

SBOS152A – AUGUST 1987 – REVISED OCTOBER 2003

Precision Gain = 10 DIFFERENTIAL AMPLIFIER

FEATURES

- ACCURATE GAIN: $\pm 0.025\%$ max
- HIGH COMMON-MODE REJECTION: 86dB min
- NONLINEARITY: 0.001% max
- EASY TO USE
- PLASTIC 8-PIN DIP, SO-8 SOIC PACKAGES

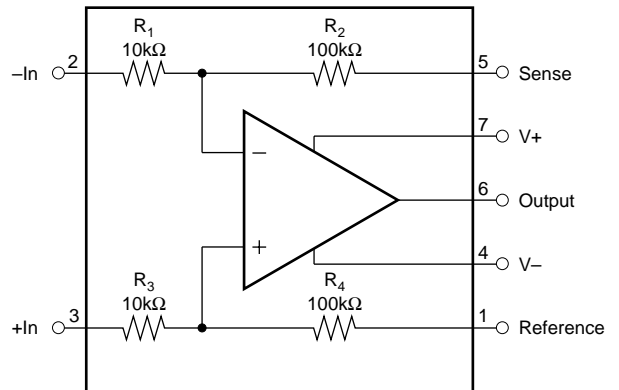
APPLICATIONS

- G = 10 DIFFERENTIAL AMPLIFIER
- G = +10 AMPLIFIER
- G = -10 AMPLIFIER
- G = +11 AMPLIFIER
- INSTRUMENTATION AMPLIFIER

DESCRIPTION

The INA106 is a monolithic Gain = 10 differential amplifier consisting of a precision op amp and on-chip metal film resistors. The resistors are laser trimmed for accurate gain and high common-mode rejection. Excellent TCR tracking of the resistors maintains gain accuracy and common-mode rejection over temperature.

The differential amplifier is the foundation of many commonly used circuits. The INA106 provides this precision circuit function without using an expensive resistor network. The INA106 is available in 8-pin plastic DIP and SO-8 surface-mount packages.



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SPECIFICATIONS

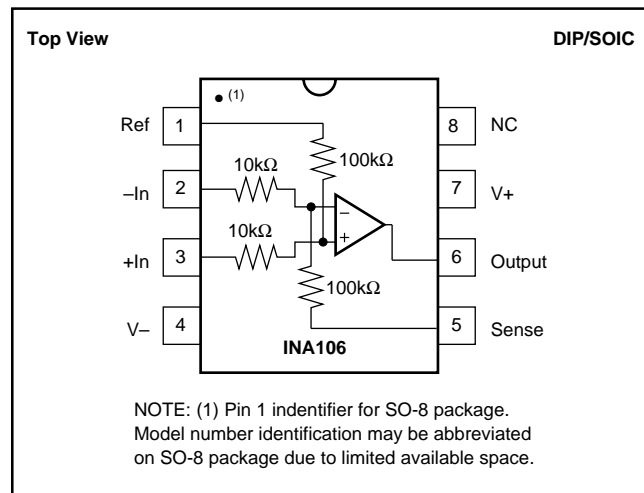
ELECTRICAL

At +25°C, $V_S = \pm 15V$, unless otherwise specified.

