16-bit Cascadable ALU

FEATURES

- ☐ High-Speed (15ns), Low Power 16-bit Cascadable ALU
- ☐ Implements Add, Subtract, Accumulate, Two's Complement, Pass, and Logic Operations
- ☐ All Registers Have a Bypass Path for Complete Flexibility
- □ DECC SMD No. 5962-89959
- ☐ Available 100% Screened to MIL-STD-883, Class B
- ☐ Package Styles Available:
 - 68-pin Plastic LCC, J-Lead
 - 68-pin Ceramic LCC
 - 68-pin Ceramic PGA

DESCRIPTION

The L4C381 is a flexible, high speed, cascadable 16-bit Arithmetic and Logic Unit. It combines four 381-type 4-bit ALUs, a look-ahead carry generator, and miscellaneous interface logic - all in a single 68-pin package. While containing new features to support high speed pipelined architectures and single 16-bit bus configurations, the L4C381 retains full performance and functional compatibility with the bipolar '381 designs.

The L4C381 can be cascaded to perform 32-bit or greater operations. See "Cascading the L4C381" toward the end of this data sheet for more information.

ARCHITECTURE

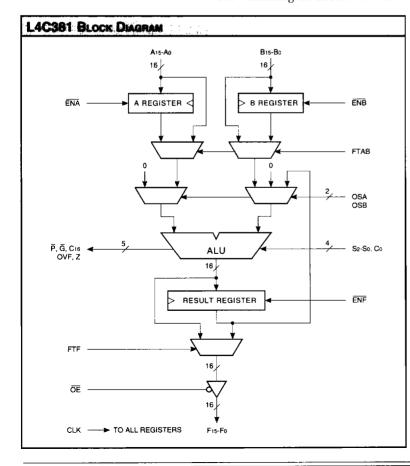
The L4C381 operates on two 16-bit operands (A and B) and produces a 16-bit result (F). Three select lines control the ALU and provide 3 arithmetic, 3 logical, and 2 initialization functions. Full ALU status is provided to support cascading to longer word lengths. Registers are provided on both the ALU inputs and the output, but these may be bypassed under user control. An internal feedback path allows the registered ALU output to be routed to one of the ALU inputs, accommodating chain operations and accumulation. Furthermore, the A or B input can be forced to Zero allowing unary functions on either operand.

ALU OPERATIONS

The S2-S0 lines specify the operation to be performed. The ALU functions and their select codes are shown in Table 1.

The two functions, B minus A and A minus B, can be achieved by setting the carry input of the least significant slice and selecting codes 001 and 010 respectively.

Time	L A. C. Europeano								
TABLE 1. ALU FUNCTIONS									
S2-S0	FUNCTION								
000	CLEAR (F = 00 ••• 00)								
001	NOT(A) + B								
010	A + NOT(B)								
011	A + B								
100	A XOR B								
101	A OR B								
110	A AND B								
111	PRESET (F = 11 ••• 11)								



Arithmetic Logic Units

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ALU STATUS

The ALU provides Overflow and Zero status bits. Carry, Propagate, and Generate outputs are also provided for cascading. These outputs are defined for the three arithmetic functions only. The ALU sets the Zero output when all 16 output bits are zero. The Generate, Propagate, C16, and OVF flags for the A + B operation are defined in Table 2. The status flags produced for NOT(A) + B and A + NOT(B) can be found by complementing Ai and Bi respectively in Table 2.

OPERAND REGISTERS

The L4C381 has two 16-bit wide input registers for operands A and B. These registers are rising edge triggered by a common clock. The A register is enabled for input by setting the \overline{ENA} control LOW, and the B register is enabled for input by setting the \overline{ENB} control LOW. When either the \overline{ENB} control or \overline{ENB} control is HIGH, the data in the corresponding input register will not change.

This architecture allows the L4C381 to accept arguments from a single 16-bit data bus. For those applications that do not require registered inputs, both the A and B operand registers can be bypassed with the FTAB control line. When the FTAB control is asserted (FTAB = HIGH), data is routed around the A and B input registers; however, they continue to function normally via the $\overline{\text{ENA}}$ and $\overline{\text{ENB}}$ controls. The contents of the input registers will again be available to the ALU if the FTAB control is released.

