}$, $V_{DD} = 15\text{ V}$, and $V_{SS} = -15\text{ V}$ (unless otherwise noted)

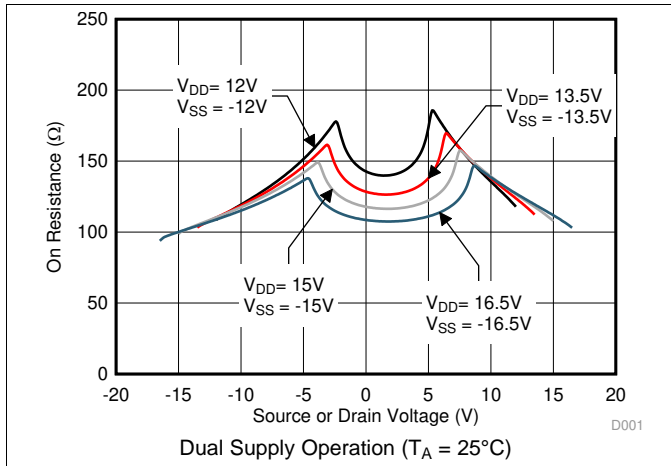


图 1. On-Resistance vs Source or Drain Voltage

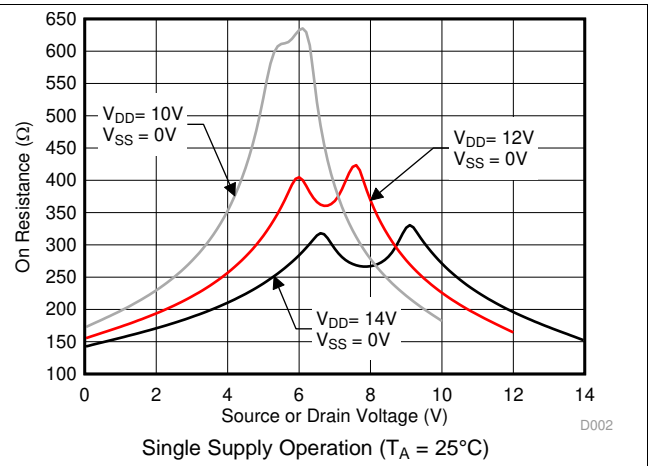


图 2. On-Resistance vs Source or Drain Voltage

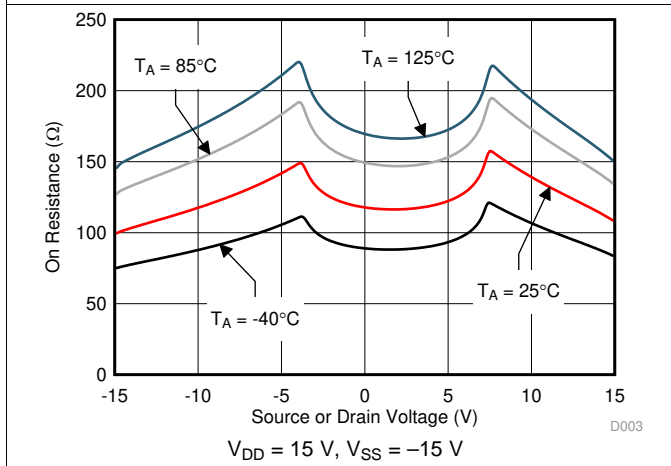


图 3. On-Resistance vs Source or Drain Voltage

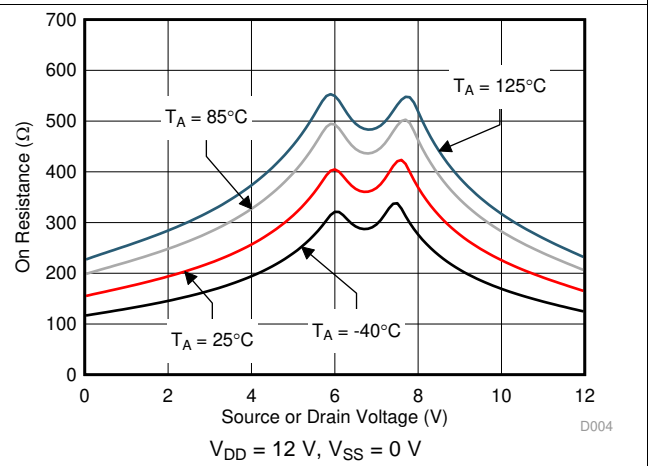


图 4. On-Resistance vs Source or Drain Voltage

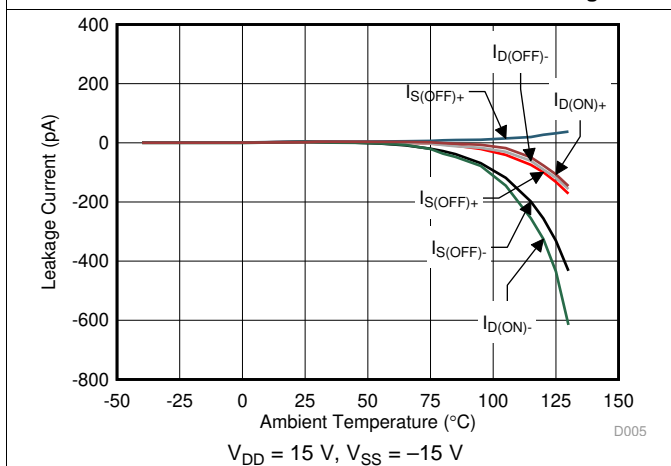


图 5. Leakage Current vs Temperature

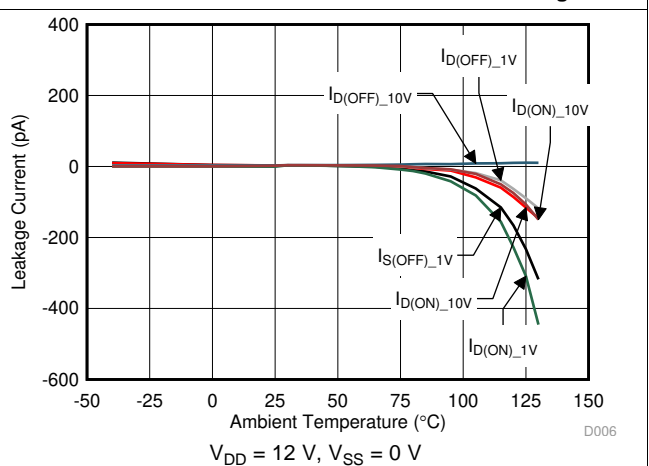
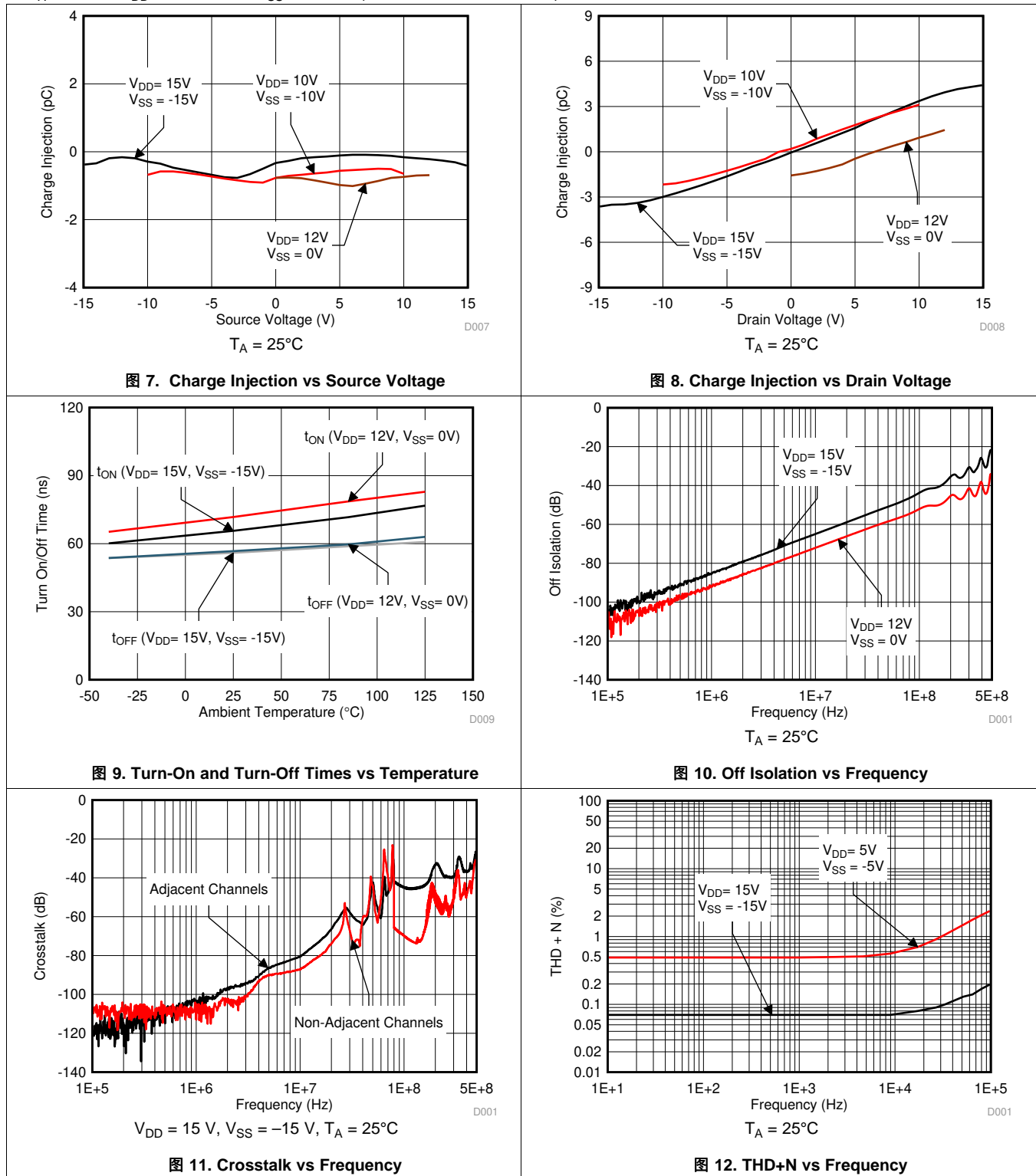


图 6. Leakage Current vs Temperature

Typical Characteristics (接下页)

at $T_A = 25^\circ\text{C}$, $V_{DD} = 15\text{ V}$, and $V_{SS} = -15\text{ V}$ (unless otherwise noted)



Typical Characteristics (接下页)

at $T_A = 25^\circ\text{C}$, $V_{DD} = 15\text{ V}$, and $V_{SS} = -15\text{ V}$ (unless otherwise noted)

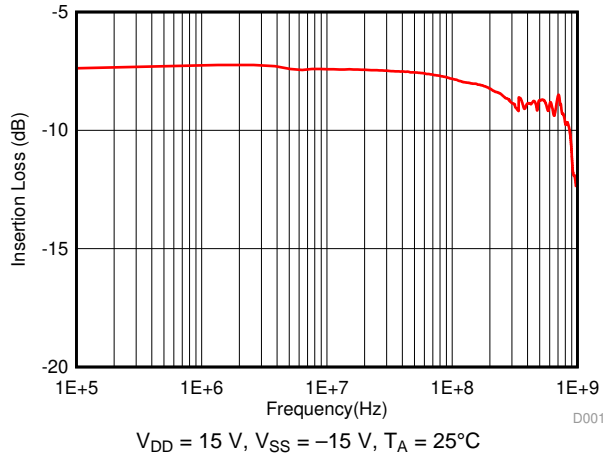


图 13. On Response vs Frequency

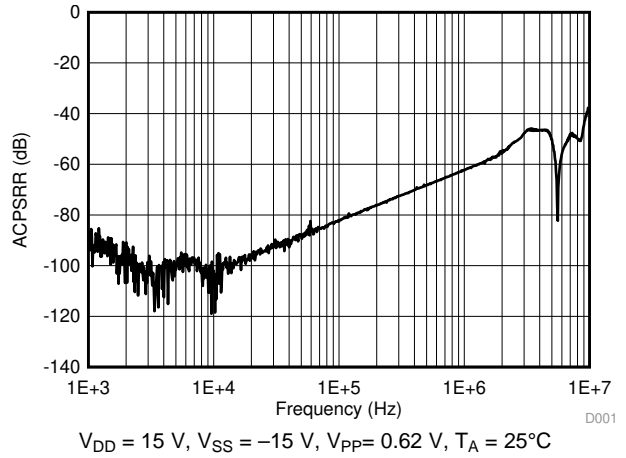


