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April 1<sup>st</sup>, 2010 Renesas Electronics Corporation

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**Quadruple Comparators** 

REJ03D0806-0100 Rev.1.00 Mar 10, 2006

# Description

The HA17901A series products are comparators designed for general purpose, especially for power control systems.

These ICs operate from a single power-supply voltage over a wide range of voltages, and feature a reduced powersupply current since the supply current is independent of the supply voltage.

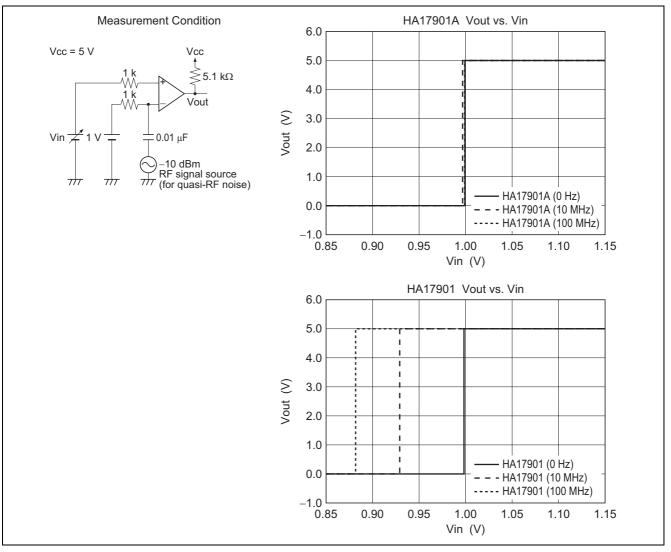
These comparators have the merit which ground is included in the common-mode input voltage range at a single-voltage power supply operation. These products have a wide range of applications, including limit comparators, simple A/D converters, pulse/square-wave/time delay generators, wide range VCO circuits, MOS clock timers, multivibrators, and high-voltage logic gates.

# Features

- Wide power-supply voltage range : 2 to 36 V
- Very low supply current : 0.8 mA Typ.
- Low input bias current : 25 nA Typ.
- Low input offset current : 5 nA Typ.
- Low input offset voltage : 2 mV Typ.
- The common-mode input voltage range includes ground
- Output voltages compatible with CMOS logic systems



• Low electro-magnetic susceptibility

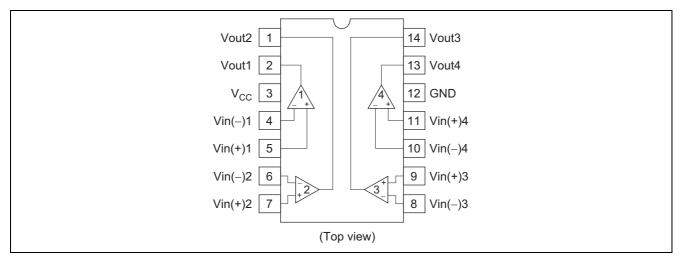


# **Ordering Information**

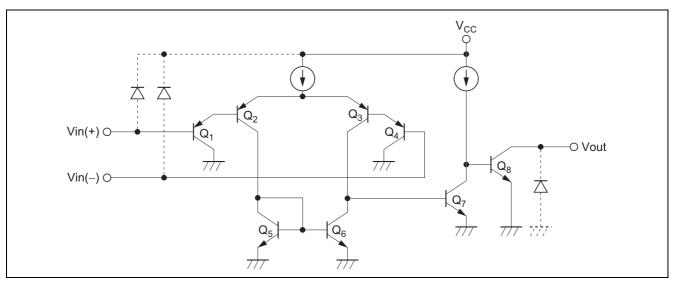
Type No.	Application	Package Name	Package Code
HA17901AP	Industry use	DIP-14 pin	PRDP0014AB-B
HA17901AFP		SOP-14 pin (JEITA)	PRSP0014DF-B
HA17901ARP		SOP-14 pin (JEDEC)	PRSP0014DE-A
HA17901AT		TSSOP-14 pin	PTSP0014JA-B



# **Pin Arrangement**



# **Circuit Structure** (1/4)



Note: If Input/Output terminals voltage over the absolute maximum ratings, there is possibility of mis-operation, characteristics deterioration and destruction, because of the current's flowing to parasitic diode in IC. The Input/Output terminals are recommended to be protected with the clamp circuit which using the diode with low forward voltage (like schottky barrier diode) when there is a possibility for the Input/Output terminals voltage exceeds the absolute maximum ratings.