OUTPUT REGISTER

The output of the ALU drives the input of a 16-bit register. This rising-edge-triggered register is clocked by the same clock as the input registers. When the ENF control is LOW, data from the ALU will be clocked into the

TABLE 2. ALU STATUS FLAGS	
Bit Carry Generate = gi = AiBi Bit Carry Propagate = pi = Ai + Bi	for i = 0 15 for i = 0 15
Po = po Pi = pi (Pi-1) and	for i = 1 15
G0 = g0 Gi = gi + pi (Gi-1) Ci = Gi-1 + Pi-1 (C0)	for i = 1 15 for i = 1 15
then	
G = NOT(G15) P = NOT(P15) C16 = G15 + P15C0 OVF = C15 XOR C16	

output register. By disabling the output register, intermediate results can be held while loading new input operands. Three-state drivers controlled by the $\overline{\text{OE}}$ input allow the L4C381 to be configured in a single bidirectional bus system.

The output register can be bypassed by asserting the FTF control signal (FTF = HIGH). When the FTF control is asserted, output data is routed around the output register, however, it continues to function normally via the $\overline{\text{ENF}}$ control. The contents of the output register will again be available on the output pins if FTF is released. With both FTAB and FTF true (HIGH) the L4C381 is functionally identical to four cascaded 54S381-type devices.

OPERAND SELECTION

The two operand select lines, OSA and OSB, control multiplexers that precede the ALU inputs. These multiplexers provide an operand force-to-zero function as well as F register feedback to the B input. Table 3 shows the inputs to the ALU as a function of the operand select inputs. Either the A or B operands may be forced to zero.

TABL	E 3.	OPERAND S	ELECTION			
OSB	OSA	OPERAND B	OPERAND A			
0	0	F	Α			
0	1	0	Α			
1	0	В	0			
1	1	В	Α			

When both operand select lines are low, the L4C381 is configured as a chain calculation ALU. The registered ALU output is passed back to the B input to the ALU. This allows accumulation operations to be performed by providing new operands via the A input port. The accumulator can be preloaded from the A input by setting OSA true. By forcing the function select lines to the CLEAR state (000), the accumulator may be cleared. Note that this feedback operation is not affected by the state of the FTF control. That is, the F outputs of the L4C381 may be driven directly by the ALU. The output register continues to function, however, and provides the ALU B operand source.



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Storage temperature	65°C to +150°C
Operating ambient temperature	55°C to +125°C
Vcc supply voltage with respect to ground	0.5 V to +7.0 V
Input signal with respect to ground	
Signal applied to high impedance output	−3.0 V to +7.0 V
Output current into low outputs	25 mA
Latchup current	> 400 mA

OPERATING CONDITIONS To meet speci	ified electrical and switching character	istics	
Mode	Temperature Range (Ambient)	Supply Voltage	
Active Operation, Commercial	0°C to +70°C	4.75 V ≤ V CC ≤ 5.25 V	
Active Operation, Military	−55°C to +125°C	$4.50 \text{ V} \le \text{V}CC \le 5.50 \text{ V}$	

ELECTRI	CAL CHARACTERISTICS Ove	er Operating Conditions (Note 4)	_			
Symbol	Parameter	Test Condition	Min	Тур	Max	Unit
V OH	Output High Voltage	Vcc = Min., Iон = −2.0 mA	2.4			٧
V OL	Output Low Voltage	VCC = Min., IOL = 8.0 mA			0.5	V
V iH	Input High Voltage		2.0		Vcc	٧
V iL	Input Low Voltage	(Note 3)	0.0		0.8	V
lix	Input Current	Ground ≤ VIN ≤ VCC (Note 12)			±20	μA
loz	Output Leakage Current	Ground ≤ VOUT ≤ VCC (Note 12)			±20	μA
ICC1	Vcc Current, Dynamic	(Notes 5, 6)		15	30	mA
ICC2	VCC Current, Quiescent	(Note 7)			1.5	mA