图 14. ACPSRR vs Frequency

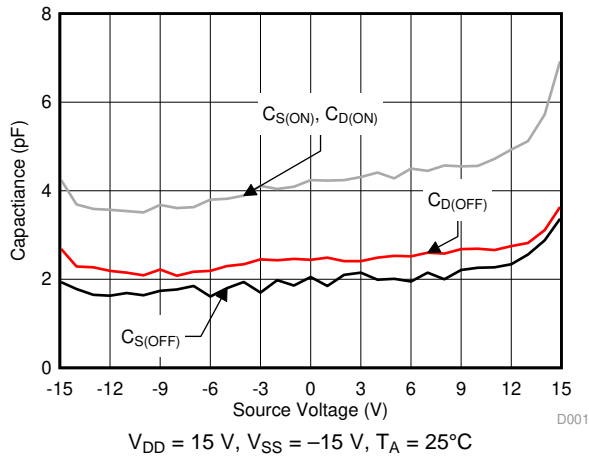


图 15. Capacitance vs Source Voltage

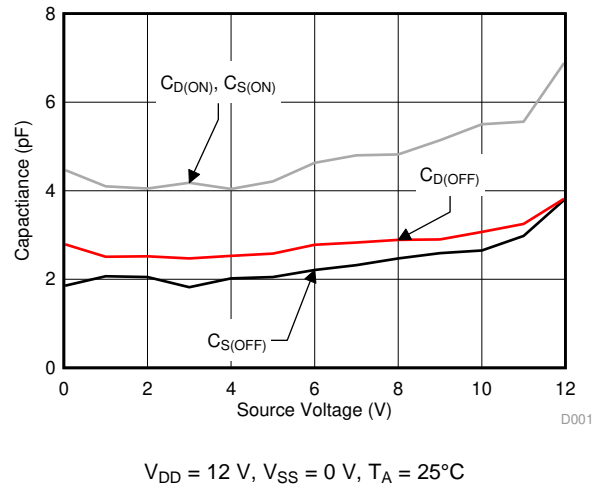


图 16. Capacitance vs Source Voltage

8 Parameter Measurement Information

8.1 Truth Tables

表 1, 表 2, 表 3 and show the truth tables for the TMUX6111, TMUX6112, and TMUX6113, respectively.

表 1. TMUX6111 Truth Table

SELx	STATE
0	All Switch ON
1	All Switch OFF

表 2. TMUX6112 Truth Table

SELx	STATE
0	All Switch OFF
1	All Switch ON

表 3. TUMUX6113 Truth Table

SELx	STATE
0	Switch 1, 4 OFF Switch 2, 3 ON
1	Switch 1, 4 ON Switch 2, 3 OFF

9 Detailed Description

9.1 Overview

The TMUX6111, TMUX6112, and TMUX6113 are 4-channel single-pole/ single-throw (SPDT) switches that supports dual supplies (± 5 V to ± 17 V) or single supply (10 V to 17 V) operation. Each channel of the switch is turned on or turned off based on the state of its corresponding SELx pin. The [Functional Block Diagram](#) section provides a top-level block diagram of the switches.

