RF2338

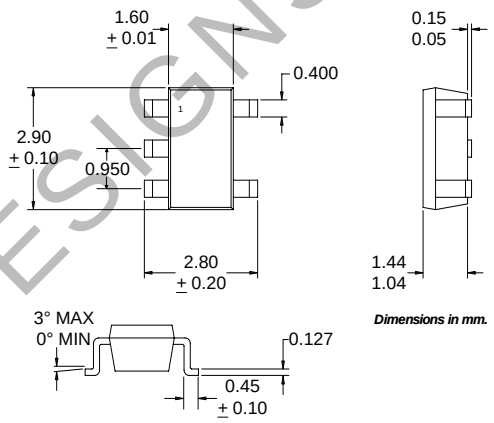
GENERAL PURPOSE AMPLIFIER

Typical Applications

- Broadband, Low Noise Gain Blocks
- IF or RF Buffer Amplifiers
- Driver Stage for Power Amplifiers
- Final PA for Low Power Applications
- Broadband Test Equipment

Product Description

The RF2338 is a general purpose, low-cost RF amplifier IC. The device is manufactured on an advanced Gallium Arsenide Heterojunction Bipolar Transistor (HBT) process, and has been designed for use as an easily-cascadable 50Ω gain block. Applications include IF and RF amplification in wireless voice and data communication products operating in frequency bands up to 6000MHz. The device is self-contained with 50Ω input and output impedances and requires only two external DC biasing elements to operate as specified. The RF2338 is available in a very small industry-standard SOT23-5 surface mount package, enabling compact designs which conserve board space.



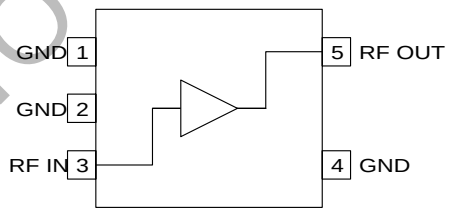
Package Style: SOT23-5

Optimum Technology Matching® Applied

- ☐ Si BJT
- ☑ GaAs HBT
- ☐ GaAs MESFET
- ☐ Si Bi-CMOS
- ☐ SiGe HBT
- ☐ Si CMOS
- ☐ InGaP/HBT
- ☐ GaN HEMT
- ☐ SiGe Bi-CMOS

Features

- DC to 6000MHz Operation
- Internally Matched Input and Output
- 12dB Small Signal Gain
- +24dBm Output IP3
- +11dBm Output Power
- Single Positive Power Supply



Functional Block Diagram

Ordering Information

RF2338	General Purpose Amplifier
RF2338 PCBA	Fully Assembled Evaluation Board

RF Micro Devices, Inc.
7628 Thorndike Road
Greensboro, NC 27409, USA
Tel (336) 664 1233
Fax (336) 664 0454
http://www.rfmd.com

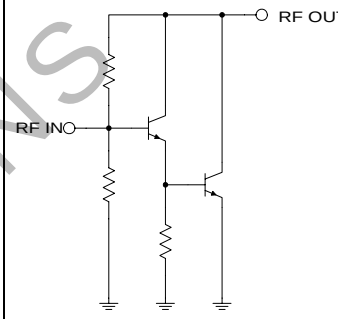
Absolute Maximum Ratings

Parameter	Rating	Unit
Input RF Power	+15	dBm
Operating Ambient Temperature	-40 to +85	°C
Storage Temperature	-60 to +150	°C

Caution! ESD sensitive device.

