

# 3 Amp Positive Adjustable Regulator

### **FEATURES**

- Guaranteed 1% Initial Voltage Tolerance
- Guaranteed 3A Output Current
- Guaranteed 0.3% Load Regulation
- Guaranteed 0.01%/V Line Regulation
- 100% Thermal Limit Burn-in

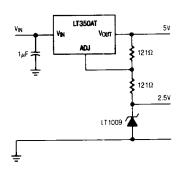
## **APPLICATIONS**

- Improved Linear Regulators
- Adjustable Power Supplies
- Constant Current Regulation
- Battery Chargers

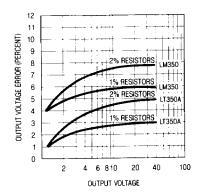
## DESCRIPTION

The LT150A Series are 3-terminal positive adjustable voltage regulators which offer improved performance over earlier devices. A major feature of the LT150A/350A is the output voltage tolerance is guaranteed at a maximum of  $\pm$  1%, allowing an overall power supply tolerance to be better than 3% using inexpensive 1% resistors. Line and load regulation performance has been improved as well. Additionally, the LT150A/350A reference voltage is guaranteed not to exceed 2% when operating over the full load, line and power dissipation conditions. The LT150A/350A adjustable regulators offer an improved solution for all positive voltage regulator requirements with load currents up to 3 amps.

#### Regulator With Reference



#### **Output Voltage Error**

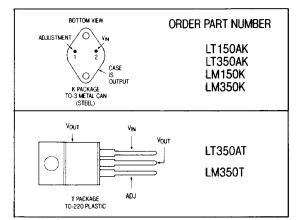




## **ABSOLUTE MAXIMUM RATINGS**

Power Dissipation Internally Limited Input to Output Voltage Differential 35V
Operating Junction Temperature Range
LT150A/LM15055°C to 150°C
LT350A/LM350 0°C to 125°C
Storage Temperature Range
LT150A/LM15065°C to 150°C
LT350A/LM35065°C to 150°C
Lead Temperature (Soldering, 10 sec.) 300°C

# PACKAGE/ORDER INFORMATION



## PRECONDITIONING:

100% THERMAL LIMIT BURN-IN

# **ELECTRICAL CHARACTERISTICS (See Note 1)**

SYMBOL	PARAMETER	CONDITIONS		MIN	LT150A TYP	MAX	MIN	LM150 TYP	MAX	UNITS
V <sub>REF</sub>	Reference Voltage	$I_{OUT} = 10$ mA, $T_i = 25$ °C		1.238	1.250	1.262				٧
		$3V \leqslant (V_{\text{IN}} - V_{\text{OUT}}) \leqslant 35V$ $10\text{mA} \leqslant I_{\text{OUT}} \leqslant 3\text{A}, P \leqslant 30\text{W}$	•	1.225	1.250	1,270	1.20	1.25	1.30	٧
ΔV <sub>QUT</sub> ΔV <sub>IN</sub>	Line Regulation	$3\text{V} \leqslant (\text{V}_{\text{IN}} - \text{V}_{\text{OUT}}) \leqslant 35\text{V}, \text{ (See Note 2)}$	•		0.005 0.02	0.01 0.05		0.005 0.02	0.01 0.05	%/V %/V
ΔV <sub>OUT</sub> ΔI <sub>OUT</sub>	Load Regulation	$\begin{array}{ll} \mbox{10mA} \leqslant \mbox{I}_{OUT} \leqslant \mbox{3A, (See Note 2)} \\ \mbox{T}_{A} = 25^{\circ}\mbox{C} \\ \mbox{V}_{OUT} \leqslant 5\mbox{V} \\ \mbox{V}_{OUT} \geqslant 5\mbox{V} \\ \end{array}$			5 0.1	15 0.3		5 0.1	15 0.3	mV %
		$V_{OUT} \le 5V$ $V_{OUT} \ge 5V$	•		15 0.3	50 1		20 0.3	50 1	m∨ %
	Thermal Regulation	T <sub>A</sub> = 25°C, 20msec Pulse			0.002	0.01		0.002	0.01	%/W
	Ripple Rejection	$\begin{array}{c} V_{OUT} = 10V, f = 120Hz \\ C_{ADJ} = 0 \\ C_{ADJ} = 10\mu F \end{array}$	•	66	65 86		66	65 86		dB dB
I <sub>ADJ</sub>	Adjust Pin Current		•		50	100		50	100	μA
$\Delta I_{ADJ}$	Adjust Pin Current Change	$\begin{array}{l} 10\text{mA} \leqslant i_L \leqslant 3\text{A} \\ 3\text{V} \leqslant (\text{V}_{\text{IN}} - \text{V}_{\text{OUT}}) \leqslant 35\text{V} \end{array}$	•		0.2	5		0.2	5	μΑ
	Minimum Load Current	$(V_{IN} - V_{OUT}) = 35V$	•		3.5	5		3.5	5	mA
	Current Limit	$(V_{IN} - V_{OUT}) \leq 10V$	•	3	4.5		3.0	4.5		A
		$(V_{IN} - V_{OUT}) = 30V$	•	0.3	1		0.3	1		A
ΔV <sub>OUT</sub> ΔTemp	Temperature Stability	$-55^{\circ}\text{C} \leqslant \text{T}_{\text{j}} \leqslant +150^{\circ}\text{C}$	•		1	2		1		%
ΔV <sub>out</sub> ΔTime	Long Term Stability	T <sub>A</sub> = 125°C			0.3	1		0.3	1	%
en	RMS Output Noise (% of V <sub>OUT</sub> )	$T_A = 25^{\circ}C$ , $10Hz \leqslant f \leqslant 10kHz$	T		0.001		•	0.001		%
Θ <sub>jc</sub>	Thermal Resistance Junction to Case	K Package				1.5			1.5	°C/W