9.1.1 On-Resistance

The on-resistance of the TMUX6111, TMUX6112, and TMUX6113 is the ohmic resistance across the source (Sx) and drain (Dx) pins of the device. The on-resistance varies with input voltage and supply voltage. The symbol R_{ON} is used to denote on-resistance. The measurement setup used to measure R_{ON} is shown in [图 17](#). Voltage (V) and current (I_{CH}) are measured using this setup, and R_{ON} is computed as shown in [公式 1](#):

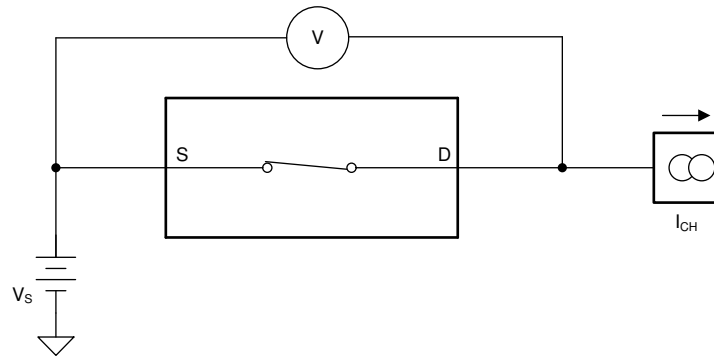


图 17. On-Resistance Measurement Setup

$$R_{ON} = V / I_{CH} \quad (1)$$

9.1.2 Off-Leakage Current

There are two types of leakage currents associated with a switch during the off state:

1. Source off-leakage current
2. Drain off-leakage current

Source leakage current is defined as the leakage current flowing into or out of the source pin when the switch is off. This current is denoted by the symbol $I_{S(OFF)}$.

Drain leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is off. This current is denoted by the symbol $I_{D(OFF)}$.

The setup used to measure both off-leakage currents is shown in [图 18](#)

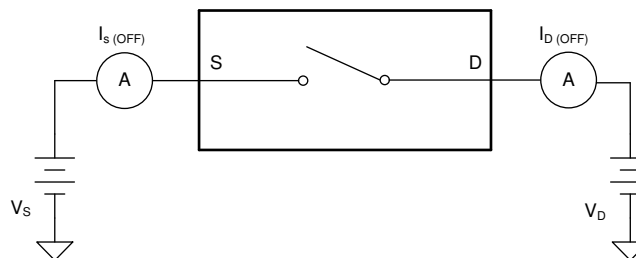


图 18. Off-Leakage Measurement Setup

Overview (接下页)

9.1.3 On-Leakage Current

On-leakage current is defined as the leakage current that flows into or out of the drain pin when the switch is in the on state. The source pin is left floating during the measurement. 图 19 shows the circuit used for measuring the on-leakage current, denoted by $I_{D(ON)}$.

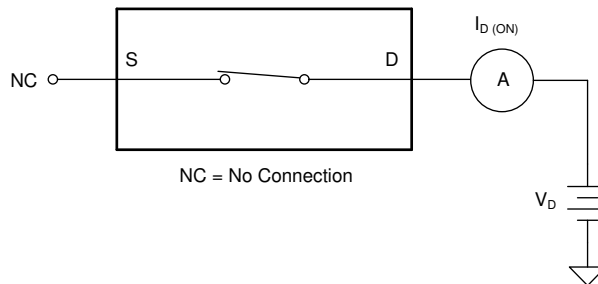


图 19. On-Leakage Measurement Setup

9.1.4 Break-Before-Make Delay

The break-before-make delay is a safety feature of the TMUX6113 switch. The TMUX6113's ON switches first break the connection before the OFF switches make connection. The time delay between the *break* and the *make* is known as break-before-make delay. 图 20 shows the setup used to measure break-before-make delay, denoted by the symbol t_{BBM} .

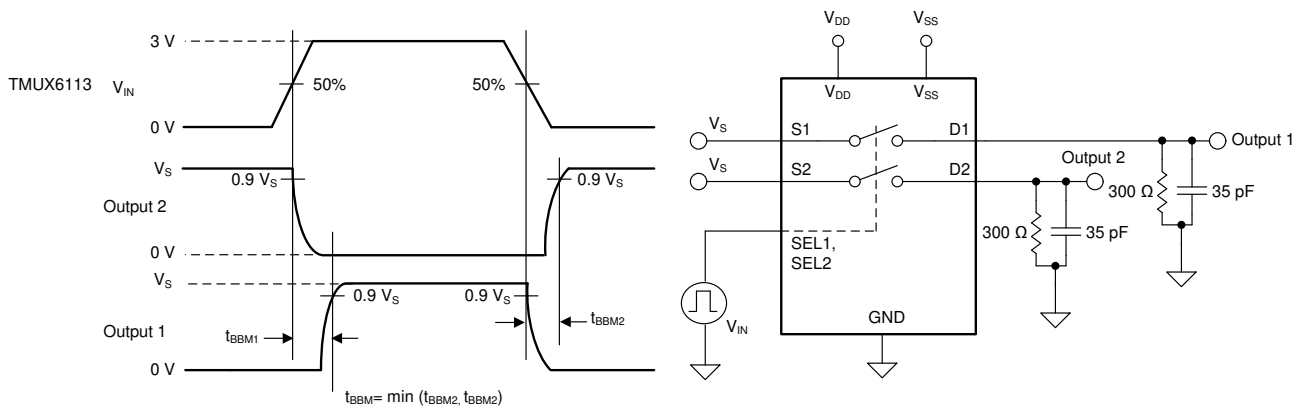


图 20. Break-Before-Make Delay Measurement Setup

9.1.5 Turn-On and Turn-Off Time

Turn-on time is defined as the time taken by the output of the TMUX6111, TMUX6112, and TMUX6113 to rise to a 90% final value after the SELx signal has risen (for NC switches) or fallen (for NO switches) to a 50% final value. 图 21 shows the setup used to measure turn-on time. Turn-on time is denoted by the symbol t_{ON} .

Turn off time is defined as the time taken by the output of the TMUX6111, TMUX6112, and TMUX6113 to fall to a 10% initial value after the SELx signal has fallen (for NC switches) or risen (for NO switches) to a 50% initial value. 图 21 shows the setup used to measure turn-off time. Turn-off time is denoted by the symbol t_{OFF} .

Overview (接下页)

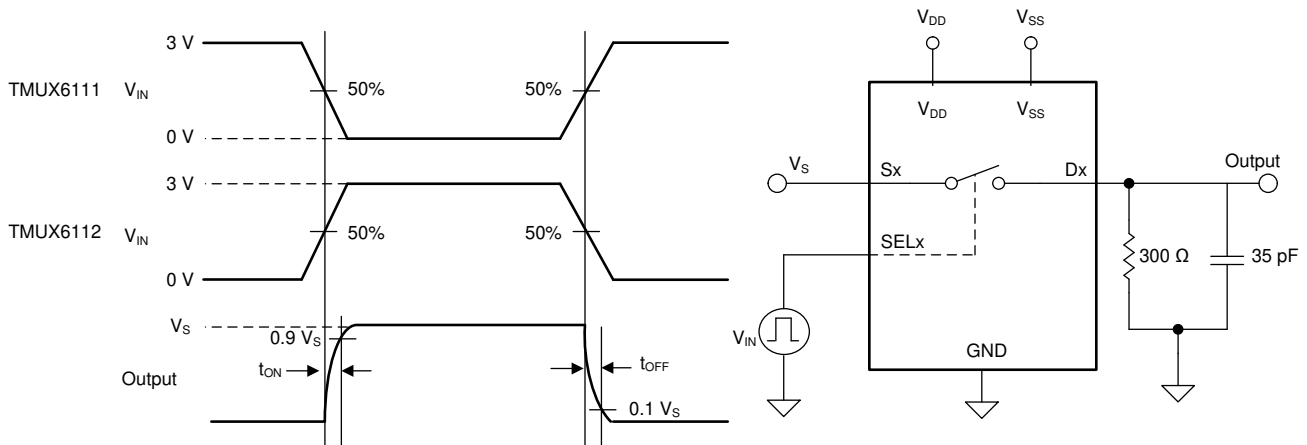


图 21. Turn-On and Turn-Off Time Measurement Setup

9.1.6 Charge Injection

The TMUX6111, TMUX6112, and TMUX6113 have a simple transmission-gate topology. Any mismatch in capacitance between the NMOS and PMOS transistors results in a charge injected into the drain or source during the falling or rising edge of the gate signal. The amount of charge injected into the source or drain of the device is known as charge injection, and is denoted by the symbol Q_{INJ} . 图 22 shows the setup used to measure charge injection.