PARAMETER	CONDITIONS	INA106KP, U			UNITS
		MIN	TYP	MAX	
GAIN Initial ⁽¹⁾ Error vs Temperature Nonlinearity ⁽²⁾			10 0.01 -4 0.0002	0.025	V/V % ppm/°C %
OUTPUT Related Voltage Rated Current Impedance Current Limit Capacitive Load	$I_O = +20mA, -5mA$ $V_O = 10V$ To Common Stable Operation	10 +20, -5	12 0.01 +40/-10 1000		V mA Ω mA pF
INPUT Impedance Voltage Range Common-Mode Rejection ⁽³⁾	Differential Common-Mode Differential Common-Mode $T_A = T_{MIN}$ to T_{MAX}	 ± 1 ± 11 86	10 110 100		k Ω k Ω V V dB
OFFSET VOLTAGE Initial vs Temperature vs Supply vs Time	RTI ⁽⁴⁾ $\pm V_S = 6V$ to $18V$		50 0.2 1 10	200 10	μV $\mu V/^\circ C$ $\mu V/V$ $\mu V/mo$
NOISE VOLTAGE $f_B = 0.01Hz$ to $10Hz$ $f_O = 10kHz$	RTI ⁽⁵⁾		1 30		$\mu Vp-p$ nV/ \sqrt{Hz}
DYNAMIC RESPONSE Small Signal Full Power BW Slew Rate Settling Time: 0.1% 0.01% 0.01%	-3dB $V_O = 20Vp-p$ $V_O = 10V$ Step $V_O = 10V$ Step $V_{CM} = 10V$ Step, $V_{DIFF} = 0V$	30 2	5 50 3 5 10 5		MHz kHz V/ μs μs μs μs
POWER SUPPLY Rated Voltage Range Quiescent Current	Derated Performance $V_O = 0V$	± 5	± 15 ± 1.5	± 18 ± 2	V V mA
TEMPERATURE RANGE Specification Operation Storage		0 -40 -65		+70 +85 +150	°C °C °C

NOTES: (1) Connected as difference amplifier (see Figure 1). (2) Nonlinearity is the maximum peak deviation from the best-fit straight line as a percent of full-scale peak-to-peak output. (3) With zero source impedance (see "Maintaining CMR" section). (4) Includes effects of amplifiers' input bias and offset currents. (5) Includes effect of amplifier's input current noise and thermal noise contribution of resistor network.

PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS

Power Supply Voltage	±18V
Input Voltage Range	±V _S
Operating Temperature Range: P, U	-40°C to +85°C
Storage Temperature Range	-40°C to +85°C
Lead Temperature (soldering, 10s): P	+300°C
Wave Soldering (3s, max) U	+260°C
Output Short Circuit to Common	Continuous

NOTE: (1) Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.

ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

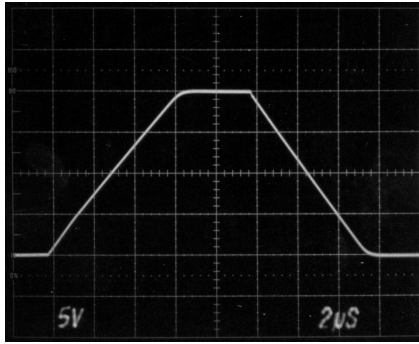
PACKAGE/ORDERING INFORMATION

For the most current package and ordering information, see the Package Option Addendum located at the end of this data sheet.

TYPICAL PERFORMANCE CURVES

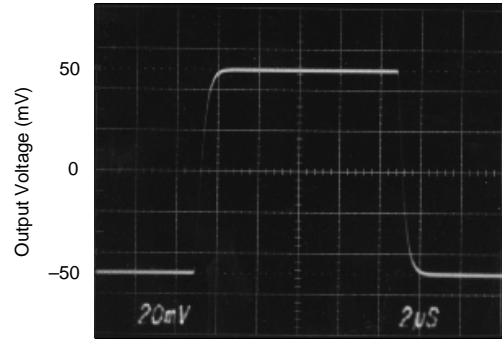
At $T_A = +25^\circ\text{C}$, $V_S = \pm 15\text{V}$, unless otherwise noted.