# **Absolute Maximum Ratings**

				$(1a = 25^{\circ}C)$
ltem		Symbol	Ratings	Unit
Power supply voltage		V <sub>CC</sub>	36	V
Differential input voltage		Vin(diff)	±V <sub>CC</sub>	V
Input voltage		Vin	–0.3 to +V <sub>CC</sub>	V
Output pin voltage		Vout	–0.3 to +36	V
Output current		lout *1	20	mA
Allowable power dissipation	DIP	PT	625 * <sup>2</sup>	mW
	SOP		625 * <sup>3</sup>	
	TSSOP		400 *4	
Operating temperature		Topr	-40 to +85	°C
Storage temperature		Tstg	–55 to +125	°C

Notes: 1. These products can be destroyed if the output and V<sub>CC</sub> are shorted together. The maximum output current is the allowable value for continuous operation.

These are the allowable values up to  $Ta = 50^{\circ}C$ . Derate by 8.3 mW/°C above that temperature. 3. HA17901AFP/ARP:

When it is mounted on glass epoxy board of 40 mm  $\times$  40 mm  $\times$  1.6 mmt with 10% wiring density, value at Ta  $\leq$  25°C. If Ta > 25°C, derated by 6.25 mW/°C.

When it is mounted on glass epoxy board of 40 mm  $\times$  40 mm  $\times$  1.6 mmt with 30% wiring density. If Ta > 32°C, derated by 6.70 mW/°C.

4. HA17901AT:

These are the allowable values up to  $Ta = 25^{\circ}C$ . Derate by 4 mW/ $^{\circ}C$  above that temperature.

# **Electrical Characteristics**

 $(V_{CC} = 5 V, Ta = 25^{\circ}C)$ 

 $(T_{0} - 25^{\circ}C)$ 

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Input offset voltage	V <sub>IO</sub>	_	2	7	mV	Output switching point:
						when $V_0 = 1.4V$ , $R_S = 0\Omega$
Input offset current	I <sub>IO</sub>		5	50	nA	$  I_{IN(+)} - I_{IN(-)}  $
Input bias current	I <sub>IB</sub>		25	250	nA	I <sub>IN(+)</sub> or I <sub>IN(-)</sub>
Common-mode input voltage *1	V <sub>CM</sub>	0	—	V <sub>CC</sub> -1.5	V	
Supply current	Icc	_	0.8	2	mA	R <sub>L</sub> = ∞
Voltage Gain * <sup>3</sup>	Av	_	(200)	_	V/mV	$R_L = 15k\Omega$
Response time * <sup>2,3</sup>	t <sub>R</sub>	_	(1.3)	_	μs	$V_{RL} = 5V, R_L = 5.1k\Omega$
Output sink current	I <sub>O(sink)</sub>	6	16		mA	$V_{IN(-)} = 1V,  V_{IN(+)} = 0,  V_O \le 1.5V$
Output saturation voltage	V <sub>O(sat)</sub>		200	400	mV	$V_{IN(-)} = 1V, V_{IN(+)} = 0, Iosink = 3mA$
Output leakage current *3	I <sub>LO</sub>		(0.1)		nA	$V_{IN(+)} = 1V, V_{IN(-)} = 0, V_O = 5V$

Notes: 1. Voltages more negative than –0.3 V are not allowed for the common-mode input voltage or for either one of the input signal voltages.