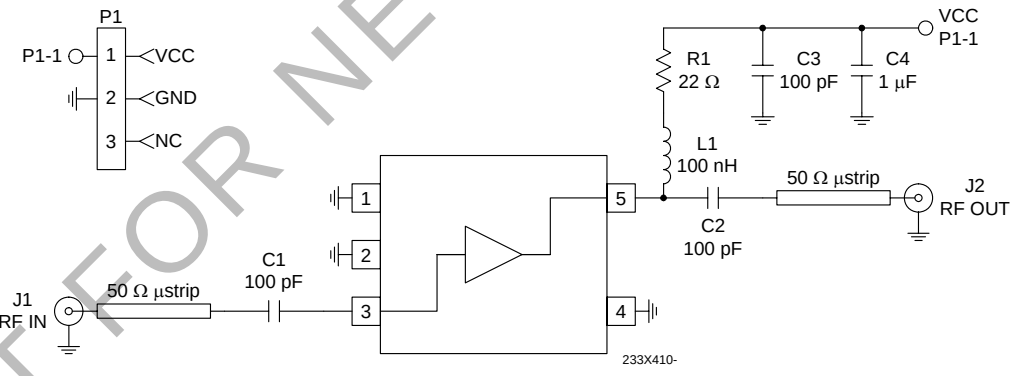
Parameter	Specification			Unit	Condition
	Min.	Typ.	Max.		
Overall					T=25°C, I_{CC}=40mA
Frequency Range		DC to 6000		MHz	
3dB Bandwidth		3.5		GHz	
Gain		12.1		dB	Freq=100MHz
		11.8		dB	Freq=1000MHz
		11.2		dB	Freq=2000MHz
		9.7		dB	Freq=3000MHz
		8.7		dB	Freq=4000MHz
		8		dB	Freq=5000MHz
		7.3		dB	Freq=6000MHz
Gain Flatness		±0.5		dB	100MHz to 2000MHz
Noise Figure		5.3		dB	Freq=2000MHz
Input VSWR		2.0:1			In a 50Ω system, DC to 3000MHz
Output VSWR		2.0:1			In a 50Ω system, DC to 3000MHz
Output IP_3		+23.5		dBm	Freq=2000MHz±50kHz, P_{TONE}=-10dBm
Output P_{1dB}		+10.5		dBm	Freq=2000MHz
Reverse Isolation		15.6		dB	Freq=2000MHz
Thermal					I_{CC}=40mA, P_{DISS}=141mW (See Note.)
$Theta_{JC}$		322		°C/W	
Maximum Measured Junction Temperature		130		°C	T_{AMB}=+85°C, V_{PIN}=3.53V
Mean Time Between Failures		20,000		years	See Note.
Power Supply					With 22Ω bias resistor
Device Operating Voltage		3.7		V	At pin 5 with I_{CC}=40mA
Supply Voltage		4.6		V	At evaluation board connector, I_{CC}=40mA
Operating Current		40	43	mA	See note.

Note: Because of process variations from part to part, the current resulting from a fixed bias voltage will vary. As a result, caution should be used in designing fixed voltage bias circuits to ensure the worst case bias current does not exceed 43mA over all intended operating conditions.

Pin	Function	Description	Interface Schematic
1	GND	Ground connection. For best performance, keep traces physically short and connect immediately to ground plane.	
2	GND	Same as pin 1.	
3	RF IN	RF input pin. This pin is NOT internally DC-blocked. A DC-blocking capacitor, suitable for the frequency of operation, should be used in most applications. DC coupling of the input is not allowed, because this will override the internal feedback loop and cause temperature instability.	
4	GND	Same as pin 1.	
5	RF OUT	RF output and bias pin. Biasing is accomplished with an external series resistor and choke inductor to V_{CC}. The resistor is selected to set the DC current into this pin to a desired level. The resistor value is determined by the following equation: $$R = \frac{(V_{SUPPLY} - V_{DEVICE})}{I_{CC}}$$ Care should also be taken in the resistor selection to **ensure that the current into the part never exceeds 43mA over the planned operating temperature**. This means that a resistor between the supply and this pin is always required, even if a supply near 3.6V is available, to provide DC feedback to prevent thermal runaway. Because DC is present on this pin, a DC blocking capacitor, suitable for the frequency of operation, should be used in most applications. The supply side of the bias network should also be well bypassed.	

