

# FLAT PACK CRYSTAL CLOCK OSCILLATORS -5.2 to -4.5Vdc & 1.8 to 15Vdc - 0.12Hz to 200MHz

# Description

Q-Tech's flat pack crystal oscillators consist of a source clock square wave generator, logic output buffers and/or logic divider stages, and a round AT highprecision quartz crystal built in an all metal flat package.

# Features

- Made in the USA
- ECCN: EAR99
- DFARS 252-225-7014 Compliant: Electronic Component Exemption
- Wide frequency range from 0.12Hz to 200MHz
- Available as QPL MIL-PRF-55310/21 (TTL) QT24 only
- Choice of flat packs and pin outs
- Choice of supply voltages
- Choice of output logic options
- AT-Cut crystal
- All metal hermetically sealed package
- Tight or custom symmetry available
- Capacitive load drive capability (Z output)
- Low height
- External tuning capacitor option
- Fundamental and third overtone designs
- Tristate function option D
- Three-point crystal mounts
- Custom design available tailors to meet customer's needs
- Q-Tech does not use pure lead or pure tin in its products

# Applications

- Designed to meet today's requirements for all voltage applications
- Wide military clock applications
- Industrial controls
- Microcontroller driver



# **Ordering Information**



(\*) Hot Solder Dip Sn60 per MIL-PRF 55310 is optional for an additional cost (\*\*) Please specify supply voltage when ordering CMOS

For frequency stability vs. temperature options not listed herein, request a custom part number.

For Non-Standard requirements, contact Q-Tech Corporation at Sales@Q-Tech.com

# **Packaging Options**

Standard packaging in a locked anti-static cardboard

# Other Options Available For An Additional Charge

- Lead forming available on all packages. Please contact for details.
- P. I. N. D. test (MIL-STD 883, Method 2020)
- Lead trimming

All Flat Pack packages are available in surface mount form. Specifications subject to change without prior notice.



# **Electrical Characteristics**

Parameters	C	AC	HC	Т	L (*)	ECL / PECL (**)	
Output freq. range QT21, 24, 25	500Hz — 15MHz	500Hz — 125MHz	0.12Hz — 125MHz	0.12Hz — 125MHz	0.12Hz — 160MHz	1MHz — 200MHz	
(Fo) QT22, 26, 28	500Hz — 15MHz	500Hz – 85MHz	500Hz — 85MHz	500Hz — 85MHz	500Hz — 85MHz	8MHz — 85MHz	
Supply voltage (Vdd)	$5V \sim 15Vdc \pm 10\%$	$5.0$ Vdc $\pm 10\%$		$3.3$ Vdc $\pm 10\%$	-5.2Vdc ± 5% (10K / 10KHECL) 5Vdc ± 5% (PECL) 3.3Vdc ± 5% (LVPECL)		
Maximum Applied Voltage (Vdd max.)	-0.5 to +18Vdc	-0.5 to +7.0Vdc			-0.5 to +5.0Vdc	0 to -8.0Vdc (10K / 10KHECL) 0 to +8.0Vdc (PECL) 0 to +5.0Vdc (LVPECL)	
Freq. stability ( $\Delta F/\Delta T$ )		See Option codes					
Operating temp. (Topr)		See Option codes					
Storage temp. (Tsto)		-62°C to + 125°C					
Operating supply current (Idd) (No Load)	F and Vdd dependent 3 mA max. at 5V up to 5MHz 25 mA max. at 15V up to 15MHz	$\begin{array}{cccccccc} 20 \text{ mA max.} & - 0.12 \text{Hz} \sim < 16 \text{MHz} & 3 \text{ m} \\ 25 \text{ mA max.} & - 16 \text{MHz} \sim < 40 \text{MHz} & 6 \text{ m} \\ 35 \text{ mA max.} & - 40 \text{MHz} \sim < 60 \text{MHz} & 10 \text{ m} \\ 45 \text{ mA max.} & - 60 \text{MHz} \sim < 85 \text{MHz} & 20 \text{ m} \\ 60 \text{ mA max.} & - 85 \text{MHz} \sim & 125 \text{MHz} & 30 \text{ m} \\ & & & & & & & & & & & & & & & & & & $			$\begin{array}{llllllllllllllllllllllllllllllllllll$	45 mA max 8MHz ~ < 125MHz 75 mA max 125MHz ~ 200MHz	
Symmetry (50% of ouput waveform or 1.4Vdc for TTL)	45/55% max. Fo < 4MHz 40/60% max. Fo ≥ 4MHz	45/55% max. Fo < 12MHz 40/60% max. Fo ≥ 12MHz				45/55% max. Fo < 12MHz 40/60% max. Fo ≥ 12MHz	
Rise and Fall times (with typical load)	30ns max. (Measured from 10% to 90%)	15ns max. Fo < 15kHz 6ns max. Fo 15kHz ~ 39.999MHz 3ns max. Fo 40MHz ~ 160 MHz (Measured from 10% to 90% CMOS or from 0.8V to 2.0V TT			9 2.0V TTL)	3.5ns max. Fo < 125MHz 3ns max. Fo 125MHz ~ 200MHz (Measured from 20% to 80%)	
Output Load		15pF // 10kΩ		<b>10TTL Fo &lt; 20MHz</b> 6TTL Fo ≥ 20MHz	15pF // 10kΩ	50Ω to -2V (10K / 10KH) 50Ω to Vcc -2V (P & LP)	
Start-up time (Tstup)	10ms max.						
Output voltage (Voh/Vol)	0.9 x Vdd min.; 0.1 x Vdd max.			2.4V min.; 0.4V max.	0.9 x Vdd min.; 0.1 x Vdd max.	-1.15V min; -1.54V max. (E) 4V min.; 3.37V max. (PE) 2.27V min.; 1.68V max. (LP)	
Output Current (Ioh/Iol)	$\pm$ 1mA typ. at 5V $\pm$ 6.8mA typ. at 15V	$\pm 24mA$	±8 mA	-1.6mA / TTL +40µA / TTL	$\pm 4 \mathrm{mA}$ .	-50mA	
Enable/Disable Tristate function	Call for details	$\label{eq:VIH} \begin{array}{l} VIH \geq 2.2V \mbox{ Oscillation;} \\ VIL \leq 0.8V \mbox{ High Impedance} \end{array}$			$\label{eq:VIH} \begin{array}{l} VIH \geq 0.7 \ x \ Vdd \ Oscillation; \\ VIL \leq 0.3 \ x \ Vdd \ High \ Impedance \end{array}$	Call for details	
Jitter RMS 1 <sub>o</sub> (at 25°C)	8ps typ < 40MHz 5ps typ ≥ 40MHz				15ps typ < 40MHz 8ps typ ≥ 40MHz	Integrated phase jitter 12kHz - 20MHz 1ps typ.	
Aging (at 70°C)	$\pm$ 5ppm max. first year / $\pm$ 2ppm typ. per year thereafter						