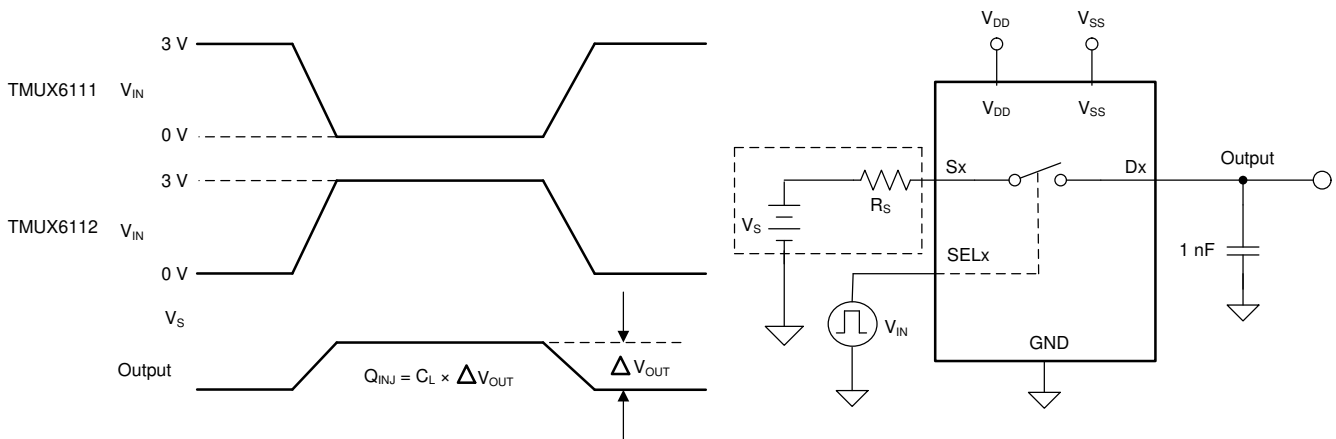


图 22. Charge-Injection Measurement Setup

9.1.7 Off Isolation

Off isolation is defined as the voltage at the drain pin (Dx) of the TMUX6111, TMUX6112, and TMUX6113 when a $1-V_{RMS}$ signal is applied to the source pin (Sx) of an OFF switch. 图 23 shows the setup used to measure off isolation. Use 公式 2 to compute off isolation.

Overview (接下页)

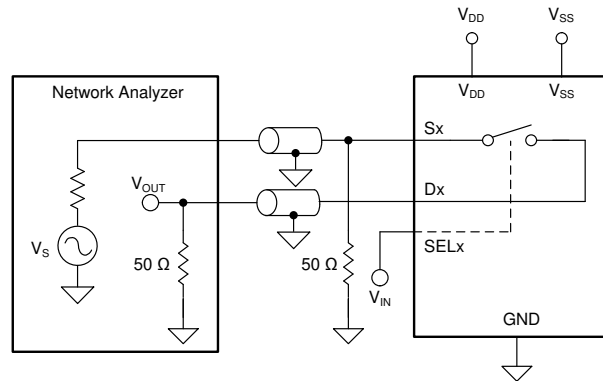


图 23. Off Isolation Measurement Setup

$$\text{Off Isolation} = 20 \cdot \text{Log} \left(\frac{V_{\text{OUT}}}{V_{\text{S}}} \right) \tag{2}$$

9.1.8 Channel-to-Channel Crosstalk

Channel-to-channel crosstalk is defined as the voltage at the source pin (Sx) of an off-channel, when a 1- V_{RMS} signal is applied at the source pin of an on-channel. 图 24 shows the setup used to measure, and 公式 3 is the equation used to compute, channel-to-channel crosstalk.

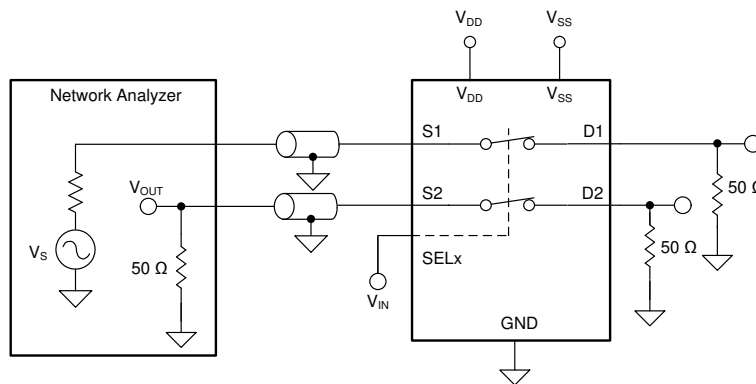


图 24. Channel-to-Channel Crosstalk Measurement Setup

$$\text{Channel-to-Channel Crosstalk} = 20 \cdot \text{Log} \left(\frac{V_{\text{OUT}}}{V_{\text{S}}} \right) \tag{3}$$

9.1.9 Bandwidth

Bandwidth is defined as the range of frequencies that are attenuated by < 3 dB when the input is applied to the source pin (Sx) of an on-channel, and the output is measured at the drain pin (Dx) of the TMUX6111, TMUX6112, and TMUX6113. 图 25 shows the setup used to measure bandwidth of the switch. Use 公式 4 to compute the attenuation.

Overview (接下页)

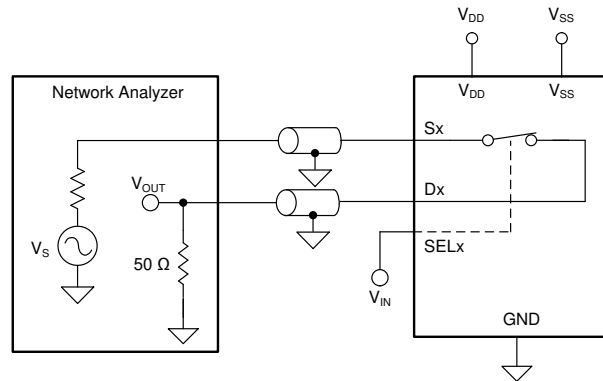


图 25. Bandwidth Measurement Setup

$$\text{Attenuation} = 20 \cdot \text{Log} \left(\frac{V_2}{V_1} \right) \quad (4)$$

9.1.10 THD + Noise

The total harmonic distortion (THD) of a signal is a measurement of the harmonic distortion, and is defined as the ratio of the sum of the powers of all harmonic components to the power of the fundamental frequency at the mux output. The on-resistance of the TMUX6111, TMUX6112, and TMUX6113 varies with the amplitude of the input signal and results in distortion when the drain pin is connected to a low-impedance load. Total harmonic distortion plus noise is denoted as THD+N. 图 26 shows the setup used to measure THD+N of the TMUX6111, TMUX6112, and TMUX6113.