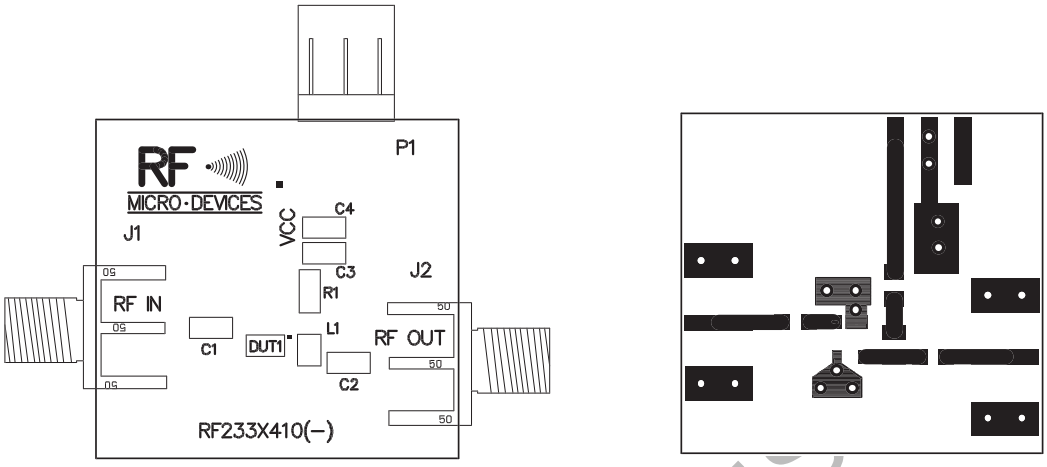
Evaluation Board Schematic

(Download Bill of Materials from www.rfmd.com.)

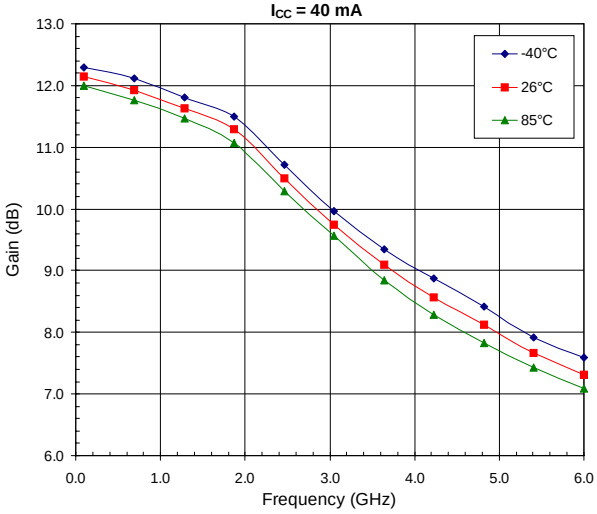


Evaluation Board Layout
Board Size 1.0” x 1.0”

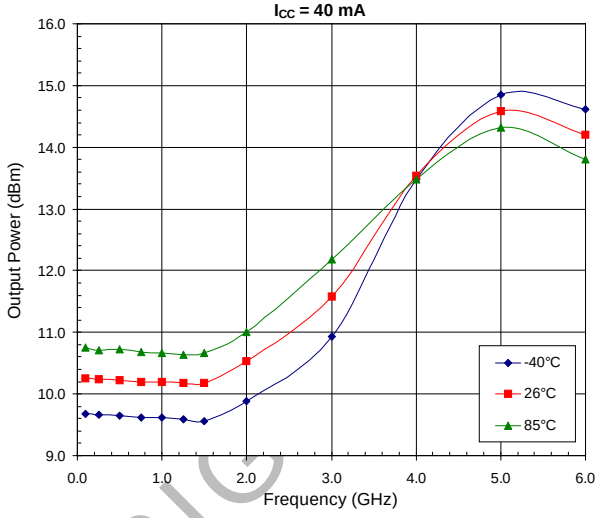
Board Thickness 0.020”, Board Material R0-4003 Rogers



