

Utilizing the circuit designs perfected for recently introduced Quad Operational Amplifiers, these dual operational amplifiers feature 1) low power drain 2) a common mode input voltage range extending to ground/V_{EE}, 3) Single Supply or Split Supply operation and 4) pin outs compatible with the popular MC1558 dual operational amplifier. The PJ358 Series is equivalent to one half of an PJ324.

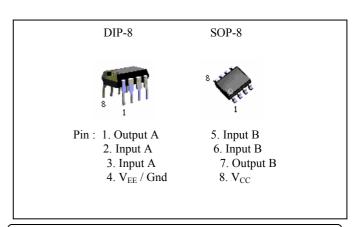
These amplifiers have several distinct advantages over standard operational amplifier types in single supply

FEATURES

- Short circuit protected outputs
- True differential input stage
- Single supply operation: 3.0 V to 32 Volts
- Low input bias currents
- Internally compensated
- Common mode range extends to negative supply
- Single and split supply operation
- Similar performance to the popular MC1558

applications. They can operate at supply voltages as low as 3.0 Volts or as high as 32 Volts with quiescent currents about one fifth of those associated with the LM741 (on a pet amplifier basis). The common mode input range includes the negative supply, thereby eliminating the necessity for external biasing components in many applications.

The output voltage range also includes the negative power supply voltage.



ORDERING INFORMATION

Device	Operating Temperature	Package
PJ358CS	20°C 4- 195°C	SOP-8
PJ358CD	-20°C to +85°C	DIP-8

MAXIMUM RATINGS ($T_A = +25^{\circ}C$, unless otherwise noted.)

Rating	Symbol	PJ358	Unit
Power Supply Voltage Single Supply	V_{CC}	32	Vdc
Split Supplies	V_{CC}, V_{EE}	±16	
Input Differential Voltage Range (1)	V_{IDR}	±32	Vdc
Input Common Mode Voltage Range (2)	V_{ICR}	-0.3 to 32	Vdc
Input forward current (3) (VI0.3V)	$I_{ m IF}$	50	mA
Output Short Circuit Duration	$t_{ m S}$	Continuous	
Junction Temperature Plastic Packages	T_{J}	150	$^{\circ}\mathbb{C}$
Storage Temperature Range Plastic Packages	$T_{\rm stg}$	-55 to +125	$^{\circ}\mathbb{C}$

NOTE: 1. Split Power Supplies.

2. For supply. Voltages less than 32V for the PJ358 the absolute maximum input voltage is equal to the supply voltage.

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3. This input curent will only exist when the voltage is negative at any of the input teads. Normal output states will retablish when the input voltage returns to a voltage greater than -0.3V.

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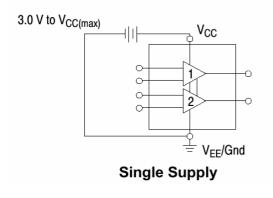
ELECTRICAL CHARACTERISTICS($T_A = 25$, $^{\circ}$ C Vcc = 5V unless otherwise noted).