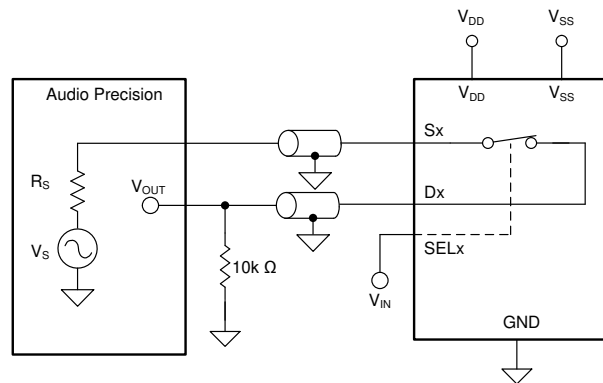
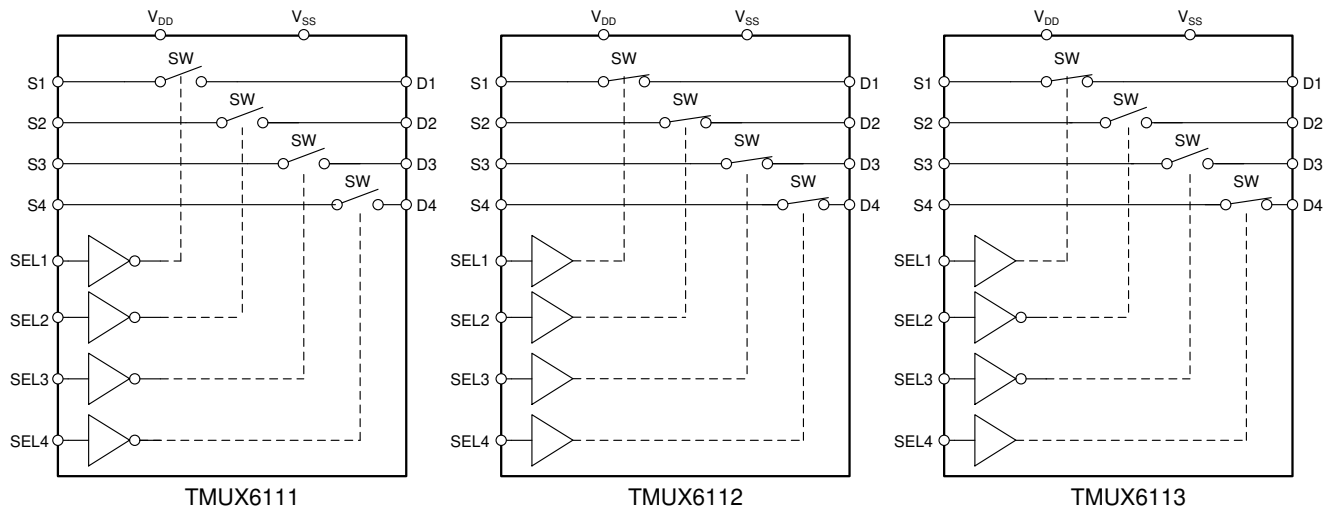


图 26. THD+N Measurement Setup

9.2 Functional Block Diagram



9.3 Feature Description

9.3.1 Ultra-low Leakage Current

The TMUX6111, TMUX6112, and TMUX6113 provide extremely low on- and off-leakage currents. The devices are capable of switching signals from high source-impedance inputs into a high input-impedance op amp with minimal offset error because of the ultralow leakage currents. 图 27 shows typical leakage currents of the devices versus temperature.

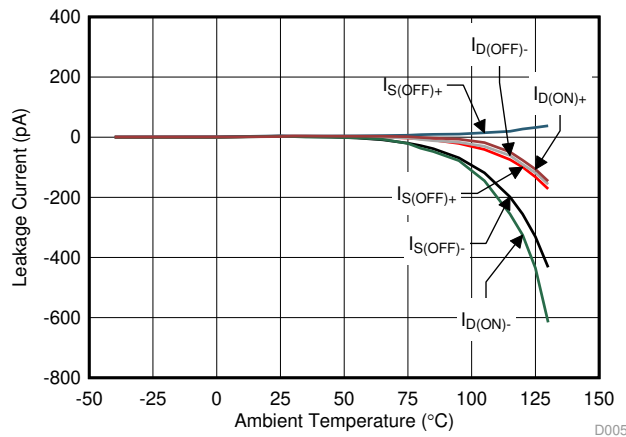


图 27. Leakage Current vs Temperature

9.3.2 Ultra-low Charge Injection

The TMUX6111, TMUX6112, and TMUX6113 are implemented with simple transmission gate topology, as shown in 图 28. Any mismatch in the stray capacitance associated with the NMOS and PMOS causes an output level change whenever the switch is opened or closed. The devices utilize special charge-injection cancellation circuitry that reduces the source (Sx)-to-drain (Dx) charge injection to as low as 0.6 pC at $V_S = 0$ V, as shown in 图 29.

Feature Description (接下页)

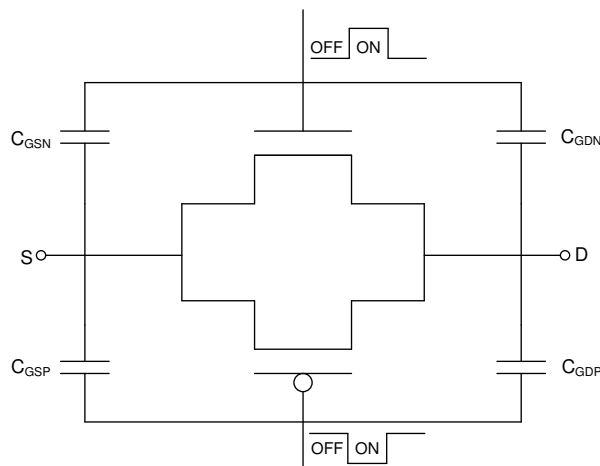


图 28. Transmission Gate Topology

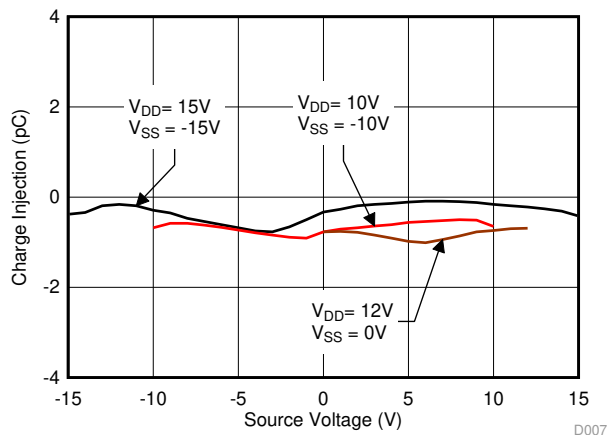


图 29. Source-to-Drain Charge Injection vs Source or Drain Voltage

9.3.3 Bidirectional and Rail-to-Rail Operation

The TMUX6111, TMUX6112, and TMUX6113 conduct equally well from source (Sx) to drain (Dx) or from drain (Dx) to source (Sx). Each channel of the switches has very similar characteristics in both directions. The input signal to the devices swings from V_{SS} to V_{DD} without any significant degradation in performance. The on-resistance of these devices varies with input signal.

9.4 Device Functional Modes

Each channel of the TMUX6111, TMUX6112, and TMUX6113 is turned on or turned off based on the state of its corresponding SELx pin. The SELx pins are weakly pulled-down through an internal 6 MΩ resistor, allowing the switches to stay in a determined state when power is applied to the devices. The SELx pins can be connected to V_{DD} .

10 Application and Implementation

注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

The TMUX6111, TMUX6112, and TMUX6113 offer outstanding input/output leakage currents and ultralow charge injection. These devices operate up to 34 (dual supply) or 17 V (single supply), and offer true rail-to-rail input and output. The on-capacitance of the TMUX6111, TMUX6112, and TMUX6113 is low. These features makes the TMUX6111, TMUX6112, and TMUX6113 a family of precision, robust, high-performance analog multiplexer for high-voltage, industrial applications.