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(\*) Available in 2.5Vdc (N) or 1.8Vdc (R)

(\*\*) Please contact Q-Tech for details on 100KECL logic (EF)

Z Output logic can drive up to 200 pF load with typical 6ns rise & fall times (tr, tf) -

Q-TECH Corporation

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# **Package Configuration Versus Pin Connections**









.060/.035

(1.52/889)

В



.010

(254)









D





Е

















Equivalent E/D Output QT # Conf Vcc GND Case MIL-PRF-55310 (\*) N/C Configuration QT21 10 N/A А 13 10 11 12 QT22 В 8 9 9 10 11 N/A С 10 /21 = QT24TQT24 13 10 11 12 QT25 N/A D 13 10 10 11 12 QT26 Е 14 7 7 8 6 N/A QT28 F 8 9 9 10 11 N/A

(\*) ECL / PECL complimentary output available on pin 12 (except QT22, 26, & 28) with a Q-Tech custom part number

# **Package Information**

- Package material (Header and Leads): Kovar
- Lead finish: Gold Plated  $50\mu \sim 80\mu$  inches
- $\label{eq:Nickel Underplate} Nickel Underplate 100\mu \sim 250\mu \ inches$   $\bullet \ Cover: Kovar, Gold Plated - 50\mu \sim 100\mu \ inches$
- Nickel Underplate  $-70\mu \sim 90\mu$  inches • Package to lid attachment: Seam weld
- Weight: 2.0g typ., 4.0g max.



#### FLAT PACK **CRYSTAL CLOCK OSCILLATORS** -5.2 to -4.5Vdc & 1.8 to 15Vdc - 0.12Hz to 200MHz

# **Test Circuit**

**Output Waveform (Typical)** 





# **Startup Time**





# **Supply Current**



#### TYPICAL TEST CIRCUIT FOR CMOS LOGIC



(\*) CL INCLUDES PROBE AND JIG CAPACITANCE

The Tristate function on pin 1 has a built-in pull-up resistor typical 50k $\Omega$ , so it can be left floating or tied to Vdd without deteriorating the electrical performance.

# **Frequency vs. Temperature Curve**





#### **Thermal Characteristics**

The heat transfer model in a hybrid package is described in figure 1.

Heat spreading occurs when heat flows into a material layer of increased cross-sectional area. It is adequate to assume that spreading occurs at a 45° angle.