# **ELECTRICAL CHARACTERISTICS (See Note 1)**

SYMBOL	PARAMETER	CONDITIONS		MIN	LT350A TYP	MAX	MIN	LM350 TYP	MAX	UNITS
V <sub>REF</sub>	Reference Voltage	$I_{OUT}=10$ mA, $T_j=25$ °C		1.238	1.250	1.262				1
		$3V \leqslant (V_{IN} - V_{OUT}) \leqslant 35V$ $10\text{mA} \leqslant I_{OUT} \leqslant 3A, P \leqslant 30W$	•	1.225	1.250	1.270	1.20	1.25	1.30	\
ΔV <sub>QUT</sub> ΔV <sub>IN</sub>	Line Regulation	$ \begin{array}{ll} 3 \text{V} \leqslant (\text{V}_{\text{IN}} - \text{V}_{\text{OUT}}) \leqslant 35 \text{V, (See} \\ \text{Note 2)} \\ \text{T}_{\text{A}} = 25 ^{\circ} \text{C} \end{array} $	•		9.005 0.02	0.01 0.05		0.005 0.02	0.03 0.07	%/\ %/\
7N <sup>O</sup> IZ	Load Regulation	$\begin{array}{l} 10\text{mA} \leqslant I_{\text{OUT}} \leqslant 3\text{A, (See Note 2)} \\ T_{\text{A}} = 25^{\circ}\text{C} \\ V_{\text{OUT}} \leqslant 5\text{V} \\ V_{\text{OUT}} \geqslant 5\text{V} \end{array}$			5 0.1	15 0.3		5 0.1	25 0.5	m\ %
		$V_{OUT} \le 5V$ $V_{OUT} \ge 5V$	••		15 0.3	50 1		20 0.3	70 1.5	m\
	Thermal Regulation	T <sub>A</sub> = 25°C, 20msec Pulse			0.002	0.01		0.002	0.03	%/V
	Ripple Rejection	$V_{OUT}=10V,f=120Hz$ $C_{ADJ}=0$ $C_{ADJ}=10\mu F$	•	66	65 86		66	65 86		dE dE
I <sub>ADJ</sub>	Adjust Pin Current	100	•		50	100		50	100	μΑ
ال <sub>ADJ</sub>	Adjust Pin Current Change	$\begin{array}{l} 10\text{mA} \leqslant I_{\text{OUT}} \leqslant 3\text{A} \\ 3\text{V} \leqslant (\text{V}_{\text{IN}} - \text{V}_{\text{OUT}}) \leqslant 35\text{V} \end{array}$	•		0.2	5		0.2	5	μΑ
	Minimum Load Current	$(V_{IN} - V_{OUT}) \leq 35V$	•		3.5	10		3.5	10	m/
	Current Limit	$(V_{IN} - V_{OUT}) \le 10V$	•	3	4.5		3	4.5		A
		$(V_{IN} - V_{OUT}) = 30V, T_j = 25^{\circ}C$		0.25	1		0.25	1		A
∆V <sub>OUT</sub> ∆Temp	Temperature Stability		•		1	2		1		94
ΔV <sub>OUT</sub> ΔTime	Long Term Stability	$T_A = 125$ °C			0.3	1	_	0.3	1	%
e <sub>n</sub>	RMS Output Noise (% of V <sub>OUT</sub> )	$T_A = 25^{\circ}C$ , $10Hz \le f \le 10kHz$			0.001			0.001		%
$\Theta_{jc}$	Thermal Resistance Junction to Case	K Package T Package		·	1.2 3	1.5 4		1.2	1.5	°C/W