10.2 Typical Application

One useful application to take advantage of TMUX6111, TMUX6112, and TMUX6113's precision performance is the sample and hold circuit. A sample and hold circuit can be useful for an analog to digital converter (ADC) to sample a varying input voltage with improved reliability and stability. It can also be used to store the output samples from a single digital-to-analog converter (DAC) in a multi-output application. A simple sample and hold circuit can be realized using an analog switch like one of the TMUX6111, TMUX6112, and TMUX6113 analog switches.

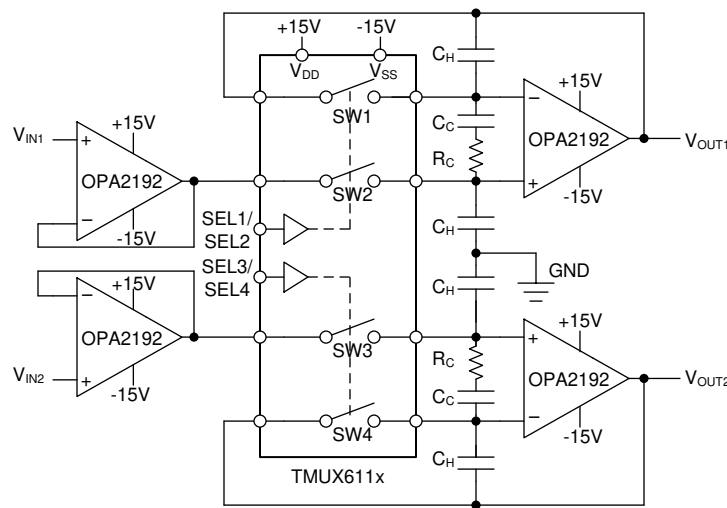


图 30. A 2-output Sample and Hold Circuit Realized Using the TMUX611x Analog Switch

Typical Application (接下页)

10.2.1 Design Requirements

The purpose of this precision design is to implement an optimized 2-output sample and hold circuit using a 4-channel SPST switch. The sample and hold circuit needs to be capable of supporting high voltage output swing up to $\pm 15\text{V}$ with minimized pedestal error and fast settling time. The overall system block diagram is illustrated in 图 30.

10.2.2 Detailed Design Procedure

The TMUX6111, TMUX6112, or TMUX6113 switch is used in conjunction with the voltage holding capacitors (C_H) to implement the sample and hold circuit. The basic operation is:

1. When the switch (SW2 or SW3) is closed, it samples the input voltage and charges the holding capacitors (C_H) to the input voltages values.
2. When the switch (SW2 or SW3) is open, the holding capacitors (C_H) holds its previous value, maintaining stable voltage at the amplifier output (V_{OUT}).

Ideally, the switch delivers only the input signals to the holding capacitors. However, when the switch gets toggled, some amount of charge also gets transferred to the switch output in the form of charge injection, resulting slight sampling error. The TMUX6111, TMUX6112, and TMUX6113 switches have excellent charge injection performance of only 0.6 pC, making them ideal choices for this implementation to minimize sampling error.

Due to switch and capacitor leakage current, the voltage on the hold capacitors droops with time. The TMUX6111, TMUX6112, and TMUX6113 minimize the droops due to its ultra-low leakage performance. At 25°C, the TMUX6111, TMUX6112, and TMUX6113 have extremely tiny leakage current at 1 pA typical and 20 pA max.

The TMUX6111, TMUX6112, and TMUX6113 devices also support high voltage capability. The devices support up to $\pm 17\text{V}$ dual supply operation, making it an ideal solution in this high voltage sample and hold application.

A second switch SW1 (or SW4) is also included to operate in parallel with SW2 (or SW3) to reduce pedestal error during switch toggling. Because both switches are driven at the same potential, they act as common-mode signal to the op-amp, thereby minimizing the charge injection effects caused by the switch toggling action. Compensation network consisting of R_C and C_C is also added to further reduce the pedestal error, whiling reducing the hold-time glitch and improving the settling time of the circuit.

10.3 Application Curves

TMUX6111, TMUX6112, and TMUX6113 have excellent charge injection performance of only 0.6 pC (typical), making them ideal choices to minimize sampling error for the sample and hold application. 图 31 shows the plot for the charge injection vs. source input voltage for TMUX6111, TMUX6112, and TMUX6113.

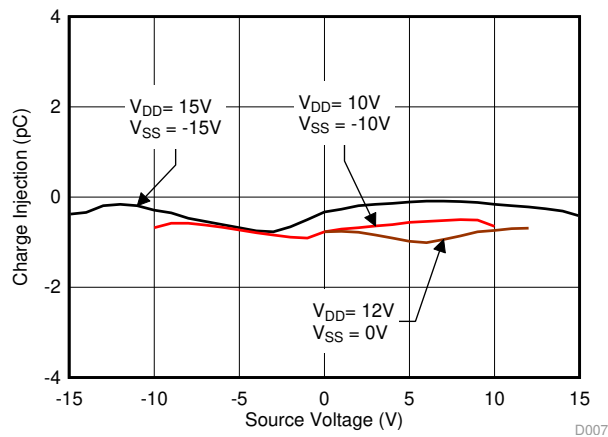


图 31. Charge injection vs. Source Voltage for TMUX6111, TMUX6112 and TMUX6113

11 Power Supply Recommendations

The TMUX6111, TMUX6112, and TMUX6113 operate across a wide supply range of ± 5 V to ± 17 V (10 V to 17 V in single-supply mode). They also perform well with asymmetrical supplies such as $V_{DD} = 12$ V and $V_{SS} = -5$ V. For improved supply noise immunity, use a supply decoupling capacitor ranging from 0.1 μ F to 10 μ F at both the V_{DD} and V_{SS} pins to ground. Always ensure the ground (GND) connection is established before supplies are ramped. As a best practice, it is recommended to ramp V_{SS} first before V_{DD} in dual or asymmetrical supply applications.

The on-resistance of the devices varies with supply voltage, as illustrated in [Figure 32](#)

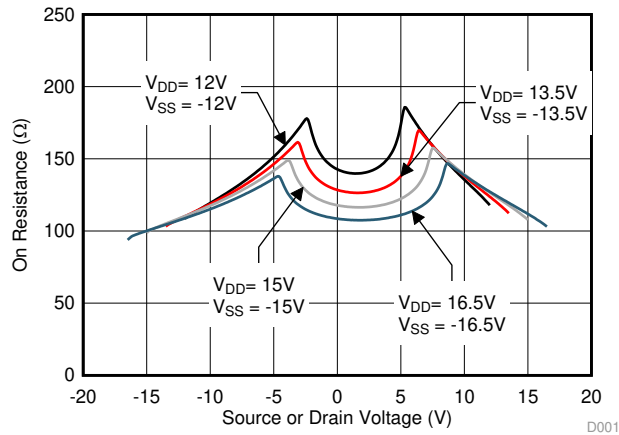


Figure 32. On-Resistance Variation With Supply and Input Voltage

12 Layout

12.1 Layout Guidelines

图 33 illustrates an example of a PCB layout with the TMUX6112PW. Some key considerations are:

- Decouple the V_{DD} and V_{SS} pins with a 0.1- μF capacitor, placed as close to the pin as possible. Make sure that the capacitor voltage rating is sufficient for the V_{DD} and V_{SS} supplies.
- Keep the input lines as short as possible.
- Use a solid ground plane to help distribute heat and reduce electromagnetic interference (EMI) noise pickup.
- Do not run sensitive analog traces in parallel with digital traces. Avoid crossing digital and analog traces if possible, and only make perpendicular crossings when necessary.

