

Substrate

FAIRCHILD

A Schlumberger Company

μ A78H12A 5-Amp Voltage Regulator

Hybrid Products

Description

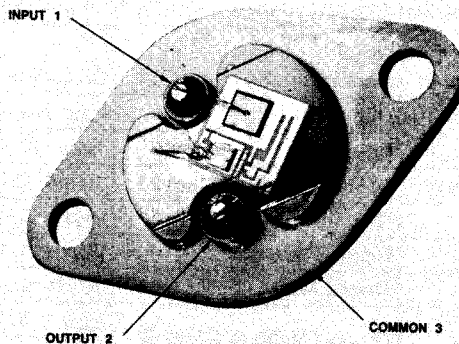
The μ A78H12A is a hybrid regulator with 12.0 V fixed output and 5.0 A output capability. It has the inherent characteristics of the monolithic 3-terminal regulators; i.e., full thermal overload, short-circuit and safe-area protection. All devices are packaged in hermetically sealed TO-3s providing 50 W power dissipation. If the safe operating area is exceeded, the device shuts down, rather than failing or damaging other system components (Note 1). This feature eliminates costly output circuitry and overly conservative heat sinks typical of high-current regulators built from discrete components.

- 5.0 A OUTPUT CURRENT
- INTERNAL CURRENT AND THERMAL OVERLOAD PROTECTION
- INTERNAL SHORT CIRCUIT PROTECTION
- LOW DROPOUT VOLTAGE (TYPICALLY 2.3 V @ 5.0 A)
- 50 W POWER DISSIPATION
- STEEL TO-3 PACKAGE

Note

1. This voltage regulator offers output transistor safe-area protection. However, to maintain full protection, the device must be operated within the maximum input-to-output voltage differential ratings, as listed on this data sheet under "Absolute Maximum Ratings." For applications violating these limits, device will not be fully protected.

Connection Diagram TO-3 Metal Package



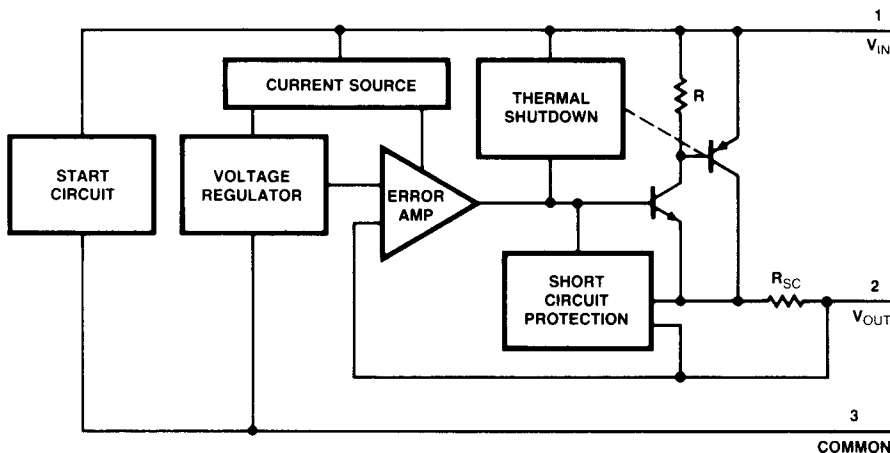
(Top View)

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Order Information

Type	Package	Code	Part No.
μ A78H12A	Metal	GN	μ A78H12ASC
μ A78H12A	Metal	GN	μ A78H12ASM

Block Diagram



μA78H12A

Absolute Maximum Ratings

Input Voltage	40 V	Commercial Temperature Range	0°C to +150°C
Input-to-Output Voltage Differential, Output Short-Circuited	35 V	Storage Temperature Range	-55°C to +150°C
Internal Power Dissipation	50 W @ 25°C Case	Pin Temperature (Soldering, 60 s)	300°C
Operating Junction Temperature	150°C		
Military Temperature Range μA78H12ASM	-55°C to +150°C		

μA7812A

Electrical Characteristics $T_J = 25^\circ\text{C}$, $V_{IN} = 19\text{ V}$, $I_{OUT} = 2.0\text{ A}$ unless otherwise specified