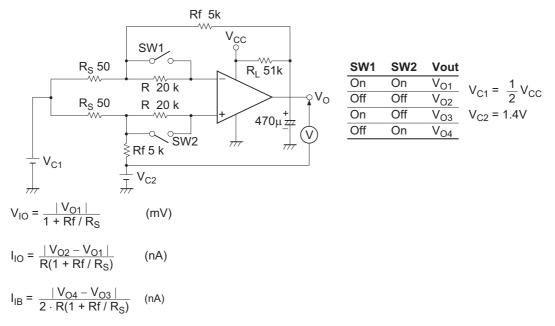
2. The stipulated response time is the value for a 100 mV input step voltage that has a 5 mV overdrive.

3. Design spec.

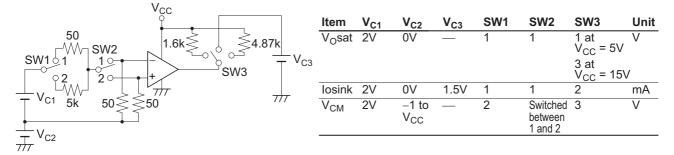
<sup>2.</sup> HA17901AP:

# **Test Circuits**

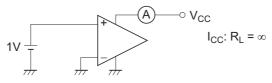
1. Input offset voltage (V<sub>IO</sub>), input offset current (I<sub>IO</sub>), and Input bias current (I<sub>IB</sub>) test circuit



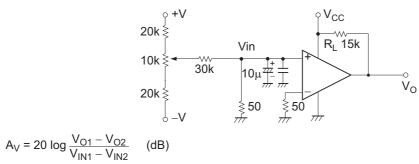
2. Output saturation voltage (Vo sat) output sink current (Iosink), and common-mode input voltage (Vom) test circuit



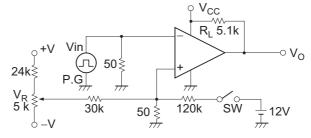
3. Supply current (I<sub>CC</sub>) test circuit



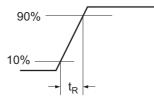
4. Voltage gain (A<sub>V</sub>) test circuit ( $R_L = 15 \text{ k}\Omega$ )