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SWITCHING CHARACTERISTICS -- COMMERCIA: OPERATING RANGE (0°C to +70°C)

GUARANTEES MANNEY		TOTAL	DELAYS	Note	8 9, 10 (i	16)				94.		<u> </u>
To Output		L4C3	81-55			L4C3	81-40			L4C3	81-26	
From Input	F15-F0	P, G	OVF, Z	C16	F15-F0	P, G	OVF, Z	C16	F15-F0	P, G	OVF, Z	C16
FTAB = 0, FTF = 0							_					
Clock	32	38	53	36	26	30	44	32	22	22	26	22
C ₀	_	_	34	22	_		28	20	—	_	18	18
S2-S0, OSA, OSB	_	42	42	42	_	32	34	35	_	22	22	22
FTAB = 0, FTF = 1							_					
Clock	56	38	53	36	46	30	44	32	28	22	26	22
Co	37	_	34	22	30	_	28	20	22	-	18	18
S2-S0, OSA, OSB	55	42	42	42	40	32	34	35	26	22	22	22
FTAB = 1, FTF = 0												
A15-A0, B15-B0	—	36	46	37	_	30	40	32	_	22	22	22
Clock	32			_	26		_	_	22	_		_
C ₀	_	_	34	22	_	_	28	20		_	18	18
S2-S0, OSA, OSB	_	42	42	42	_	32	34	35		22	22	22
FTAB = 1, FTF = 1												
A15-A0, B15-B0	55	36	46	37	40	30	40	32	26	22	22	22
Clock (OSA, OSB = 0)	56	38	53	36	46	30	44	32	28	22	26	22
C ₀	37		34	22	30	_	28	20	22	_	18	18
S2-S0, OSA, OSB	55	42	42	42	40	32	34	35	26	22	22	22

GUARANTEED Minnesses	Serve AN	p Hou	o Tues	Wan	Respec	t to C	LOCK A	iswo E	DOE M	otes 9,	10 (ns)	
•		L4C3	81-55			L4C3	81-40			L4C3	81-26	
	FTAE	3 = 0	FTAE	3 = 1	FTAE	3 = 0	FTAE	3 = 1	FTAE	3 = 0	FTAE	3 = 1
Input	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold
A15-A0, B15-B0	8	2	35	2	8	2	28	2	8	2	16	2
Co	21	0	21	0	16	0	16	0	8	0	8	0
S2-S0, OSA, OSB	44	0	44	0	32	0	32	0	18	0	18	0
ENA, ENB, ENF	10	2	10	2	10	2	10	2	8	2	8	2

3-6

THI-STAT	E ENALE/Disco	e Times Notes	9, 10, 11 (ns)
	L4C381-55	L4C381-40	L4C381-26
tENA	20	18	16
tois	20	18	16

CLOCK CYCLE TIN	E AND PULSE	WIDTH Notes	9, 10 (ns)
	L4C381-55	L4C381-40	L4C381-26
Minimum Cycle Time	43	34	20
Highgoing Pulse	15	10	10
Lowgoing Pulse	15	10	10



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SWITCHING CHARACTERISTICS - COMMENDAL OPERATING RANGE (0°C to +70°C)

CHARLES BETWEEN		TIONAL	DELAYS	Note	s 9, 10 (i	(0)			
To Output		L4C3	81-20			L4C3	81-15		
From input	F15-F0	P, G	OVF, Z	C16	F15-F0	P, G	OVF, Z	C16	
FTAB = 0, FTF = 0									
Clock	11	20	20	20	11	15	15	15	
Co		_	14	14			13	13	
S2-S0, OSA, OSB	_	18	20	18		14	15	14	
FTAB = 0, FTF = 1									_
Clock	20	20	20	20	15	15	15	15	
Co	18	_	14	14	14	_	13	13	
S2-S0, OSA, OSB	20	18	20	18	15	14	15	14	
FTAB = 1, FTF = 0									
A15-A0, B15-B0		16	20	17	_	14	15	14	
Clock	11	_	_		11		_	_	
Co	_	_	14	14	—	_	13	13	
S2-S0, OSA, OSB		18	20	18	-	14	15	14	
FTAB = 1, FTF = 1								·	
A15-A0, B15-B0	20	16	20	17	15	14	15	14	
Clock (OSA, OSB = 0)	20	20	20	20	15	15	15	15	
Co	18		14	14	14	_	13	13	
S2-S0, OSA, OSB	20	18	20	18	15	14	15	14	