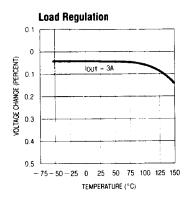
The  $\bullet$  denotes the specifications which apply over the full operating temperature range.

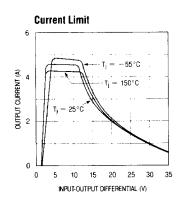
**Note 1:** Unless otherwise specified, these specifications apply for  $V_{\text{IN}} - V_{\text{OUT}} = 5V$  and  $I_{\text{OUT}} = 1.5A$ . These specifications are applicable for power dissipations up to 30W for the K package and up to 25W for the T package. Power dissipation is guaranteed at these values up to 15 Volts input-output differential. Above 15 Volts input-output differential power dissipation is limited by device internal protection circuitry.

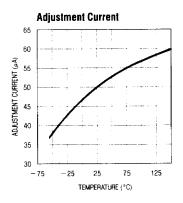
Note 2: Regulation is measured at a constant  $\mathsf{T}_j$ . Changes in output due to heating must be taken into account separately. Pulse testing with low duty cycle is used.

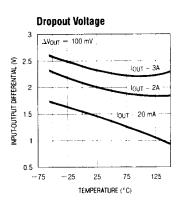


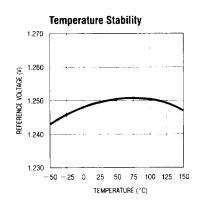
## TYPICAL PERFORMANCE CHARACTERISTICS

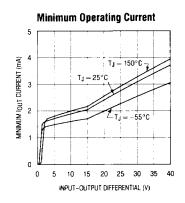


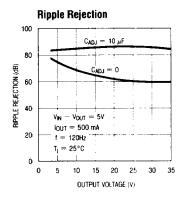


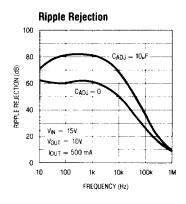


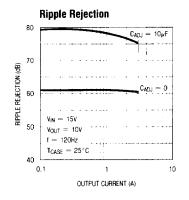




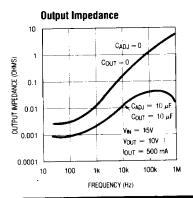


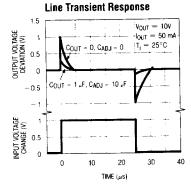


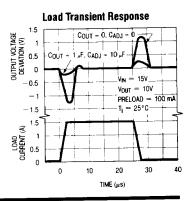




# TYPICAL PERFORMANCE CHARACTERISTICS



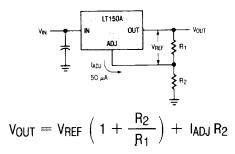




## APPLICATION INFORMATION

#### General

The LT150A develops a 1.25V reference voltage between the output and the adjustable terminal (see Figure 1). By placing a resistor, R1, between these two terminals, a constant current is caused to flow through R1 and down through R2 to set the overall output voltage. Normally this current is the specified minimum load current of 5mA or 10mA.