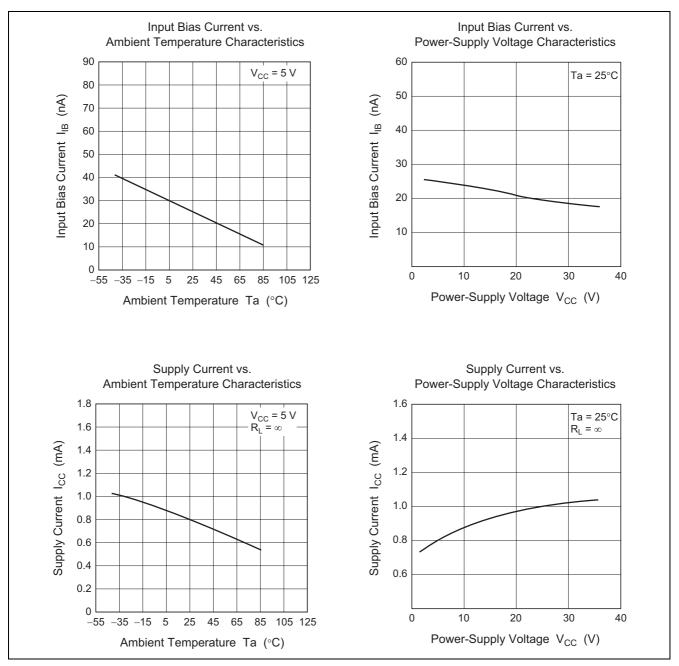
5. Response time  $(t_R)$  test circuit

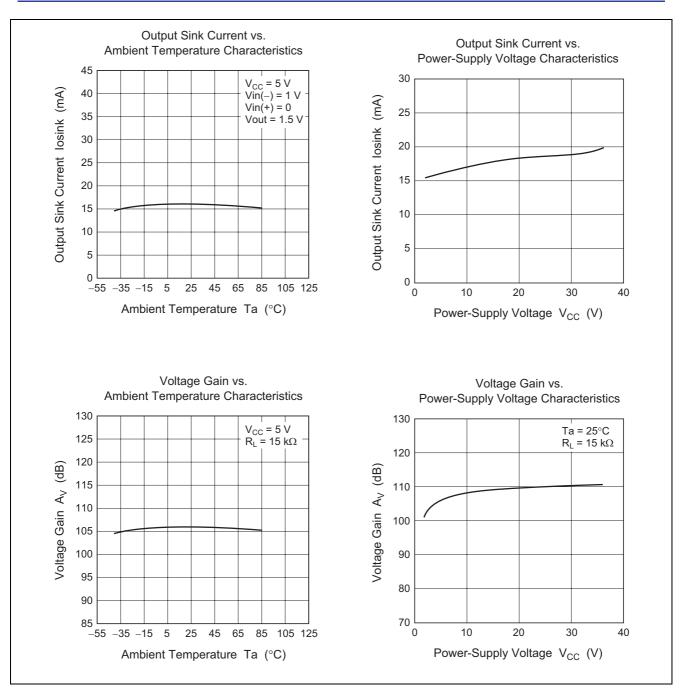


- $t_R$ :  $R_L = 5.1 \text{ k}\Omega$ , a 100 mV input step voltage that has a 5 mV overdrive
- With  $V_{IN}$  not applied, set the switch SW to the off position and adjust  $V_R$  so that  $V_O$  is in the vicinity of 1.4 V.
- Apply  $V_{IN}$  and turn the switch SW on.



# **Characteristic Curves**





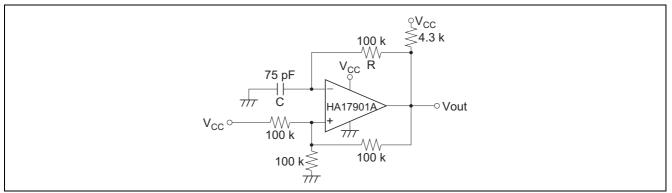


# HA17901A Application Examples

The HA17901A houses four independent comparators in a single package, and operates over a wide voltage range at low power from a single-voltage power supply. Since the common-mode input voltage range starts at the ground potential, the HA17901A is particularly suited for single-voltage power supply applications. This section presents several sample HA17901A applications.

### 1. Square-Wave Oscillator

The circuit shown in figure 1 has the same structure as a single-voltage power supply astable multivibrator. Figure 2 shows the waveforms generated by this circuit.





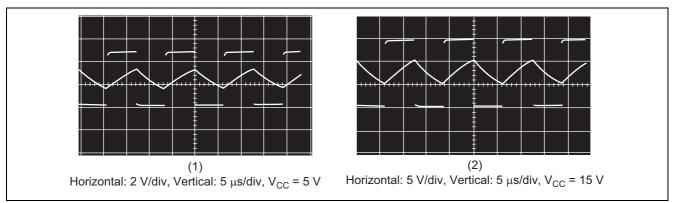


Figure 2 Operating Waveforms



#### 2. Pulse Generator

The charge and discharge circuits in the circuit from figure 1 are separated by diodes in this circuit. (See figure 3.) This allows the pulse width and the duty cycle to be set independently. Figure 4 shows the waveforms generated by this circuit.