	EUP A	e Hou	e Pares	Wire!	RESPUE	r ru C	Leek P	en l		9, 10 (mg)
		L4C3	81-20		!	L4C3	81-15			
	FTAE	3 = 0	FTAE	3 = 1	FTAE	3 = 0	FTAE	3 = 1		
Input	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold		
A15-A0, B15-B0	5	0	14	0	5	0	12	0		
Co	12	0	12	0	10	0	10	0		
S2-S0, OSA, OSB	15	0	15	0	12	0	12	0	Ì	
ENA, ENB, ENF	5	0	5	0	5	0	5	0		

THE STATE	: Even#Date	a Tues Moles	9, 10, 11 (ns)
	L4C381-20	L4C381-15	
tENA	8	6	
tDIS	8	6	

CLOCK CYCLE THE AND PULSE WHEN NOISE 9, 10 (118)									
	L4C381-20	L4C381-15							
Minimum Cycle Time	18	14							
Highgoing Pulse	5	4							
Lowgoing Pulse	5	4							

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SWITCHING CHARACTERISTICS - MILITARY OPERATING RANGE (-55°C to +125°C)

To Output	L4C381-65					L4C3	81-45		L4C381-30			
From Input	F15-F0	P, G	OVF, Z	C16	F15-F0	Ē, Ğ	OVF, Z	C 16	F15-F0	P, G	OVF, Z	C16
FTAB = 0, FTF = 0						-						
Clock	37	44	63	45	28	34	50	34	26	28	34	28
Co	-		42	25	_		32	23	_	-	22	22
S2-S0, OSA, OSB	_	48	48	48		38	38	38		28	28	28
FTAB = 0, FTF = 1												_
Clock	68	44	63	45	56	34	50	34	34	28	34	28
Co	42		42	25	32		32	23	26		22	22
S2-S0, OSA, OSB	66	48	48	48	46	38	38	38	30	28	28	28
FTAB = 1, FTF = 0				_								
A15-A0, B15-B0	_	44	56	44		32	46	36		28	28	28
Clock	37				28	_	_	_	26		_	
Co	-		42	25			32	23	_		22	22
S2-S0, OSA, OSB	_	48	48	48		38	38	38	_	28	28	28
FTAB = 1, FTF = 1												
A15-A0, B15-B0	65	44	56	44	45	32	46	36	30	28	28	28
Clock (OSA, OSB = 0)	68	44	63	45	56	34	50	34	34	28	34	28
Co	42		42	25	32		32	23	26		22	22
S2-S0, OSA, OSB	66	48	48	48	46	38	38	38	30	28	28	28

GUARANTEED MINIMUM	GUARANTEED MINIMUM SETUP AND HOLD TIMES WITH RESPECT TO CLOCK RISING EDGE Notes 9, 10 (ns)											
		L4C3	81-65			L4C3	81-45		L4C381-30			
	FTAE	3 = 0	FTAE	3 = 1	FTAE	3 = 0	FTAE	3 = 1	FTAE	3 = 0	FTAE	3 = 1
Input	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold
A15-A0, B15-B0	10	3	43	3	8	3	33	3	8	3	20	3
Co	25	0	25	0	20	0	20	0	12	0	12	0
S2-S0, OSA, OSB	50	0	50	0	36	0	36	0	20	0	20	0
ENA, ENB, ENF	12	2	12	2	10	2	10	2	10	2	10	2

TRI-STATE ENABLE/DISABLE TIMES Notes 9, 10, 11 (ns)								
	L4C381-65	L4C381-45	L4C381-30					
t ENA	22	20	18					
tDIS	22	20	18					