STEP RESPONSE



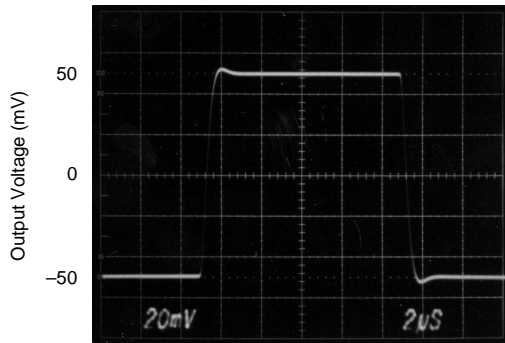
2µs/div

SMALL SIGNAL RESPONSE
(No Load)



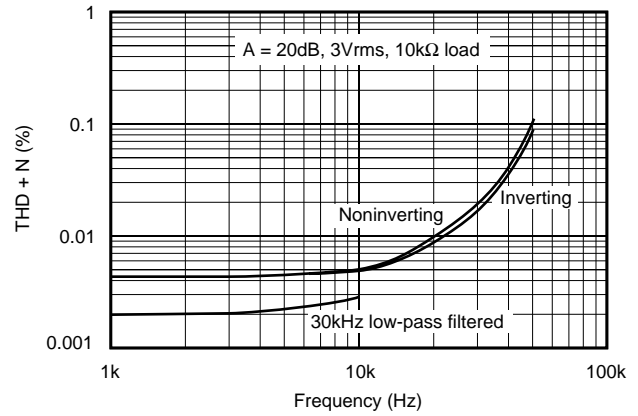
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SMALL SIGNAL RESPONSE
($R_{LOAD} = \infty$, $C_{LOAD} = 100\text{pF}$)

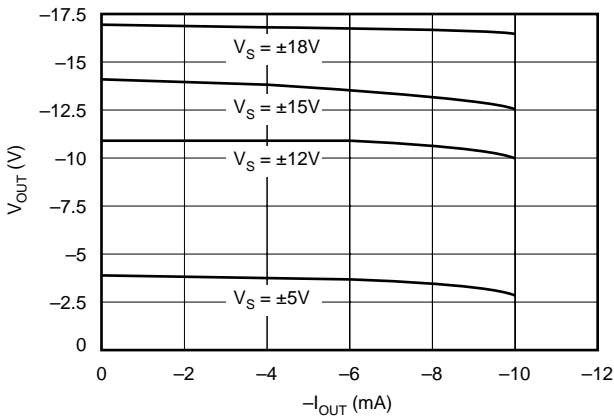


2µs/div

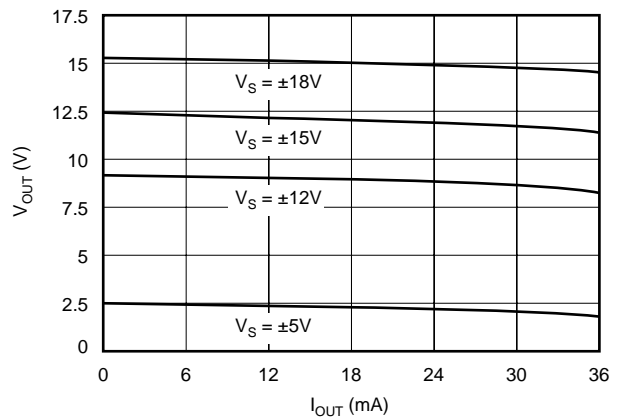
TOTAL HARMONIC DISTORTION AND NOISE
vs FREQUENCY



MAXIMUM V_{OUT} vs I_{OUT}
(Negative Swing)