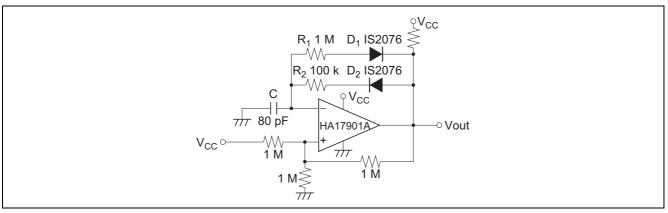


Figure 3 Pulse Generator

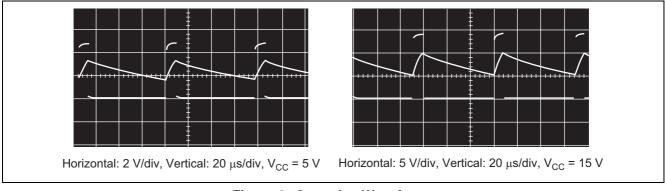


Figure 4 Operating Waveforms



### 3. Voltage Controlled Oscillator

In the circuit in figure 5, comparator  $A_1$  operates as an integrator,  $A_2$  operates as a comparator with hysteresis, and  $A_3$  operates as the switch that controls the oscillator frequency. If the output Vout1 is at the low level, the  $A_3$  output will go to the low level and the A1 inverting input will become a lower level than the A1 noninverting input. The  $A_1$  output will integrate this state and its output will increase towards the high level. When the output of the integrator  $A_1$  exceeds the level on the comparator  $A_2$  inverting input,  $A_2$  inverts to the high level and both the output Vout1 and the  $A_3$  output go to the high level. This causes the integrator to integrate a negative state, resulting in its output decreasing towards the low level. Then, when the  $A_1$  output level becomes lower than the level on the  $A_2$  noninverting input, the output Vout1 is once again inverted to the low level. This operation generates a square wave on Vout1 and a triangular wave on Vout2.

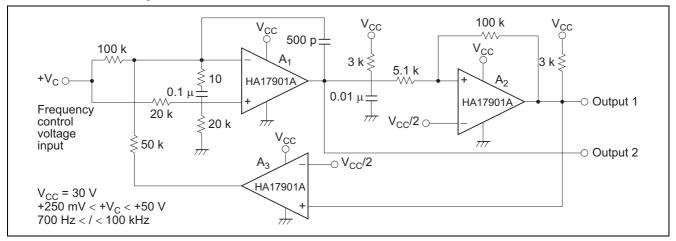


Figure 5 Voltage Controlled Oscillator

### 4. Basic Comparator

The circuit shown in figure 6 is a basic comparator. When the input voltage  $V_{IN}$  exceeds the reference voltage  $V_{REF}$ , the output goes to the high level.

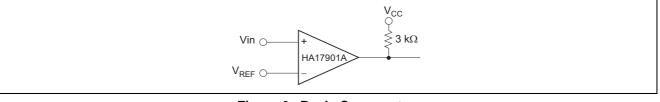


Figure 6 Basic Comparator



5. Noninverting Comparator (with Hysteresis)

Assuming  $+V_{IN}$  is 0 V, when  $V_{REF}$  is applied to the inverting input, the output will go to the low level (approximately 0 V). If the voltage applied to  $+V_{IN}$  is gradually increased, the output will go high when the value of the noninverting input,  $+V_{IN} \times R_2/(R_1 + R_2)$ , exceeds  $+V_{REF}$ . Next, if  $+V_{IN}$  is gradually lowered, Vout will be inverted to the low level once again when the value of the noninverting input,  $(Vout - V_{IN}) \times R_1/(R_1 + R_2)$ , becomes lower than  $V_{REF}$ . With the circuit constants shown in figure 7, assuming  $V_{CC} = 15$  V and  $+V_{REF} = 6$  V, the following formula can be derived, i.e.  $+V_{IN} \times 10$  M/(5.1 M + 10 M) > 6 V, and Vout will invert from low to high when  $+V_{IN}$  is > 9.06 V.

$$(Vout - V_{IN}) \times \frac{R_1}{R_1 + R_2} + V_{IN} < 6V$$
  
(Assuming Vout = 15V)

When  $+V_{IN}$  is lowered, the output will invert from high to low when  $+V_{IN} < 1.41$  V. Therefore this circuit has a hysteresis of 7.65 V. Figure 8 shows the input characteristics.