CLOCK CYCLE TIME AND PULSE WIDTH Notes 9, 10 (ns)									
	L4C381-65	L4C381-45	L4C381-30						
Minimum Cycle Time	52	38	26						
Highgoing Pulse	20	15	12						
Lowgoing Pulse	20	15	12						



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SWITCHING CHARACTERISTICS - MILITARY OPERATING RANGE (-55°C to +125°C)

GUARANTEED MAXIMUM COMBINATIONAL DELAYS Notes 9, 10 (ns)									
To Output		L4C3	81-25			L4C3	81-20		
From Input	F15-F0	P, G	OVF, Z	C16	F15-F0	₽, Ğ	OVF, Z	C16	
FTAB = 0, FTF = 0			-						
Clock	14	24	24	24	14	20	20	20	
Co	—	_	18	18	 		16	16	
S2-S0, OSA, OSB		22	24	22		18	20	18	
FTAB = 0, FTF = 1									
Clock	25	24	24	24	20	20	20	20	
Co	21		18	18	17		16	16	
S2-S0, OSA, OSB	25	22	24	22	20	18	20	18	
FTAB = 1, FTF = 0							-		
A15-A0, B15-B0	_	20	25	22	_	17	20	17	
Clock	14	_			14	_			
Co	_		18	18		_	16	16	
S2-S0, OSA, OSB	_	22	24	22		18	20	18	
FTAB = 1, FTF = 1									
A15-A0, B15-B0	25	20	25	22	20	17	20	17	
Clock (OSA, OSB = 0)	25	24	24	24	20	20	20	20	
Co	21	_	18	18	17		16	16	
S2-S0, OSA, OSB	25	22	24	22	20	18	20	18	

		L4C3	81-25			L4C3	81-20			
	FTAE	3 = 0	FTAE	3 = 1	FTAE	B = 0	FTAE	B = 1		
input	Setup	Hold	Setup	Hold	Setup	Hold	Setup	Hold		
A15-A0, B15-B0	7	2	14	2	6	2	12	2	,	
Co	14	0	14	0	12	0	12	0		
S2-S0, OSA, OSB	19	0	19	0	16	0	16	0		
ENA, ENB, ENF	7	0	7	0	6	0	6	0		

TRI-STATE ENABLE/DISABLE TIMES Notes 9, 10, 11 (ns)								
	L4C381-25	L4C381-20						
tENA	14	10						
t DIS	14	10						

CLOCK CYCLE TIME AND PULSE WIDTH Notes 9, 10 (ns)								
	L4C381-25	L4C381-20						
Minimum Cycle Time	20	18						
Highgoing Pulse	8	6						
Lowgoing Pulse	8	6						

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NOTES

- 1. Maximum Ratings indicate stress specifications only. Functional operation of these products at values bevond those indicated in the Operating Conditions table is not implied. Exposure to maximum rating conditions for extended periods may affect reliability.
- 2. The products described by this specification include internal circuitry designed to protect the chip from damaging substrate injection currents and accumulations of static charge. Nevertheless, conventional precautions should be observed during storage, handling, and use of these circuits in order to avoid exposure to excessive electrical stress values.
- 3. This device provides hard clamping of transient undershoot and overshoot. Input levels below ground or above VCC will be clamped beginning at -0.6 V and VCC + 0.6 V. The device can withstand indefinite operation with inputs in the range of -0.5 V to +7.0 V. Device operation will not be adversely affected, however, input current levels will be well in excess of 100 mA.
- 4. Actual test conditions may vary from those designated but operation is guaranteed as specified.
- 5. Supply current for a given application can be accurately approximated by:

where

N = total number of device outputs

C = capacitive load per output V = supply voltage

F = clock frequency

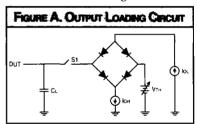
- 6. Tested with all outputs changing every cycle and no load, at a 5 MHz clock rate.
- 7. Tested with all inputs within 0.1 V of VCC or Ground, no load.
- 8. These parameters are guaranteed but not 100% tested.