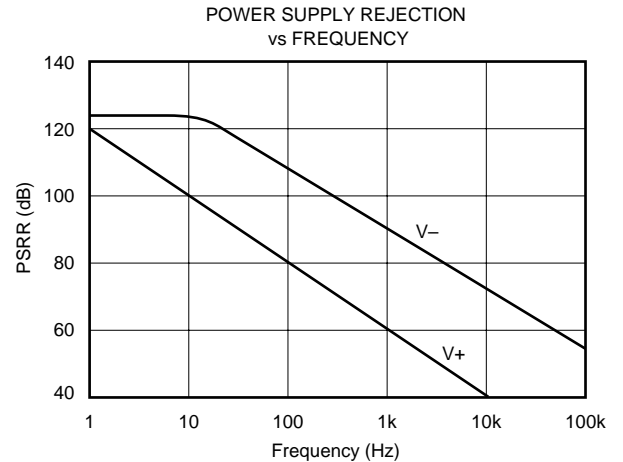
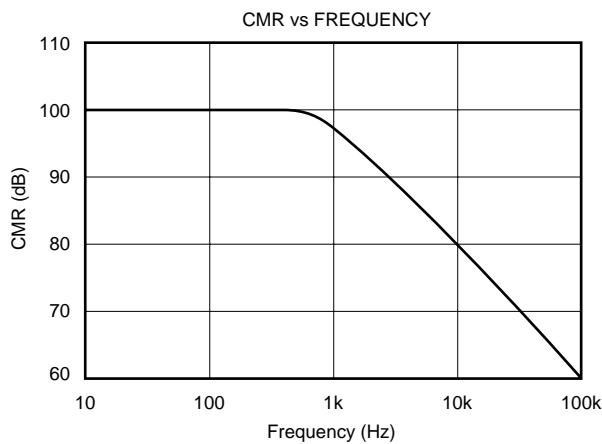


MAXIMUM V_{OUT} vs I_{OUT}
(Positive Swing)



TYPICAL PERFORMANCE CURVES (CONT)

At $T_A = +25^\circ\text{C}$, $V_S = \pm 15\text{V}$, unless otherwise noted.



APPLICATIONS INFORMATION

Figure 1 shows the basic connections required for operation of the INA106. Power supply bypass capacitors should be connected close to the device pins as shown.

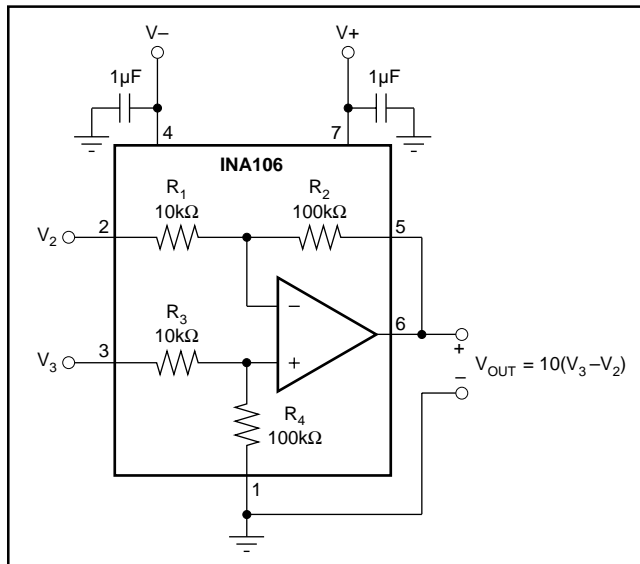


FIGURE 1. Basic Power Supply and Signal Connections.

The differential input signal is connected to pins 2 and 3 as shown. The source impedance connected to the inputs must be equal to assure good common-mode rejection. A 5Ω mismatch in source impedance will degrade the common-mode rejection of a typical device to approximately 86dB. If the source has a known source impedance mismatch, an additional resistor in series with one input can be used to preserve good common-mode rejection.

The output is referred to the output reference terminal (pin 1) which is normally grounded. A voltage applied to the

Ref terminal will be summed with the output signal. The source impedance of a signal applied to the Ref terminal should be less than 10Ω to maintain good common-mode rejection.

Figure 2 shows a voltage applied to pin 1 to trim the offset voltage of the INA106. The known 100Ω source impedance of the trim circuit is compensated by the 10Ω resistor in series with pin 3 to maintain good CMR.