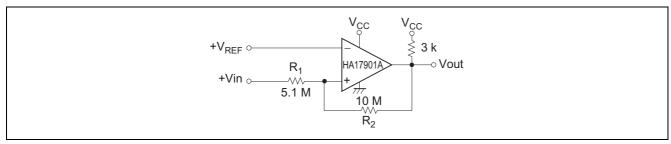


Figure 7 Noninverting Comparator

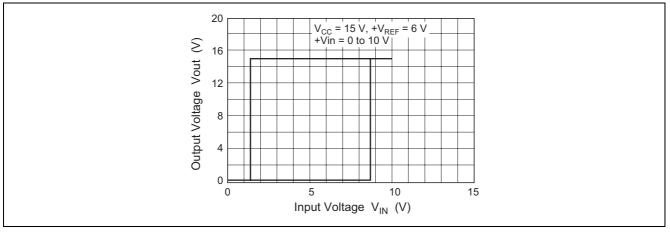


Figure 8 Noninverting Comparator I/O Transfer Characteristics



### 6. Inverting Comparator (with Hysteresis)

In this circuit, the output Vout inverts from high to low when  $+V_{IN} > (V_{CC} + Vout)/3$ . Similarly, the output Vout inverts from low to high when  $+V_{IN} < V_{CC}/3$ . With the circuit constants shown in figure 9, assuming  $V_{CC} = 15$  V and Vout = 15 V, this circuit will have a 5 V hysteresis. Figure 10 shows the I/O characteristics for the circuit in figure 9.

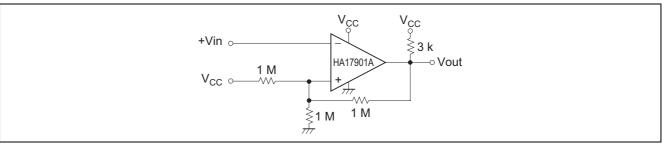


Figure 9 Inverting Comparator

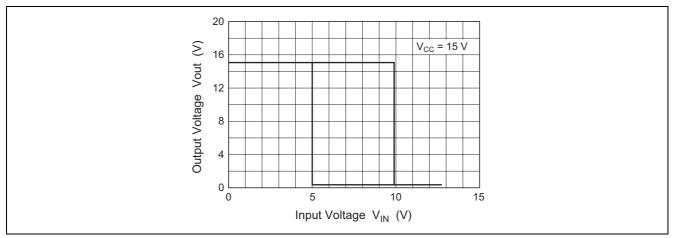


Figure 10 Inverting Comparator I/O Transfer Characteristics

7. Zero-Cross Detector (Single-Voltage Power Supply)

In this circuit, the noninverting input will essentially beheld at the potential determined by dividing  $V_{CC}$  with 100 k $\Omega$  and 10 k $\Omega$  resistors. When  $V_{IN}$  is 0 V or higher, the output will be low, and when  $V_{IN}$  is negative, Vout will invert to the high level. (See figure 11.)