12.2 Layout Example

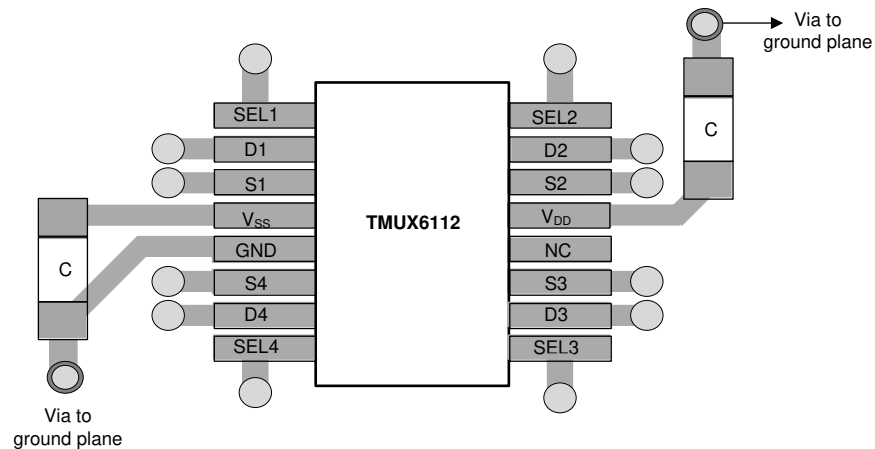


图 33. TMUX6112PW Layout Example

13 器件和文档支持

13.1 文档支持

13.1.1 相关文档

- 《采用 e-trim™ 技术的 OPAx192 36V、精密、轨到轨输入/输出、低偏移电压、低输入偏置电流运算放大器》(SBOS620E)

13.2 相关链接

下表列出了快速访问链接。类别包括技术文档、支持和社区资源、工具和软件，以及立即订购快速访问。

表 4. 相关链接

器件	产品文件夹	立即订购	技术文档	工具与软件	支持和社区
TMUX6111	请单击此处	请单击此处	请单击此处	请单击此处	请单击此处
TMUX6112	请单击此处	请单击此处	请单击此处	请单击此处	请单击此处
TMUX6113	请单击此处	请单击此处	请单击此处	请单击此处	请单击此处

13.3 接收文档更新通知

要接收文档更新通知，请导航至 ti.com 上的器件产品文件夹。单击右上角的通知我进行注册，即可每周接收产品信息更改摘要。有关更改的详细信息，请查看任何已修订文档中包含的修订历史记录。

13.4 支持资源

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

13.5 商标

E2E is a trademark of Texas Instruments.

13.6 静电放电警告



ESD 可能会损坏该集成电路。德州仪器 (TI) 建议通过适当的预防措施处理所有集成电路。如果不遵守正确的处理措施和安装程序，可能会损坏集成电路。

ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

13.7 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

14 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请查阅左侧的导航栏。

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TMUX6111PWR	ACTIVE	TSSOP	PW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	MUX6111	Samples
TMUX6111RTER	ACTIVE	WQFN	RTE	16	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TM6111	Samples
TMUX6112PWR	ACTIVE	TSSOP	PW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	MUX6112	Samples
TMUX6112RTER	ACTIVE	WQFN	RTE	16	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TM6112	Samples
TMUX6113PWR	ACTIVE	TSSOP	PW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	MUX6113	Samples
TMUX6113RTER	ACTIVE	WQFN	RTE	16	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TM6113	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TMUX6111PWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TMUX6111RTER	WQFN	RTE	16	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TMUX6112PWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TMUX6112RTER	WQFN	RTE	16	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TMUX6113PWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TMUX6113RTER	WQFN	RTE	16	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

TAPE AND REEL BOX DIMENSIONS



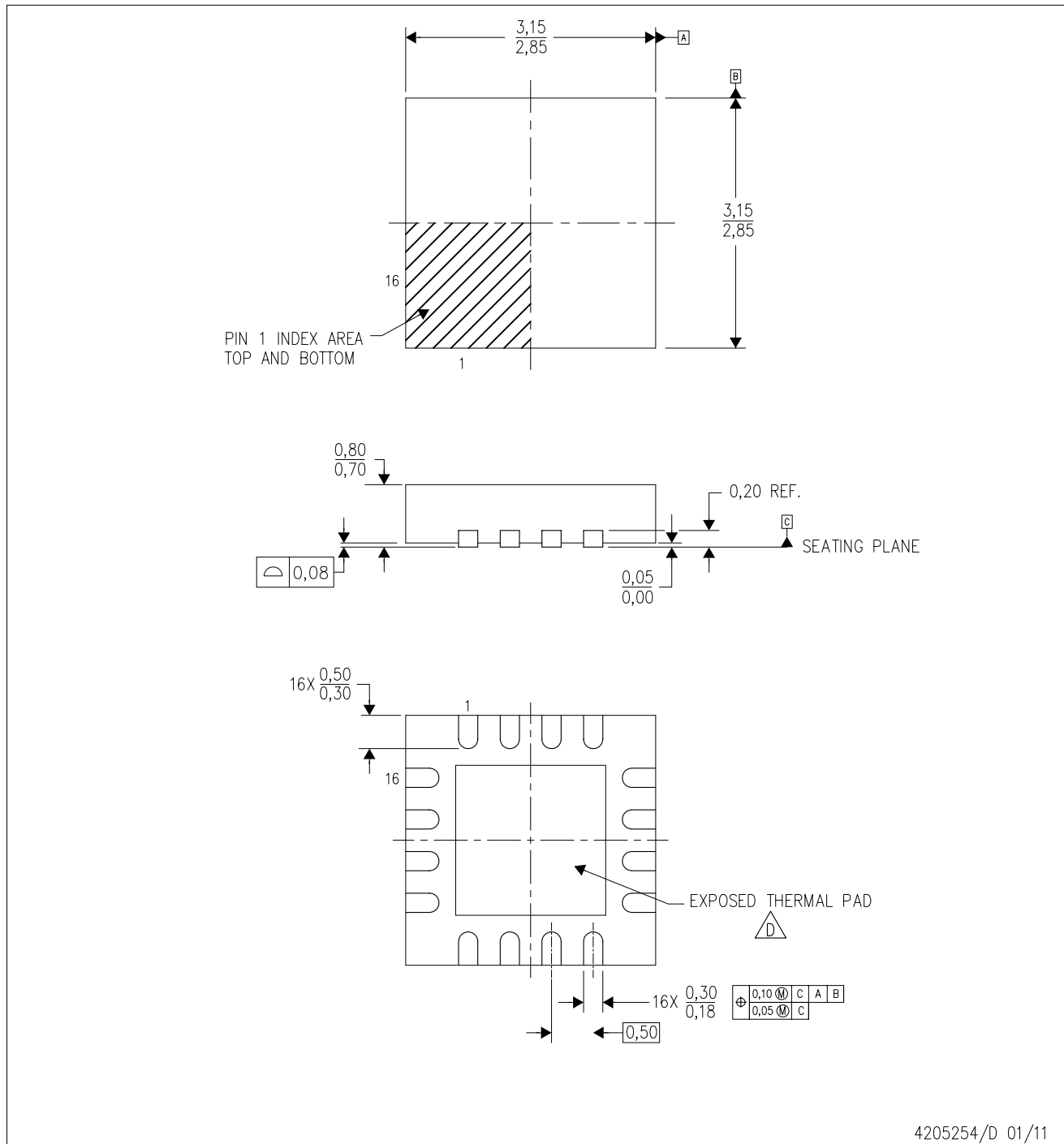
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TMUX6111PWR	TSSOP	PW	16	2000	853.0	449.0	35.0
TMUX6111RTER	WQFN	RTE	16	3000	367.0	367.0	35.0
TMUX6112PWR	TSSOP	PW	16	2000	853.0	449.0	35.0
TMUX6112RTER	WQFN	RTE	16	3000	367.0	367.0	35.0
TMUX6113PWR	TSSOP	PW	16	2000	367.0	367.0	35.0
TMUX6113RTER	WQFN	RTE	16	3000	367.0	367.0	35.0


MECHANICAL DATA

RTE (S-PWQFN-N16)

PLASTIC QUAD FLATPACK NO-LEAD



4205254/D 01/11

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. Quad Flatpack, No-leads (QFN) package configuration.
 -  The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
 - E. Falls within JEDEC MO-220.