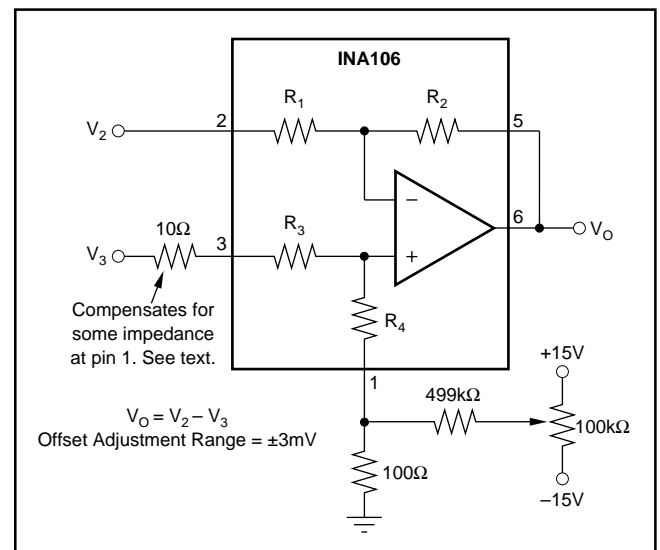


FIGURE 2. Offset Adjustment.

Referring to Figure 1, the CMR depends upon the match of the internal R_4/R_3 ratio to the R_1/R_2 ratio. A CMR of 106dB requires resistor matching of 0.005%. To maintain high CMR over temperature, the resistor TCR tracking must be better than $2\text{ppm}/^\circ\text{C}$. These accuracies are difficult and expensive to reliably achieve with discrete components.

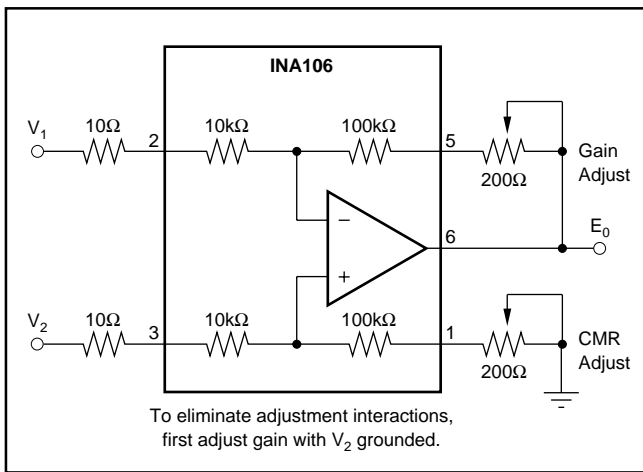


FIGURE 3. Difference Amplifier with Gain and CMR Adjust.

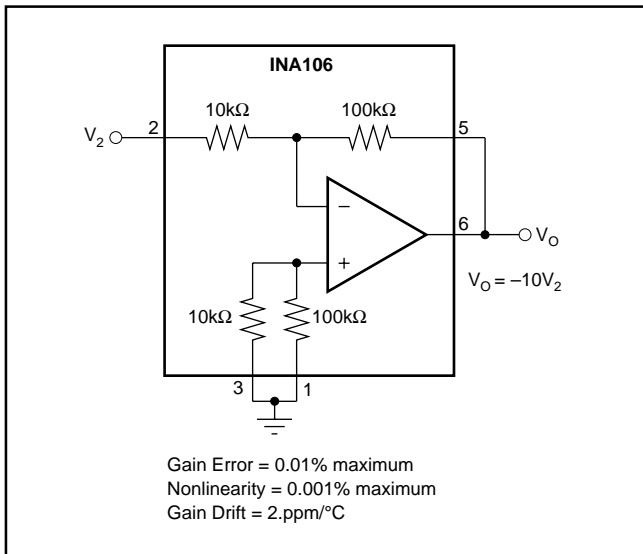


FIGURE 4. Precision $G = -10$ Inverting Amplifier.