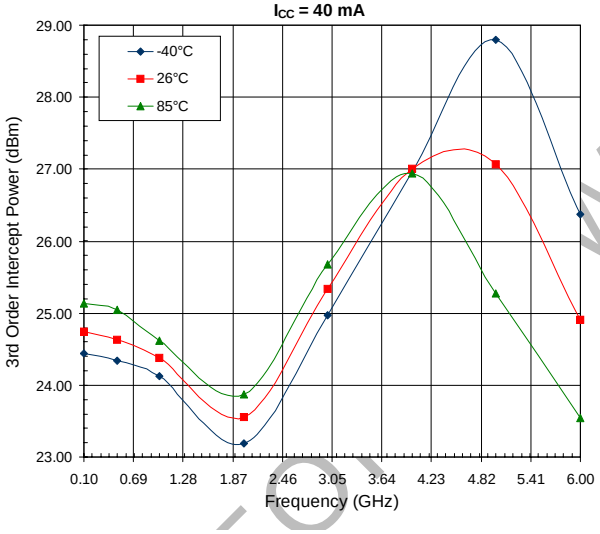
Gain versus Frequency Across Temperature
I_{CC} = 40 mA



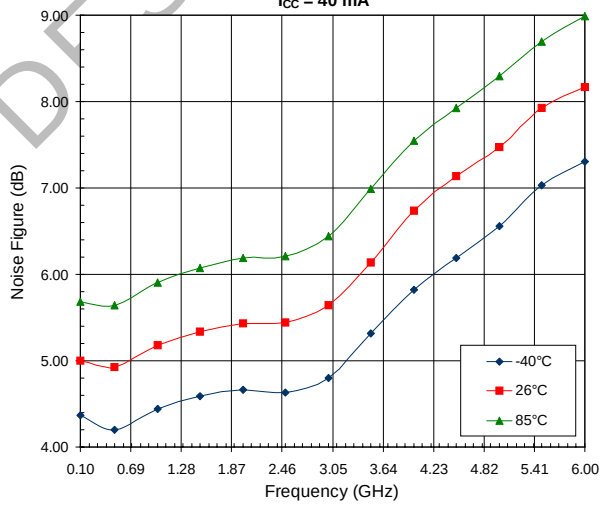
Output P1dB versus Frequency Across Temperature
I_{CC} = 40 mA



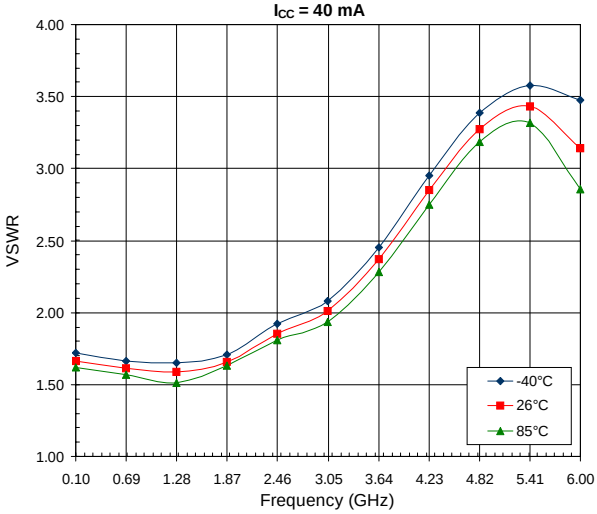
Output IP3 versus Frequency Across Temperature
I_{CC} = 40 mA



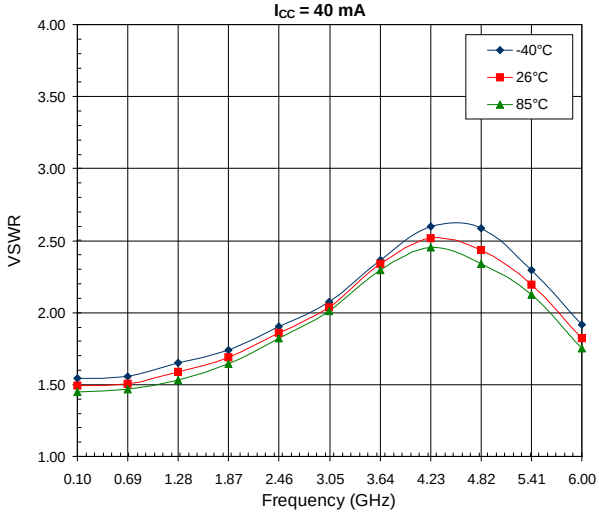
Noise Figure versus Frequency Across Temperature
I_{CC} = 40 mA