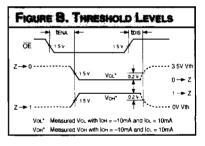
9. AC specifications are tested with 11. For the tENA test, the transition is input transition times less than 3 ns, output reference levels of 1.5 V (except tDIS test), and input levels of nominally 0 to 3.0 V. Output loading may be a resistive divider which provides for specified IOH and IOL at an output voltage of VOH min and VOL max respectively. Alternatively, a diode bridge with upper and lower current sources of IOH and IOL respectively, and a balancing voltage of 1.5 V may be used. Parasitic capacitance is 30 pF minimum, and may be distributed.

This device has high-speed outputs capable of large instantaneous current pulses and fast turn-on/turn-off times. As a result, care must be exercised in the testing of this device. The following measures are recommended:

- a. A 0.1 uF ceramic capacitor should be installed between VCC and Ground leads as close to the Device Under Test (DUT) as possible. Similar capacitors should be installed between device VCC and the tester common, and device ground and tester common.
- b. Ground and VCC supply planes must be brought directly to the DUT socket or contactor fingers.
- c. Input voltages should be adjusted to compensate for inductive ground and VCC noise to maintain required DUT input levels relative to the DUT ground pin.
- Each parameter is shown as a minimum or maximum value. Input requirements are specified from the point of view of the external system driving the chip. Setup time, for example, is specified as a minimum since the external system must supply at least that much time to meet the worst-case requirements of all parts. Responses from the internal circuitry are specified from the point of view of the device. Output delay, for example, is specified as a maximum since worst-case operation of any device always provides data within that time.

- measured to the 1.5 V crossing point with datasheet loads. For the tDIS test, the transition is measured to the ±200mV level from the measured steady-state output voltage with ±10mA loads. The balancing voltage, VTH, is set at 3.5 V for Z-to-0 and 0-to-Z tests, and set at 0 V for Zto-1 and 1-to-Z tests.
- 12. These parameters are only tested at the high temperature extreme, which is the worst case for leakage current.







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CASCADING THE L4C381

Cascading the L4C381 to 32 bits is accomplished simply by connecting the C16 output of the least significant slice to the C0 input of the most significant slice. The S2-S0, OSA, OSB, ENA, ENB, and ENF lines are common to both devices. The Zero output flags should be logically ANDed to produce the Zero flag for the 32-bit result. The OVF and C16 outputs of the most significant slice are valid for the 32-bit result.

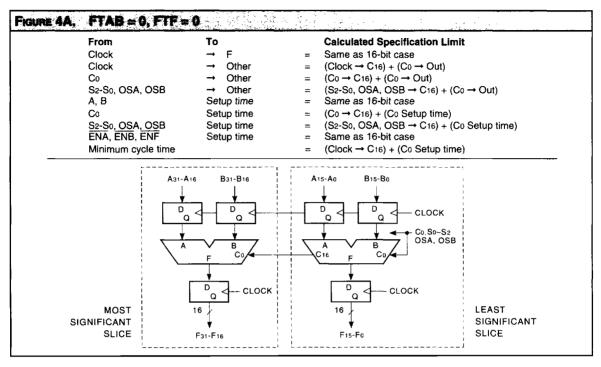
Propagation delay calculations for this configuration require two steps: First determine the propagation delay from the input of interest to the C16 output of the lower slice. Add this number to the delay from the C0 input of the upper slice to the output of interest

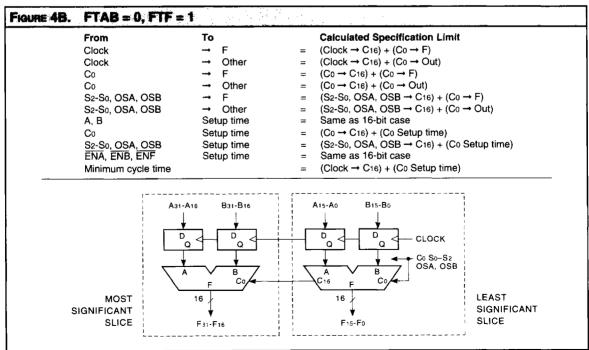
(of the C0 setup time, if the F register is used). The sum gives the overall input-to-output delay (or setup time) for the 32-bit configuration. This method gives a conservative result, since the C16 output is very lightly loaded. Formulas for calculation of all critical delays for a 32-bit system are shown in Figures 4A through 4D.