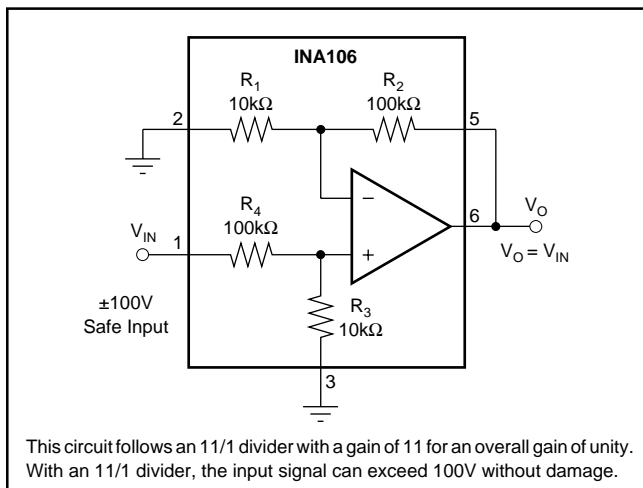
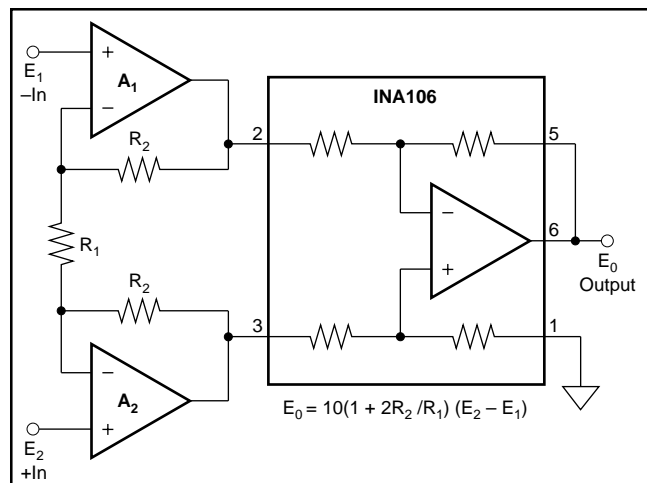


FIGURE 5. Voltage Follower with Input Protection.



To make a high performance high gain instrumentation amplifier, the INA106 can be combined with state-of-the-art op amps. For low source impedance applications, OPA37s will give the best noise, offset, and temperature drift. At source impedances above about 10kΩ, the bias current noise of the OPA37 reacting with input impedance degrades noise. For these applications, use an OPA111 or a dual OPA2111 FET input op amp for lower noise. For an electrometer grade IA, use the OPA128—see table below.

Using the INA106 for the difference amplifier also extends the input common-mode range of the instrumentation amplifier to ±10V. A conventional IA with a unity-gain difference amplifier has an input common-mode range limited to ±5V for an output swing of ±10V. This is because a unity-gain difference amp needs ±5V at the input for 10V at the output, allowing only 5V additional for common-mode.

A_1, A_2	R_1 (Ω)	R_2 (kΩ)	GAIN (V/V)	CMRR (dB)	I_b (pA)	NOISE AT 1kHz (nV/√Hz)
OPA37A	50.5	2.5	1000	128	40000	4
OPA111B	202	10	1000	110	1	10
OPA128LM	202	10	1000	118	0.075	38

FIGURE 6. Precision Instrumentation Amplifier.

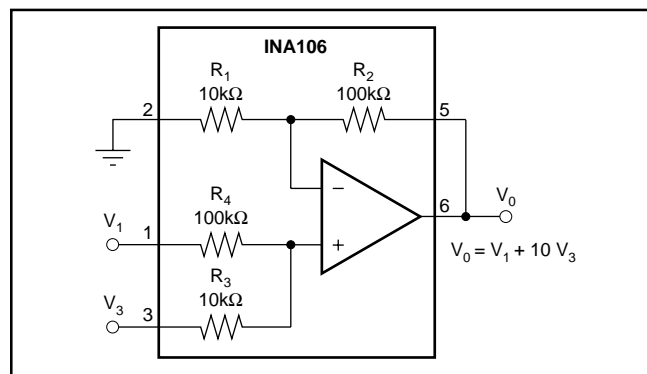


FIGURE 7. Precision Summing Amplifier.