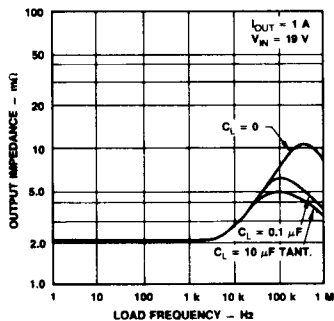
Symbol	Characteristic	Condition	Limits			Unit
			Min	Typ	Max	
V_{OUT}	Output Voltage	$I_{OUT} = 2.0\text{ A}$	11.5	12	12.5	V
ΔV_{OUT}	Line Regulation (Note 2)	$V_{IN} = 16\text{ to }25\text{ V}$		20	120	mV
ΔV_{OUT}	Load Regulation (Note 2)	$10\text{ mA} \leq I_{OUT} \leq 5.0\text{ A}$		20	120	mV
I_Q	Quiescent Current	$I_{OUT} = 0$, $V_{IN} = 17\text{ V}$		3.7	10	mA
RR	Ripple Rejection	$I_{OUT} = 1.0\text{ A}$, $f = 120\text{ Hz}$, 5.0 V_{pk-pk}	60			dB
V_n	Output Noise	$10\text{ Hz} \leq f \leq 100\text{ kHz}$, $V_{IN} = 17\text{ V}$		75		V_{RMS}
V_{DD}	Dropout Voltage (Note 3)	$I_{OUT} = 5.0\text{ A}$		2.3	2.5	V
		$I_{OUT} = 3.0\text{ A}$		2.0	2.3	V
I_{OS}	Short-Circuit Current Limit			7.0	12.0	A _{pk}

Notes

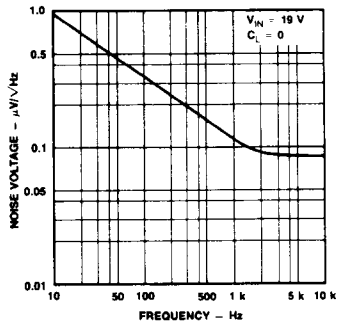
- Load and line regulation are specified at constant junction temperature. Pulse testing is required with a pulse width $\leq 1\text{ ms}$ and a duty cycle $\leq 5\%$. Full Kelvin connection methods must be used to measure these parameters.
- Dropout Voltage is the input-to-output voltage differential that causes the output voltage to decrease by 5% of its initial value.

Typical Performance Curves

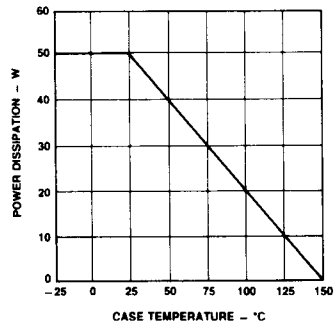
Output Impedance



Output Noise Voltage

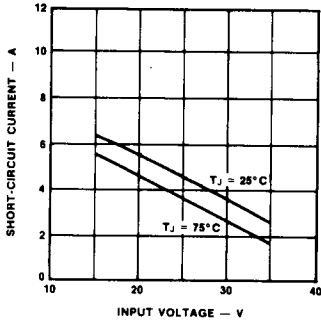


Maximum Power Dissipation

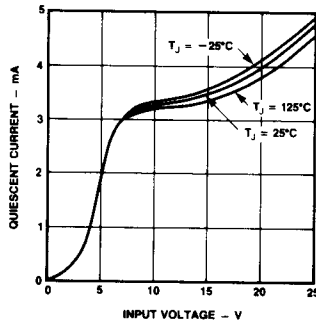


Typical Performance Curves (Cont.)