Cascading to greater than 32 bits can be accomplished in two ways: The simplest (but slowest) method is to simply connect the C16 output of each slice to the C0 input of the next more significant slice. Propagation delays are calculated as for the 32-bit case, except that the C0 to C16 delays for all intermediate slices must be added to the overall delay for each path. A

faster method is to use an external carry-lookahead generator. The P and G outputs of each slice are connected as inputs to the CLA generator, which in turn produces the C0 inputs for each slice except the least significant. The C16 outputs are not used in this case, except for the most significant one, which is the carry out of the overall system. The carry in to the system is connected to the C0 input of the least significant slice, and also to the carry lookahead generator. Propagation delays for this configuration are the sum of the time to \overline{P} , \overline{G} , for the least significant slice, the propagation delay of the carry lookahead generator, and the C0 to output time of the most significant slice.

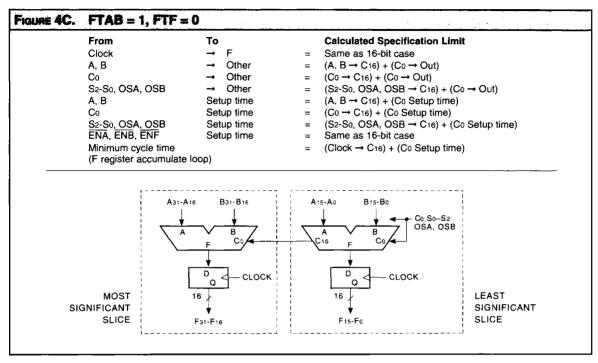
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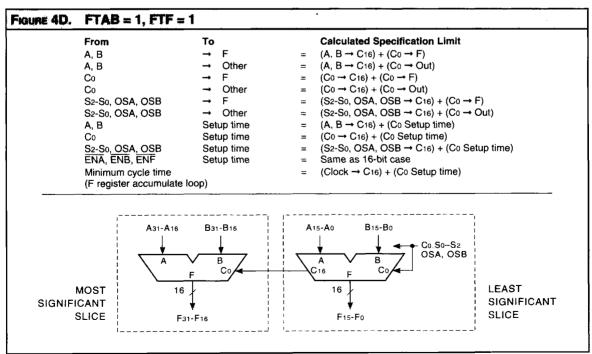






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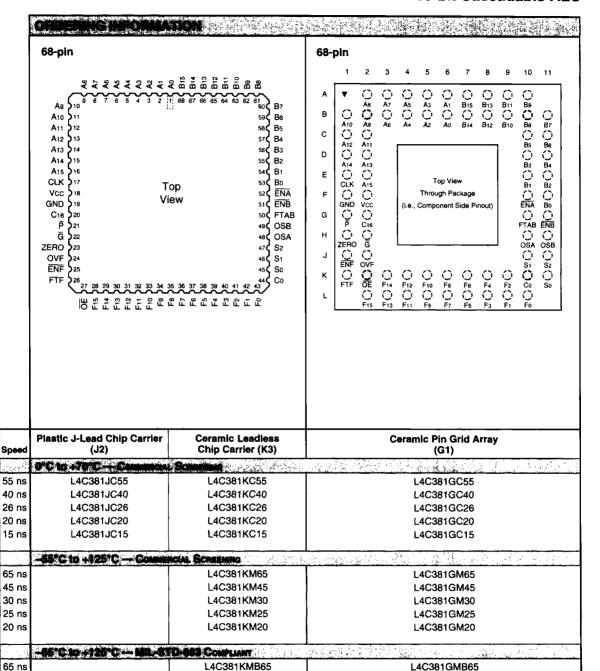
45 ns

30 ns

25 ns

20 ns

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L4C381GMB45

L4C381GMB30

L4C381GMB25

L4C381GMB20

L4C381KMB45

L4C381KMB30

L4C381KMB25

L4C381KMB20