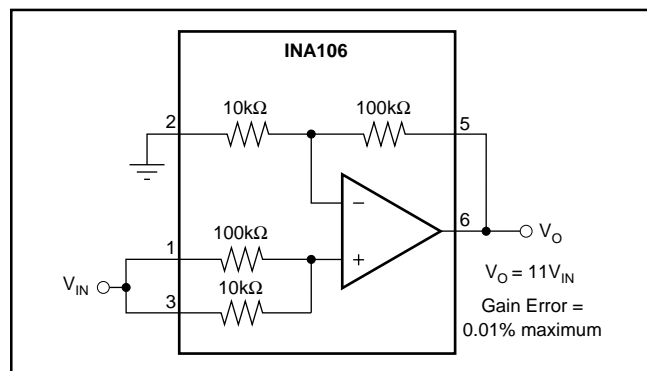


FIGURE 8. Precision $G = 11$ Buffer.

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
INA106KP	ACTIVE	PDIP	P	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
INA106KPG4	ACTIVE	PDIP	P	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
INA106U	ACTIVE	SOIC	D	8	100	Pb-Free (RoHS)	CU NIPDAU	Level-3-260C-168 HR
INA106U/2K5	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
INA106U/2K5G4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
INA106UE4	ACTIVE	SOIC	D	8	100	Pb-Free (RoHS)	CU NIPDAU	Level-3-260C-168 HR

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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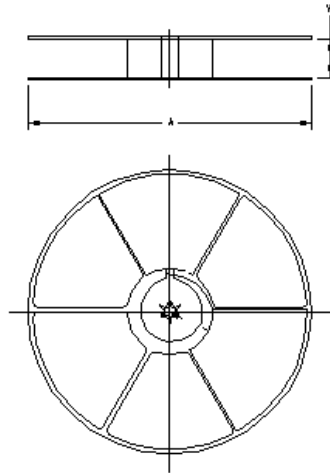
Carrier tape design is defined largely by the component length, width, and thickness.

A_o = Dimension designed to accommodate the component width.
B_o = Dimension designed to accommodate the component length.
K_o = Dimension designed to accommodate the component thickness.
W = Overall width of the carrier tape.
P = Pitch between successive cavity centers.



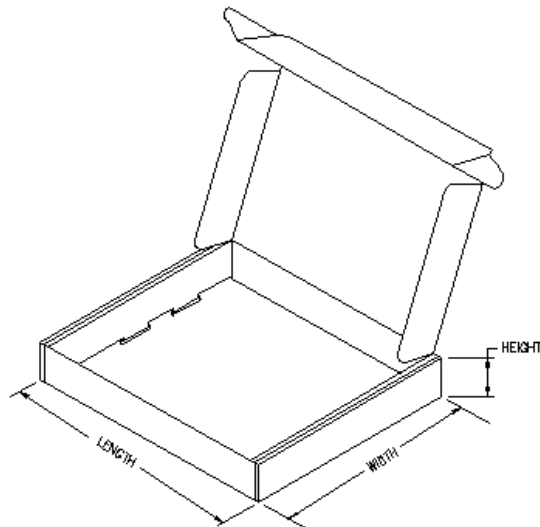
TAPE AND REEL INFORMATION

Device	Package	Pins	Site	Reel Diameter (mm)	Reel Width (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
INA106U/2K5	D	8	MLA	330	12	6.9	5.4	2.0	8	12	Q1



TAPE AND REEL BOX INFORMATION

Device	Package	Pins	Site	Length (mm)	Width (mm)	Height (mm)
INA106U/2K5	D	8	MLA	346.0	346.0	29.0