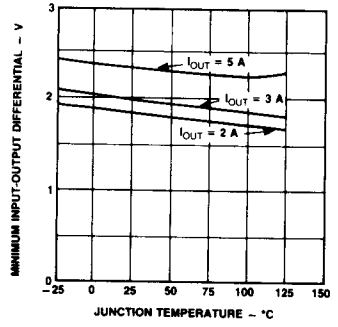
Short Circuit Current



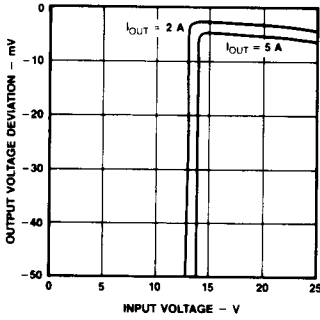
Quiescent Current



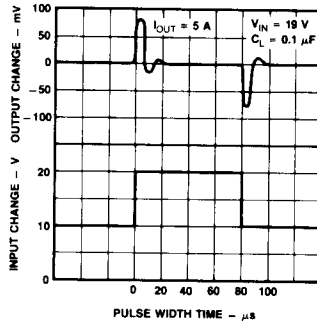
Dropout Voltage



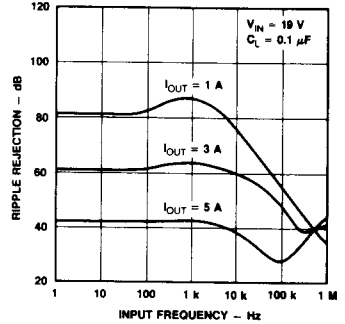
Line Regulation



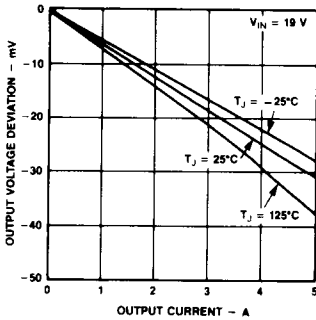
Line Transient Response



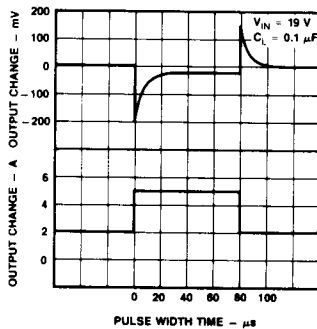
Ripple Rejection



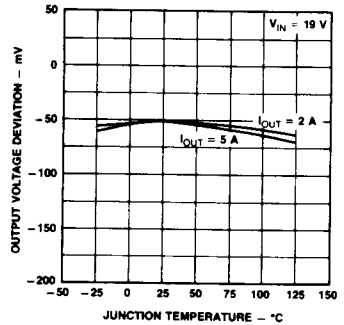
Load Regulation



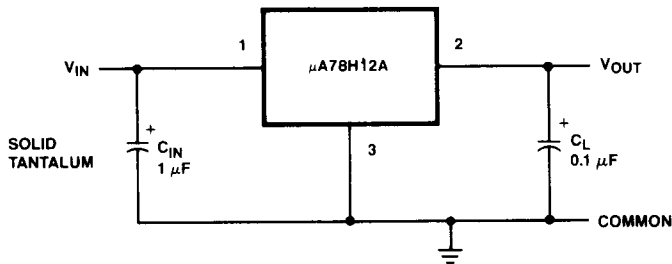
Load Transient Response



Output Voltage Deviation vs Junction Temperature



Basic Test Circuit



Design Considerations

This device has thermal-overload protection from excessive power and internal short-circuit protection which limits the circuit's maximum current. Thus, the device is protected from overload abnormalities. Although the internal power dissipation is limited, the junction temperature must be kept below the maximum specified temperature (150°C). It is recommended by the manufacturer that the maximum junction temperature be kept as low as possible for increased reliability. To calculate the maximum junction temperature or heat sink required, the following thermal resistance values should be used:

Package	Typ θ_{JC}	Max θ_{JC}
TO-3	1.8	2.5

$$P_{D(max)} = \frac{T_{J(max)} - T_A}{\theta_{JC} + \theta_{CA}}$$

$$\theta_{CA} = \theta_{CS} + \theta_{SA}$$

Solving for T_J :

$$T_J = T_A + P_D (\theta_{JC} + \theta_{JA})$$

Where:

T_J = Junction Temperature

T_A = Ambient Temperature

P_D = Power Dissipation

θ_{JC} = Junction-to-case thermal resistance

θ_{CA} = Case-to-ambient thermal resistance

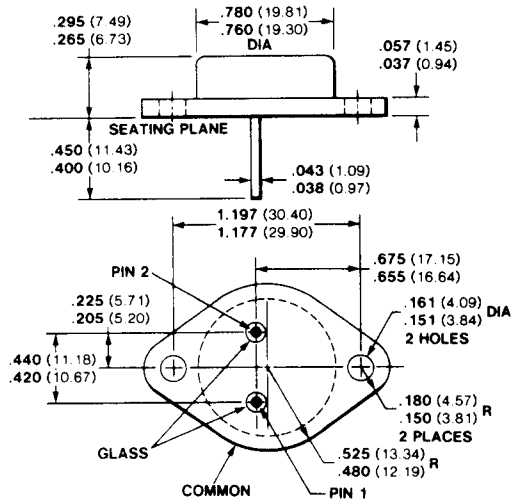
θ_{CS} = Case-to-heat sink thermal resistance

θ_{SA} = Heat sink-to-ambient thermal resistance

The devices are designed to operate without external compensation components. However, the amount of external filtering of these voltage regulators depends upon the circuit layout. If in a specific application the regulator is more than four inches from the filter capacitor, a 1 μF solid tantalum capacitor should be used at the input. A 0.1 μF capacitor should be used at the output to reduce transients created by fast switching loads, as seen in the basic test circuit. These filter capacitors must be located as close to the regulator as possible.

Caution: Permanent damage can result from forcing the output voltage higher than the input voltage. A protection diode from output to input should be used if this condition exists.

Package Outline (S Package — Steel)



Notes

All dimensions in inches bold and millimeters (parentheses)
Pins are solder-dipped alloy 52