Basic Adjustable Regulator Figure 1

Because  $I_{ADJ}$  is very small and constant when compared with the current through R1, it represents a small error and can usually be ignored. It is easily seen from the above equation, that even if the resistors were of exact value, the accuracy of the output is limited by the accuracy of  $V_{REF}$ . Earlier adjustable regulators had a reference tolerance of  $\pm 4\%$  which is

dangerously close to the  $\pm$ 5% supply tolerance required in many logic and analog systems. Further, even 1% resistors can drift 0.01%/°C, adding additional error to the output voltage tolerance.

For example, using 2% resistors and  $\pm$  4% tolerance for  $V_{REF}$ , calculations will show that the expected range of a 5V regulator design would be 4.66V  $\leq$   $V_{OUT}$   $\leq$  5.36V or approximately  $\pm$  7%. If the same example were used for a 15V regulator, the expected tolerance would be  $\pm$  8%. With these results most applications required some method of trimming, usually a trim pot. This solution is both expensive and not conducive to volume production.

One of the enhancements of Linear Technology's adjustable regulators over existing devices is the tightened initial tolerance of  $V_{\text{REF}}$ . This allows relatively inexpensive 1% or 2% film resistors to be used for R1 and R2 to set the output voltage within an acceptable tolerance.

With a guaranteed 1% reference, a 5V power supply design, using  $\pm 2\%$  resistors, would have a worst case manufacturing tolerance of  $\pm 4\%$ . If 1% resistors were used, the tolerance would drop to  $\pm 2.5\%$ . A plot of the worst case output voltage tolerance as a function of resistor tolerance is shown on the front page.



For convenience, a table of standard 1% resistor values is shown below.

Table of 1/2% and 1% Standard Resistance Values

1.00	1.47	2.15	3.16	4.64	6.81
1.02	1.50	2.21	3.24	4.75	6.98
1.05	1.54	2.26	3.32	4.87	7.15
1.07	1.58	2.32	3.40	4.99	7.32
1.10	1.62	2.37	3.48	5.11	7.50
1.13	1.65	2.43	3.57	5.23	7.68
1.15	1.69	2.49	3.65	5.36	7.87
1.18	1.74	2.55	3.74	5.49	8.06
1.21	1.78	2.61	3.83	5.62	8.25
1.24	1.82	2.67	3.92	5.76	8.45
1.27	1.87	2.74	4.02	5.90	8.66
1.30	1.91	2.80	4.12	6.04	8.87
1.33	1.96	2.87	4.22	6.19	9.09
1.37	2.00	2.94	4.32	6.34	9.31
1.40	2.05	3.01	4.42	6.49	9.53
1.43	2.10	3.09	4.53	6.65	9.76

Standard Resistance Values are obtained from the Decade Table by multiplying by multiples of 10. As an example, 1.21 can represent 1.21 $\Omega$ , 12.1 $\Omega$ , 121 $\Omega$ , 1.21 $\Omega$ 0 etc.

#### **Bypass Capacitors**

Input bypassing using a  $1\mu F$  tantalum or  $25\mu F$  electrolytic is recommended when the input filter capacitors are more than 5 inches from the device. Improved ripple rejection (80 dB) can be accomplished by adding a  $10\mu F$  capacitor from the adjust pin to ground. Increasing the size of the capacitor to  $20\mu F$  will help ripple rejection at low output voltage since the reactance of this capacitor should be small compared to the voltage setting resistor, R2. For improved AC transient response and to prevent the possibility of oscillation due to unknown reactive load, a  $1\mu F$  capacitor is also recommended at the output. Because of their low impedance at high frequencies, the best type of capacitor to use is solid tantalum.

#### **Protection Diodes**

The LT150A/350A do not require a protection diode from the adjustment terminal to the output (see Figure 2). Improved internal circuitry eliminates the need for this diode when the adjustment pin is bypassed with a capacitor to improve ripple rejection.