P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE



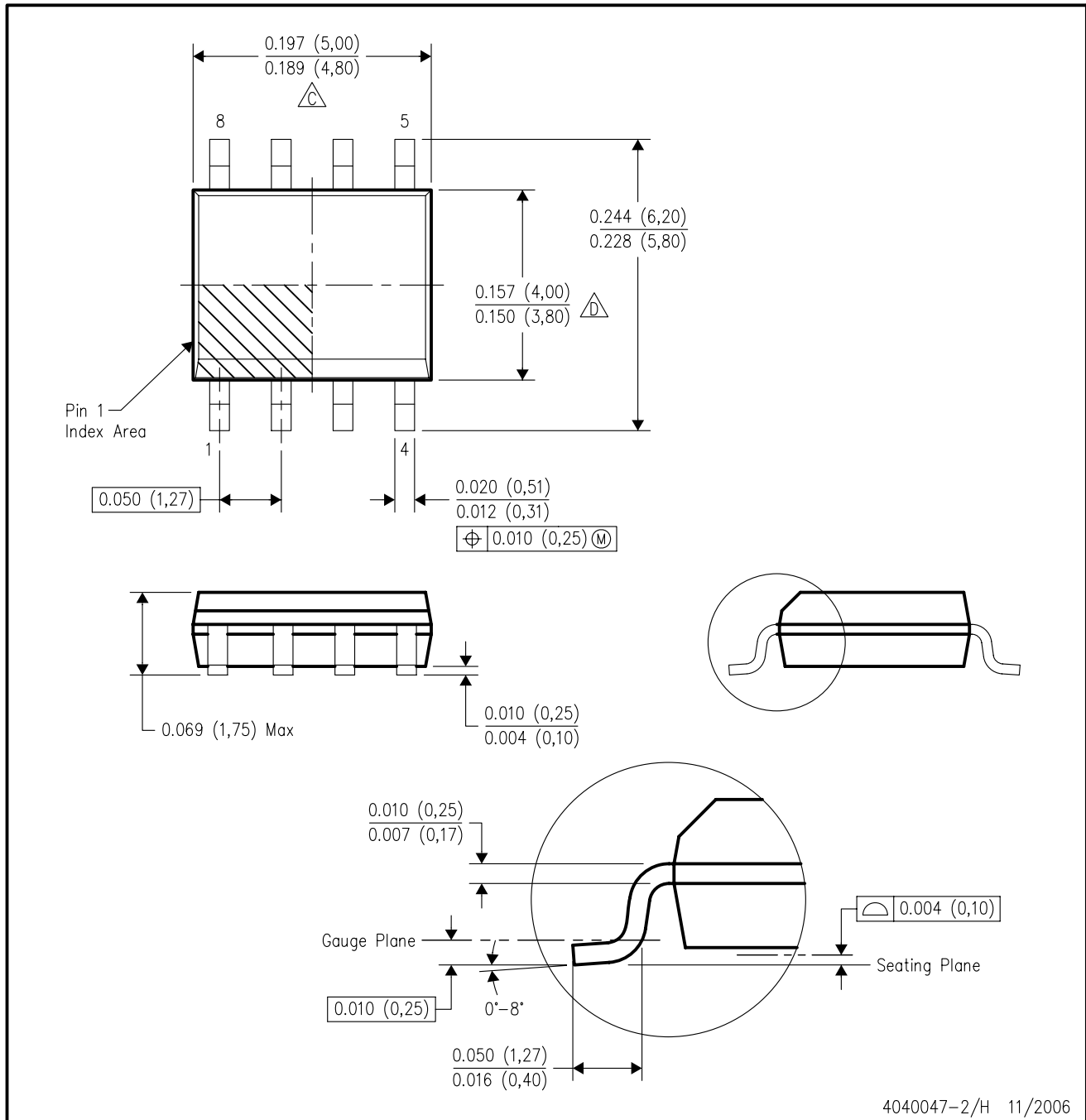
- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. Falls within JEDEC MS-001

For the latest package information, go to http://www.ti.com/sc/docs/package/pkg_info.htm



D (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.
 - D. Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.
 - E. Reference JEDEC MS-012 variation AA.

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