THERMAL PAD MECHANICAL DATA

RTE (S-PWQFN-N16)

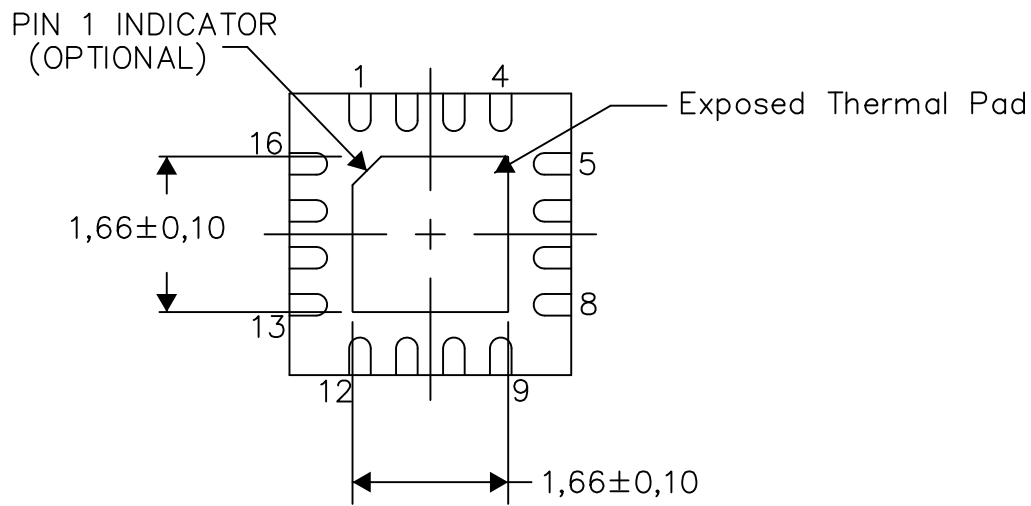
PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

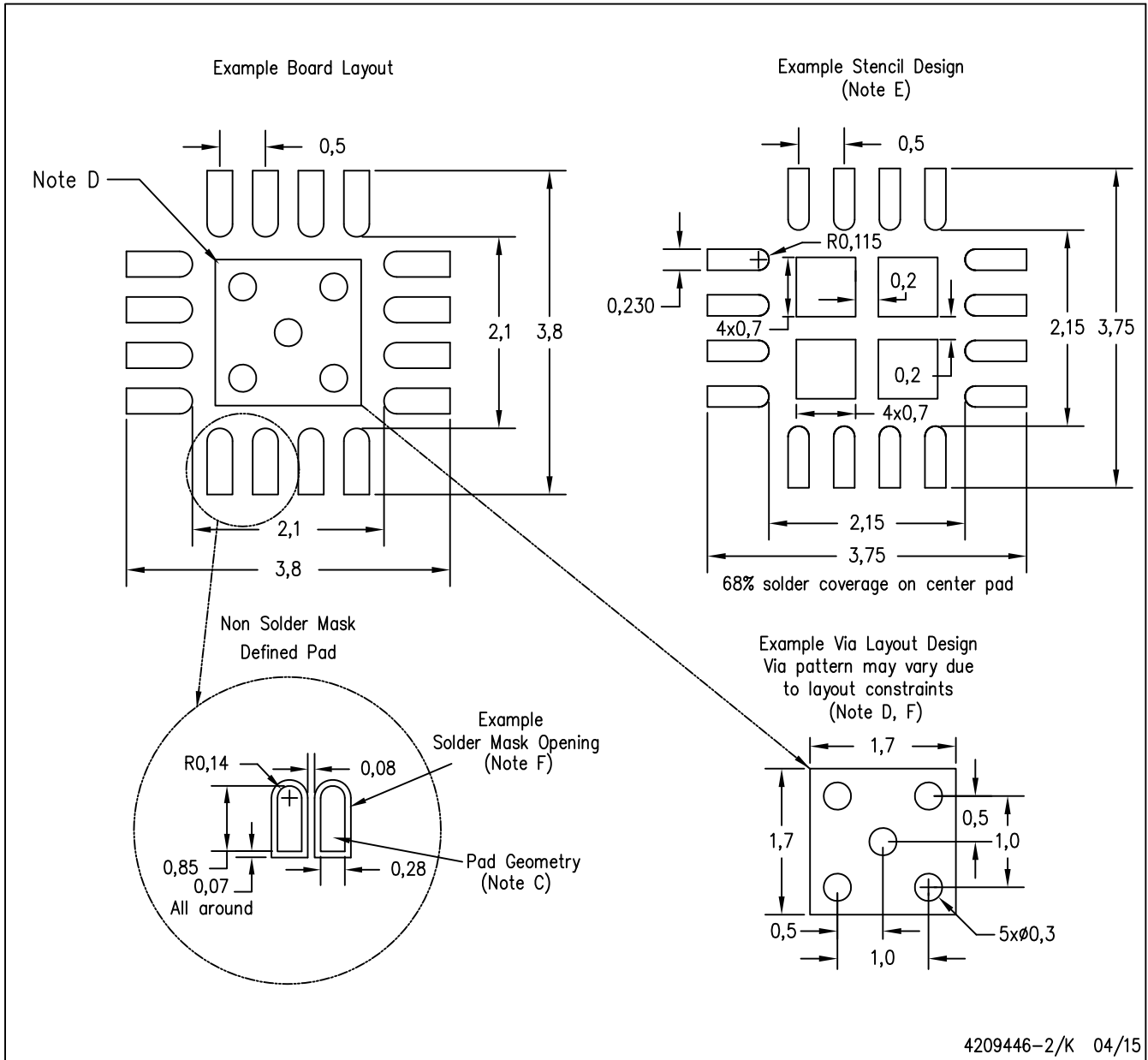


Bottom View

Exposed Thermal Pad Dimensions

4206446-8/U 08/15

NOTE: A. All linear dimensions are in millimeters



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Publication IPC-7351 is recommended for alternate designs.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
 - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - F. Customers should contact their board fabrication site for minimum solder mask web tolerances between signal pads.



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NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-153.

EXAMPLE BOARD LAYOUT

PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 10X



SOLDER MASK DETAILS

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NOTES: (continued)

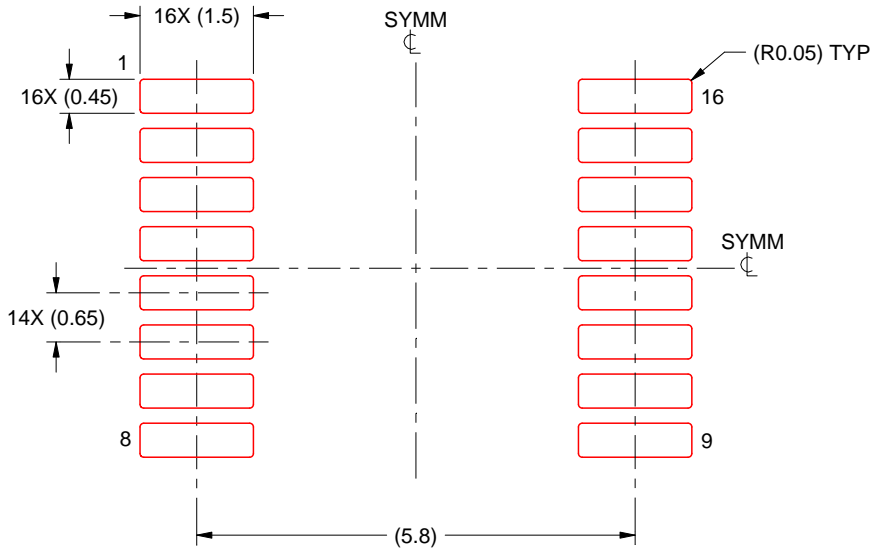
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

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NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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