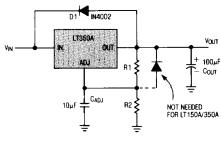


Figure 2

If a very large output capacitor is used, such as a  $100\mu F$  shown in Figure 2, the regulator could be damaged or destroyed if the input is accidentally shorted to ground or crowbarred, due to the output capacitor discharging into the output terminal of the regulator. To prevent this, a diode D1 as shown, is recommended to safely discharge the capacitor.

#### **Load Regulation**

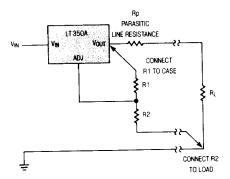
Because the LT150A is a three-terminal device, it is not possible to provide true remote load sensing. Load regulation will be limited by the resistance of the wire connecting the regulator to the load. The data sheet specification for load regulation is measured at the bottom of the package. Negative side sensing is a true Kelvin connection, with the bottom of the output divider returned to the negative side of the load. Although it may not be immediately obvious, best load regulation is obtained when the top of the resistor divider (R1) is connected directly to the case not to the load. This is illustrated in Figure 3. If R1 were connected to the load, the effective resistance between the regulator and the load would be

$$R_p \times \left(\frac{R2 + R1}{R1}\right)$$
,  $R_p = Parasitic Line Resistance$ .

Connected as shown,  $R_p$  is not multiplied by the divider ratio.  $R_p$  is about  $0.004\Omega$  per foot using 16 gauge wire. This translates to 4mV/ft at 1A load current, so it



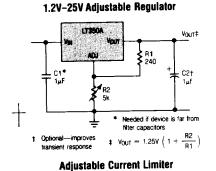
is important to keep the positive lead between regulator and load as short as possible, and use large wire or PC board traces.



Connections for Best Load Regulation Figure 3

## TYPICAL APPLICATIONS



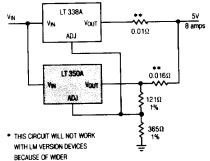


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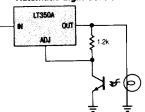
Improving Ripple Rejection

The state of the

#### \*Parallel Regulators for Higher Current



## Automatic Light Control

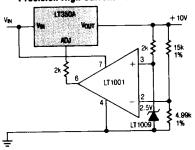


- REFERENCE VOLTAGE LIMITS
  \*\* CURRENT SHARING RESISTO
- \*\* CURRENT SHARING RESISTORS DEGRADE REGULATION TO 1%

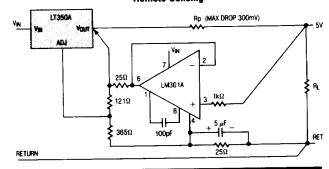
#### Precision High Current Reference

0.5Ω

1ΚΩ 500Ω



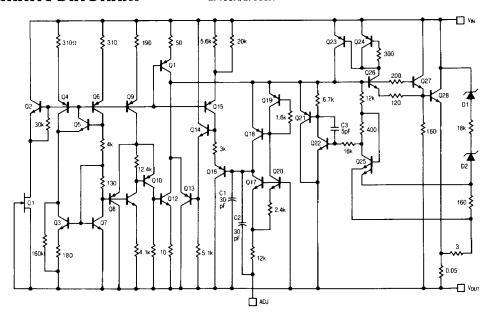
#### Remote Sensing





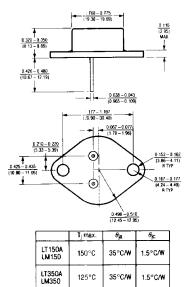
## SCHEMATIC DIAGRAM

#### LT150A/LT350A

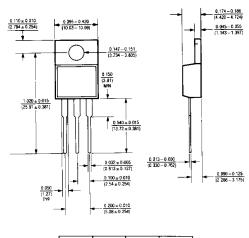


## PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.

K Package TO-3 STEEL Metal Can



T Package TO-220 Plastic



	T <sub>j</sub> max.	$\theta_{\mathrm{ja}}$	$\theta_{\text{JC}}$
LT350A LM350	125°C	50°C/W	2.5°C/W