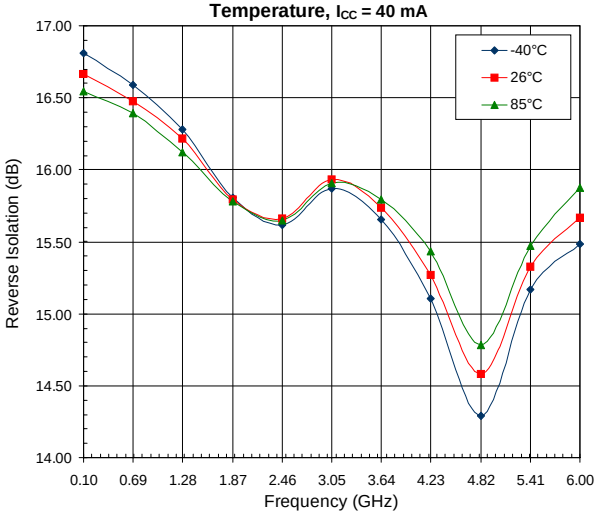
Input VSWR versus Frequency Across Temperature
I_{CC} = 40 mA



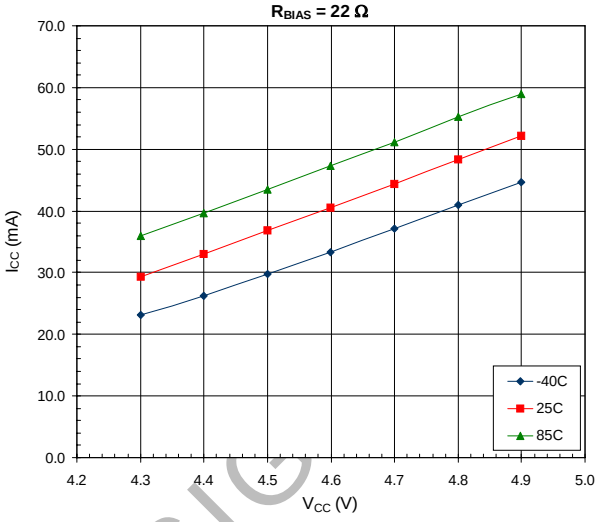
Output VSWR versus Frequency Across Temperature
I_{CC} = 40 mA



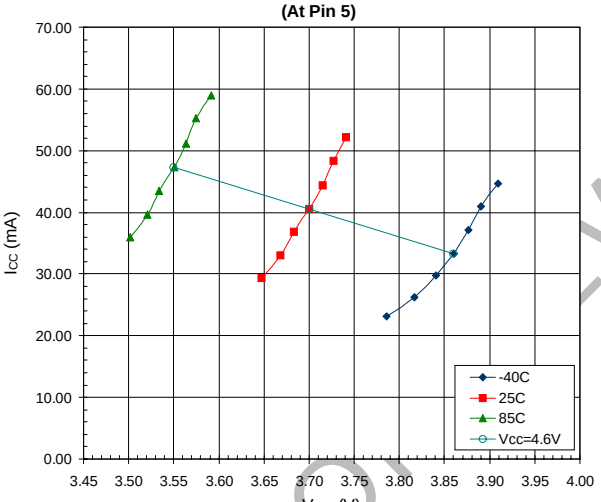
Reverse Isolation versus Frequency Across Temperature, I_{CC} = 40 mA



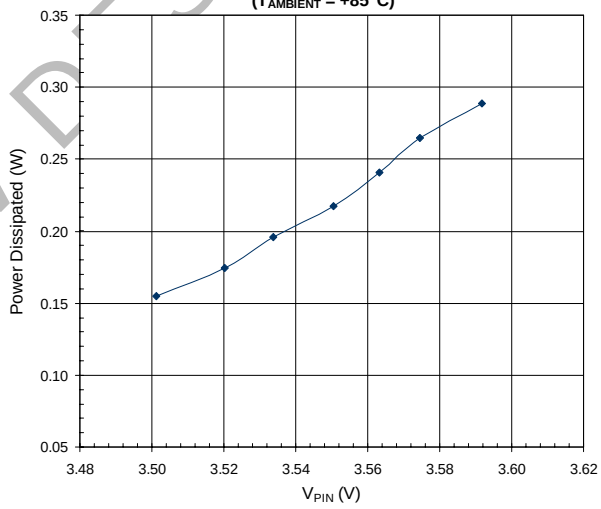
Current Voltage at Evaluation Board Connector, R_{BIAS} = 22 Ω