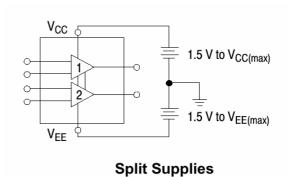
Characteristics	Symbol	Min	Тур	Max	Unit
Input Offset Voltage	$ m V_{IO}$				
$V_{CC} = 5.0 V$ to $30 V$					
V_{IC} =0V to Vcc -1.7 V, Vo =1.4V, R_S =0 Ω					
$T_A = 25^{\circ}C$			2.0	5.0	mV
$T_A = 85 ^{\circ}\text{C} \text{ to } -20 ^{\circ}\text{C}$				7.0	
Average Temperature Coefficient of Input Offset Voltage	$\triangle I_{IO}/\triangle T$		7.0		uV/°C
$T_A = 85 ^{\circ}\text{C} \text{ to } -20 ^{\circ}\text{C}$					
Input Offset Current	I_{IO}		5.0	50	nA
$T_A = 85^{\circ}C$ to $-20^{\circ}C$				150	
Average Temperature Coefficient of input Offset Current	$\triangle I_{IO}/\triangle T$		10		pA/°C
$T_A = 85^{\circ}C$ to $-20^{\circ}C$					
Input Bias Current	I_{IB}		45	-250	μΑ
$T_A = 85 \text{ t}^{\circ}\text{C o } -20^{\circ}\text{C}$			50	-500	
Input Common-Mode Voltage Range (Note1)	V_{ICR}				
$V_{CC} = 30 \text{ V}$		0		28.3	V
$V_{CC} = 30 \text{ V}, (T_A = 85^{\circ}\text{C to } -10^{\circ}\text{C})$		0		28	
Differental Input Voltage Range	V_{IDR}			V_{CC}	V
Large Signal Open-Loop Voltage Gain	A_{VOL}				V/mV
$R_L = 2.0K$, $V_{CC} = 15V$, For Large V_O Swing,		25	100		
$T_A = 85^{\circ}C$ to $-20^{\circ}C$		15			
Channel Separation			-120		dB
1.0 Khz to 20khz, Input Reterenced					
Common Mode Rejection Ratio	CMRR	65	70		dB
$R_S \le 10 \text{ k}\Omega$					
Power Supply Rejection Ratio	PSRR	65	100		dB
Output Voltage RangeRL = $2K\Omega$	V_{OR}	0		3.3	V
Output Voltage High Limit, (T _A =70 °C to 0°C)	V_{OH}				V
$V_{CC} = 30 \text{ V}, R_L = 2 \text{ k}\Omega$		26			
$V_{CC} = 30 \text{ V}, R_L = 10 \text{ k}\Omega$		27	28		
Output Voltage Low Limit , $(T_A = 70 ^{\circ}\text{C} \text{ to } 0 ^{\circ}\text{C})$	V_{OL}		5.0	20	mV
$V_{CC} = 5.0 \text{ V}, R_{L} = 10 \text{ k}\Omega$					
Output Source Current V _{ID} =+1.0V,V _{CC} =15V	I_{O^+}	20	40		mA
Output Sink Current	I _{O-}				mA
$V_{ID} = -1.0 \text{ V}, V_{CC} = 15 \text{ V}$	-	10	20		
$V_{ID} = -1.0 \text{ V}, V_{O} = 200 \text{ mV}$		12	50		μΑ
Output Short Circuit to Ground (Note 2)	I _{os}		40	60	mA
Power Supply Current, $(T_A = -20 ^{\circ}\text{C to } 85^{\circ}\text{C})$	I _{CC}				mA
$V_{CC} = 30 \text{ VV}_{O} = 0 \text{ V}, R_{L} = \infty$			1.5	3.0	
$V_{CC} = 5.0 \text{ V}, V_{O} = 0 \text{ V}, R_{L} = \infty$		-	0.7	1.2	

Notes:

- 1. The input common mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V. The upper end of the common mode voltage range is Vcc 17 V, but either or both inputs can go to +32 V.
- 2. Short circuits from the output to Vcc can cause excessive heating and eventual destruction. Destructive dissipation can result from simultaneous shorts on all amplifiers.

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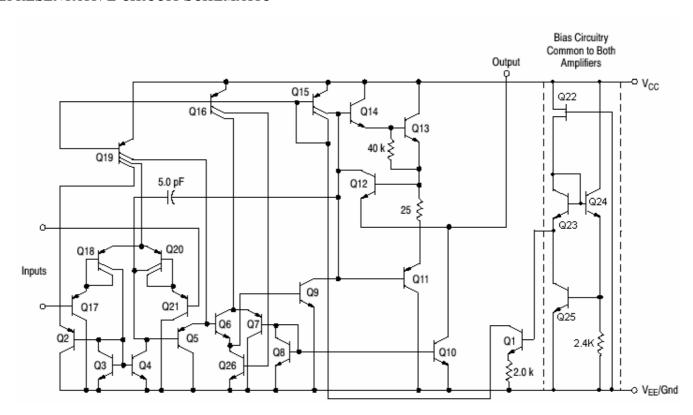