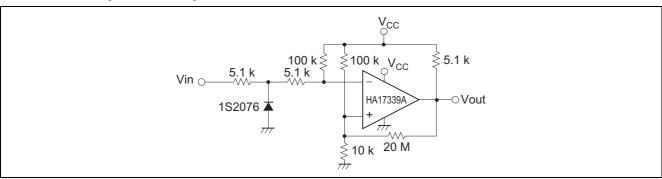
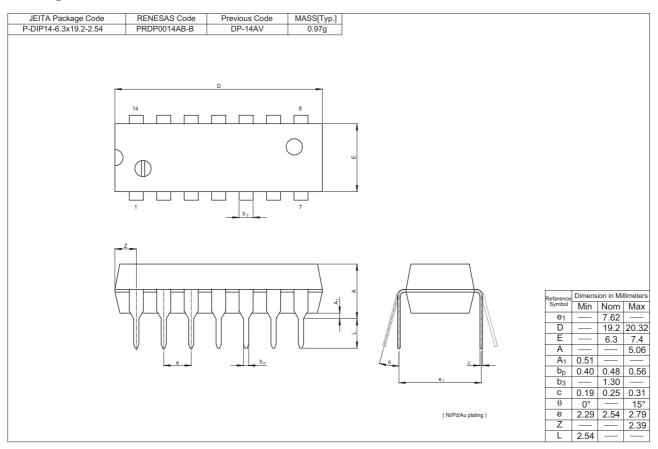
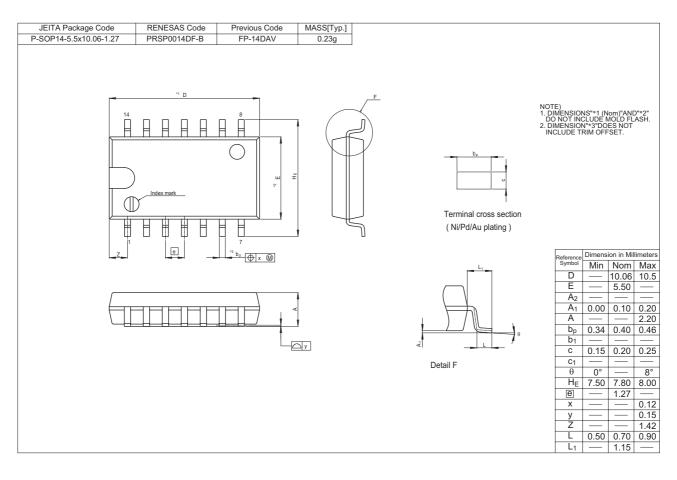


Figure 11 Zero-Cross Detector

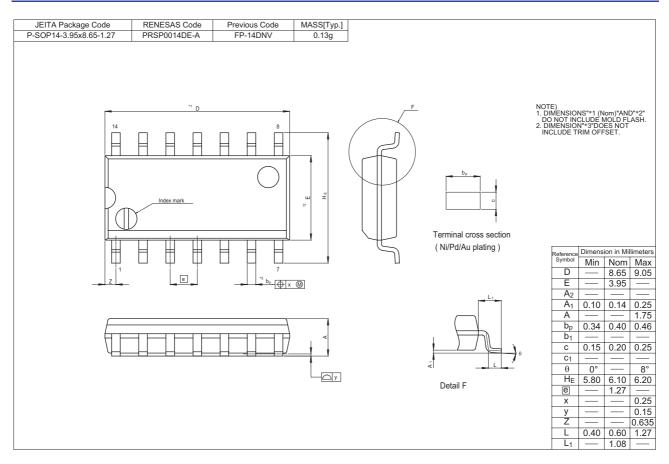
# **Package Dimensions**

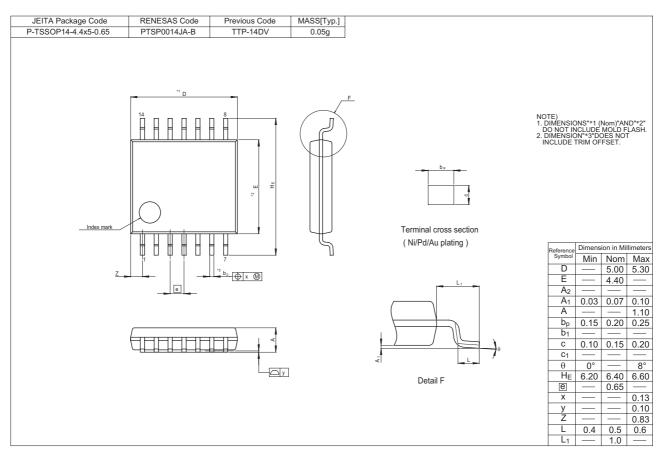




Rev.1.00 Mar 10, 2006 page 14 of 15









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