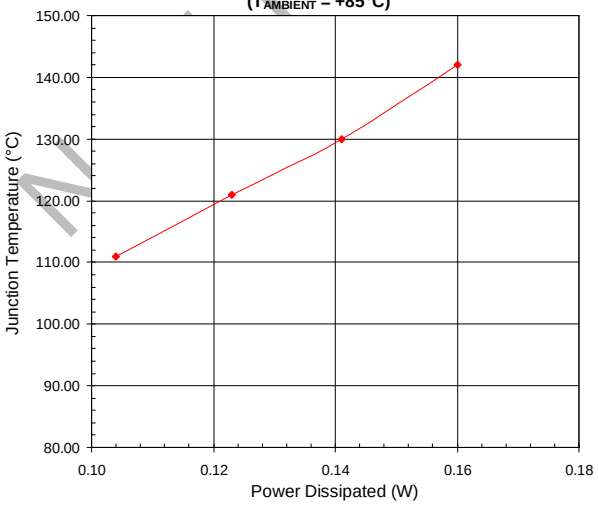
Current versus Voltage (At Pin 5)



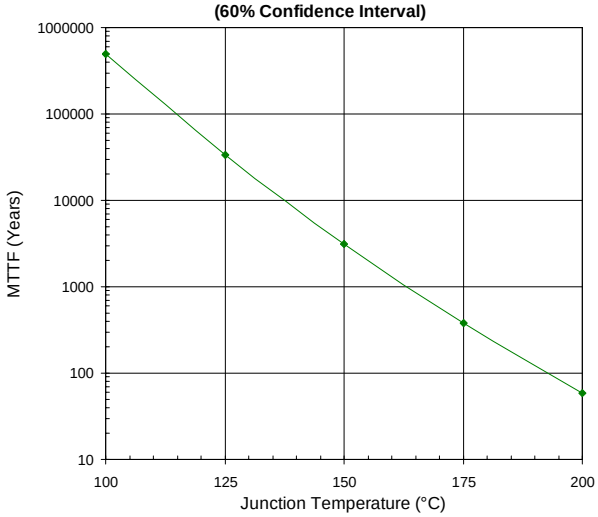
Power Dissipated versus Voltage at Pin 5 ($T_{AMBIENT}$ = +85°C)



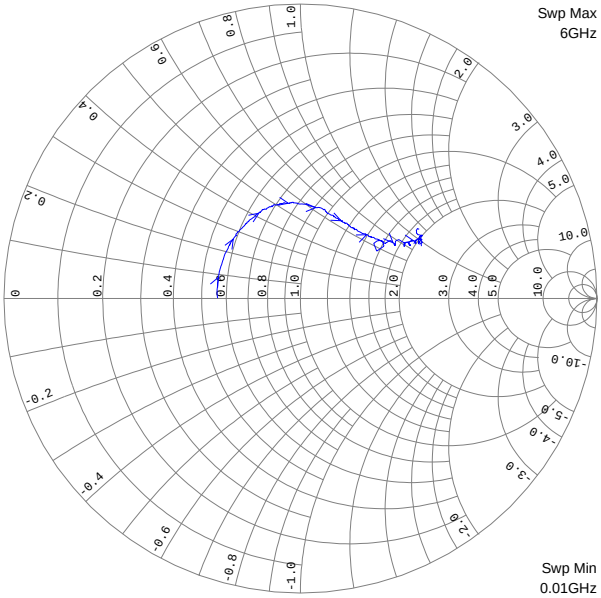
Junction Temperature versus Power Dissipated ($T_{AMBIENT}$ = +85°C)



MTTF versus Junction Temperature (60% Confidence Interval)



De-Embedded S11, V_{CC} = 3.67V, I_{CC} = 40mA, T = 25°C



De-Embedded S22, V_{CC} = 3.67V, I_{CC} = 40mA, T = 25°C