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REPRESENTATIVE CIRCUIT SCHEMATIC



CIRCUIT DESCRIPTION

The PJ358 series made using two internally compensated, two-stage operational amplifiers. The first stage of each consists of differential input devices Q20 and Q18 with input buffer transistors Q21 and Q17 and differential to single ended converter Q3 and Q4. The first stage performs not only the first stage gain function but also performs the level shifting and transconductance reduction functions. By reducing the transconductance, a smaller compensation capacitor (only 5.0pF) can be empllyed, thussaving chip area. The transconductance reduction is accomplished by splitting the collectors of Q20 and Q18. Another feature of this input stage is that the input common mode range can include the negative supply or ground, in single supply operation, without saturating either the input devices or the differential to single-ended converter. The second stage consists of a standard current source load amplifier stage.

Each amplifier is biased from an internal-voltage regulator which has a low temperature coefficient thus giving each amplifier good temperature characteristics as well as excellent power supply rejection.

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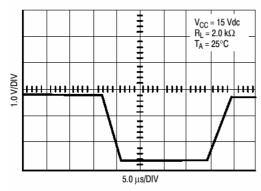
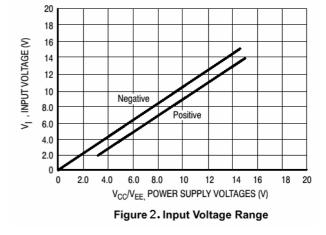


Figure 1. Large Signal Voltage Follower Response



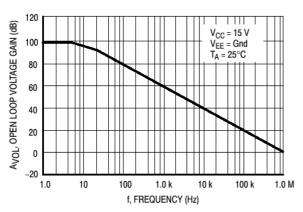


Figure 3. Large-Signal Open Loop Voltage Gain

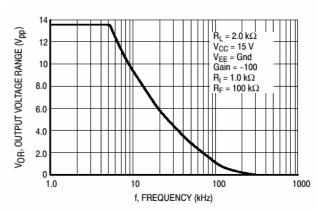


Figure 4. Large-Signal Frequency Response

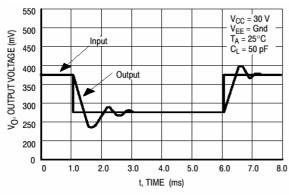


Figure 5. Small Signal Voltage Follower Pulse Response (Noninverting)