The total thermal resistance is calculated by summing the thermal resistances of each material in the thermal path between the device and hybrid case.

$$RT = R1 + R2 + R3 + R4 + R5$$

The total thermal resistance RT (see figure 2) between the heat source (die) to the hybrid case is the Theta Junction to Case (Theta JC) in°C/W.

- Theta junction to case (Theta JC) for this product is 30°C/W.
- Theta case to ambient (Theta CA) for this part is 100°C/W.
- Theta Junction to ambient (Theta JA) is 130°C/W.

Maximum power dissipation PD for this package at 25°C is:

- PD(max) = (TJ (max) TA)/Theta JA
- With  $TJ = 175^{\circ}C$  (Maximum junction temperature of die)
- PD(max) = (175 25)/130 = 1.15W





# **Environmental Specifications**

Q-Tech Standard Screening/QCI (MIL-PRF55310) is available for all of our Flat Packs. Q-Tech can also customize screening and test procedures to meet your specific requirements. The Flat Packs are designed and processed to exceed the following test conditions:

Environmental Test	Test Conditions			
Temperature cycling	MIL-STD-883, Method 1010, Cond. B			
Constant acceleration	MIL-STD-883, Method 2001, Cond. A, Y1			
Seal: Fine and Gross Leak	MIL-STD-883, Method 1014, Cond. A and C			
Burn-in	160 hours, 125°C with load			
Aging	30 days, 70°C, ± 1.5ppm max			
Vibration sinusoidal	MIL-STD-202, Method 204, Cond. D			
Shock, non operating	MIL-STD-202, Method 213, Cond. I			
Thermal shock, non operating	MIL-STD-202, Method 107, Cond. B			
Ambient pressure, non operating	MIL-STD-202, 105, Cond. C, 5 minutes dwell time minimum			
Resistance to solder heat	MIL-STD-202, Method 210, Cond. C			
Moisture resistance	MIL-STD-202, Method 106			
Terminal strength	MIL-STD-202, Method 211, Cond. C			
Resistance to solvents	MIL-STD-202, Method 215			
Solderability	MIL-STD-202, Method 208			
ESD Classification	MIL-STD-883, Method 3015, Class 1HBM 0 to 1,999V			
Moisture Sensitivity Level	J-STD-020, MSL=1			

#### Please contact Q-Tech for higher shock requirements



# **Period Jitter**

As data rates increase, effects of jitter become critical with its budgets tighter. Jitter is the deviation of a timing event of a signal from its ideal position. Jitter is complex and is composed of both random and deterministic jitter components. Random jitter (RJ) is theoretically unbounded and Gaussian in distribution. Deterministic jitter (DJ) is bounded and does not follow any predictable distribution. DJ is also referred to as systematic jitter. A technique to measure period jitter (RMS) one standard deviation (1 $\sigma$ ) and peak-to-peak jitter in time domain is to use a high sampling rate (>8G samples/s) digitizing oscilloscope. Figure shows an example of peak-to-peak jitter and RMS jitter (1 $\sigma$ ) of a QT24L-20MHz, at 3.3Vdc.



#### **Phase Noise and Phase Jitter Integration**

Phase noise is measured in the frequency domain, and is expressed as a ratio of signal power to noise power measured in a 1Hz bandwidth at an offset frequency from the carrier, e.g. 10Hz, 100Hz, 1kHz, 10kHz, 100kHz, etc. Phase noise measurement is made with an Agilent E5052A Signal Source Analyzer (SSA) with built-in outstanding low-noise DC power supply source. The DC source is floated from the ground and isolated from external noise to ensure accuracy and repeatability.

In order to determine the total noise power over a certain frequency range (bandwidth), the time domain must be analyzed in the frequency domain, and then reconstructed in the time domain into an rms value with the unwanted frequencies excluded. This may be done by converting L(f) back to  $S\varphi(f)$  over the bandwidth of interest, integrating and performing some calculations.

Symbol	Definition
$\int \mathcal{L}(\mathbf{f})$	Integrated single side band phase noise (dBc)
$S\phi(f)=(180/\Pi)x\sqrt{2\int \mathcal{L}(f)df}$	Spectral density of phase modulation, also known as RMS phase error (in degrees)
RMS jitter = $S\phi$ (f)/(fosc.360°)	Jitter(in seconds) due to phase noise. Note $S\phi$ (f) in degrees.

The value of RMS jitter over the bandwidth of interest, e.g. 10kHz to 20MHz, 10Hz to 20MHz, represents 1 standard deviation of phase jitter contributed by the noise in that defined bandwidth.

Figure below shows a typical Phase Noise/Phase jitter of a QT24HC, 5.0Vdc, 24MHz and QT24L, 3.3Vdc, 24MHz clock at offset frequencies 10Hz to 5MHz, and phase jitter integrated over the bandwidth of 12kHz to 1MHz.