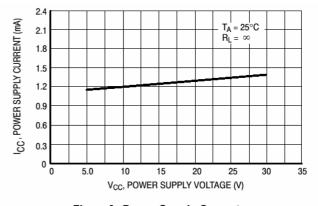


Figure 6. Power Supply Current versus Power Supply Voltage

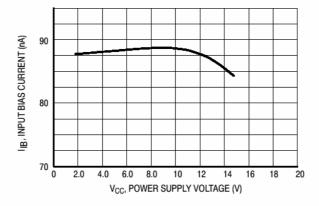


Figure 7. Input Bias Current versus Supply Voltage



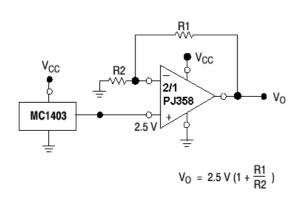


Figure 8 . Voltage Reference

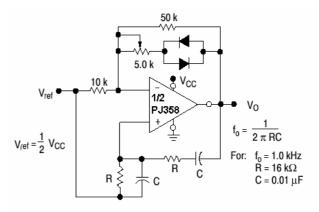


Figure 9. Wien Bridge Oscillator

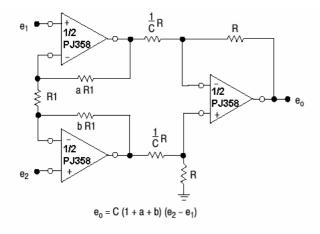


Figure 10. High Impedance Differential Amplifier

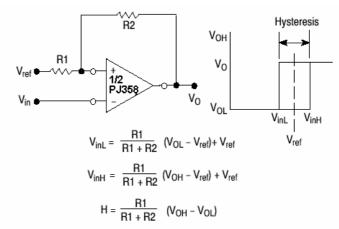


Figure 11. Comparator with Hysteresis

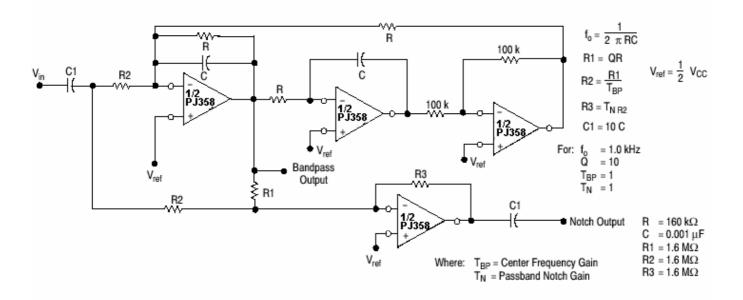


Figure 12. Bi-Quad Filter



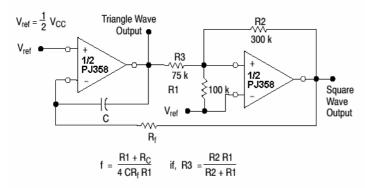
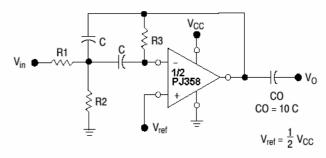


Figure 13. Function Generator



Given:
$$f_0$$
 = center frequency $A(f_0)$ = gain at center frequency

Choose value fo, C

Then: R3 =
$$\frac{Q}{\pi f_0 C}$$

R1 = $\frac{R3}{2 A(f_0)}$
R2 = $\frac{R1 R3}{4Q^2 R1 - R3}$

For less than 10% error from operational amplifier. $\frac{Q_0 f_0}{RW} < 0.1$

Where f_0 and BW are expressed in Hz.

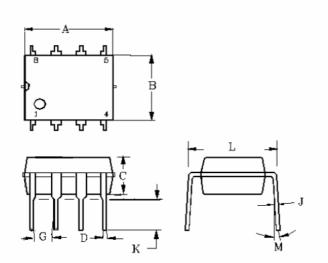
If source impedance varies, filter may be preceded with voltage follower buffer to stabilize filter parameters.

Figure 14. Multiple Feedback Bandpass Filter



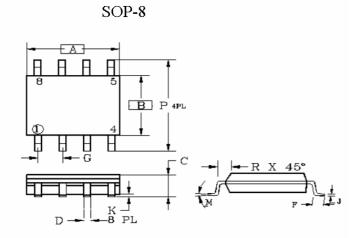
DIP-8 Mechanical drawing

DIP-8



	MILLIMETERS		INCHES	
DIM	MIN	MAX	MIN	MAX
A	9.07	9.32	0.357	0.367
В	6.22	6.48	0.245	0.255
С	3.18	4.43	0.125	0.135
D	0.35	0.55	0.019	0.020
G	2.54BSC		0.10BSC	
J	0.29	0.31	0.011	0.012
K	3.25	3.35	0.128	0.132
L	7.75	8.00	0.305	0.315
M	-	10°	-	10°

SOP-8 Mechanical drawing



	MILLIMETERS		INCHES		
DIM	MIN	MAX	MIN	MAX	
Α	4.80	5.00	0.189	0.196	
В	3.80	4.00	0.150	0.157	
С	1.35	1.75	0.054	0.068	
D	0.35	0.49	0.014	0.019	
F	0.40	1.25	0.016	0.049	
G	1.27BSC		0.05BSC		
K	0.10	0.25	0.004	0.009	
M	0°	7°	0°	7°	
P	5.80	6.20	0.229	0.244	
R	0.25	0.50	0.010 0.01		