

TPS2223, TPS2224, TPS2226 DUAL-SLOT CARDBUS POWER-INTERFACE SWITCHES FOR SERIAL PCMCIA CONTROLLERS

SLVS317 – MAY 2001

features

- Fully Integrated V_{CC} and V_{pp} Switching for 3.3 V, 5 V, and 12 V (no 12 V on TPS2223)
- Meets Current PC Card™ Standards
- V_{pp} Output Selection Independent of V_{CC}
- 12-V and 5-V Supplies Can Be Disabled
- TTL-Logic Compatible Inputs
- Short-Circuit and Thermal Protection
- 24-Pin HTSSOP, 24- or 30-Pin SSOP
- 140- μ A (Typical) Quiescent Current from 3.3-V Input
- Break-Before-Make Switching
- Power-On Reset
- -40°C to 85°C Operating Ambient Temperature Range

applications

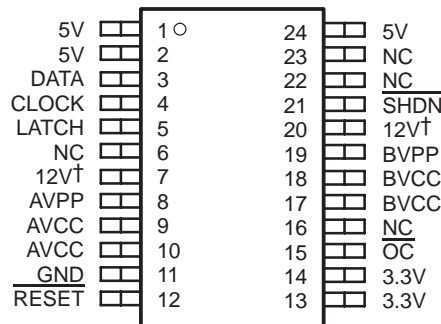
- Notebook and Desktop Computers
- Bar Code Scanners
- Digital Cameras
- Set-Top Boxes
- PDAs

description

The TPS2223, TPS2224 and TPS2226 Card-Bus™ power-interface switches provide an integrated power-management solution for two PC Card sockets. These devices allow the controlled distribution of 3.3 V, 5 V, and 12 V to each card slot. The current-limiting and thermal-protection features eliminate the need for fuses. Current-limit reporting helps the user isolate a system fault. The switch $r_{DS(on)}$ and current-limit values have been set for the peak and average current requirements stated in the PC Card specification, and optimized for cost.

Like the TPS2214 and TPS2214A and the TPS2216 and TPS2216A, this family of devices supports independent VPP/VCC switching; however, the standby and interface-mode pins are not supported. Shutdown mode is now supported independently on SHDN as well as in the serial interface. Optimized for lower power implementation, the TPS2223 does not support 12-V switching to VPP. See the available options table for pin-compatible device information.

TPS2223, TPS2224
DB OR PWP PACKAGE
(TOP VIEW)



NC – No internal connection

† Pin 7 and 20 are NC for TPS2223.

AVAILABLE OPTIONS

| T_A | PACKAGED DEVICES | | | | |
|---------------|-----------------------|-------------------|-----------------|----------------------------|---|
| | PLASTIC SMALL OUTLINE | | | | PowerPAD™ PLASTIC SMALL OUTLINE (PWP-24)† |
| | DB-24 | | DB-30 | | |
| -40°C to 85°C | TPS2223DB, TPS2224DB | | TPS2226DB | | TPS2223PWP, TPS2224PWP |
| | Pin compatibles | TPS2214, TPS2214A | Pin compatibles | TPS2216, TPS2216A, TPS2206 | |

† The DB and PWP packages are also available taped and reeled. Add R suffix to device type (e.g., TPS2223PWPR) for taped and reeled.



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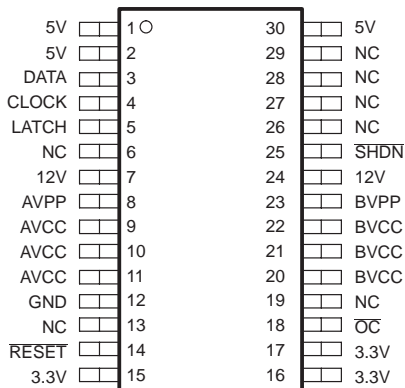
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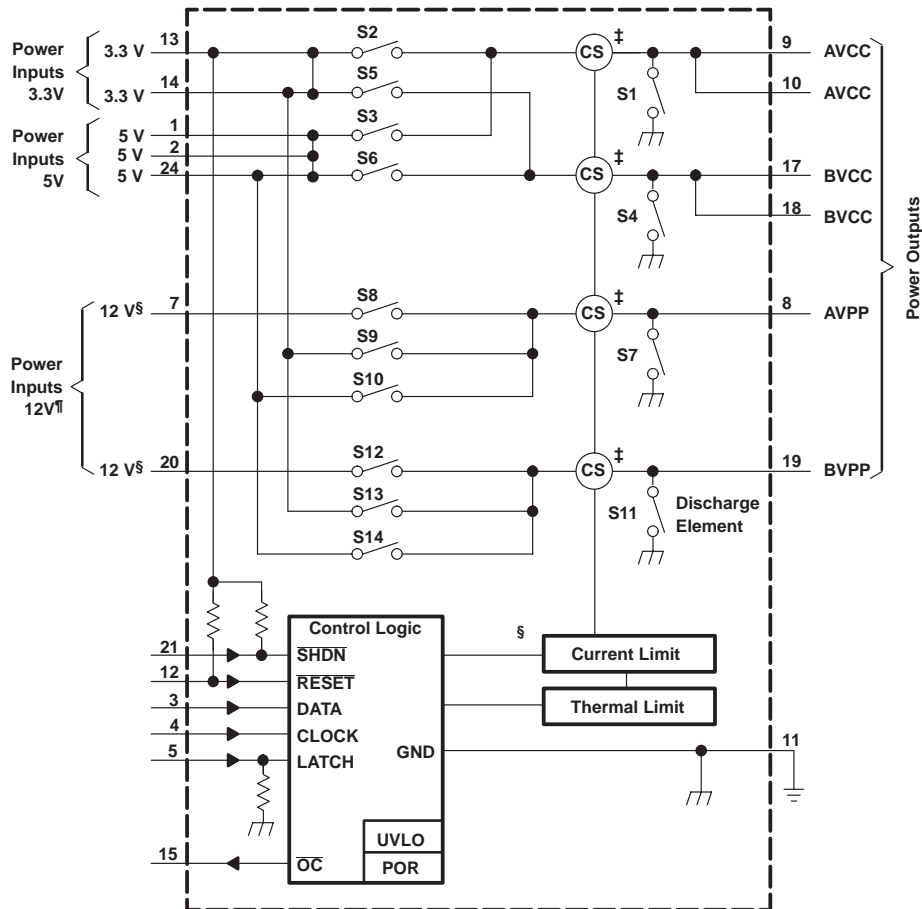
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TPS2226
DB PACKAGE
(TOP VIEW)



NC – No internal connection

functional block diagram†



† Diagram shown for 24-pin DB package.

‡ Current sense

§ The two 12-V pins must be externally connected.

¶ No Connections for TPS2223



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Terminal Functions

| TERMINAL | | | | I/O | DESCRIPTION |
|---------------------------|-------------------------|------------------|---------------------|-----|---|
| NAME | NO. | | | | |
| | TPS2223 | TPS2224 | TPS2226 | | |
| 3.3V | 13, 14 | 13, 14 | 15, 16, 17 | I | 3.3-V input for card power and chip power |
| 5V | 1, 2, 24 | 1, 2, 24 | 1, 2, 30 | I | 5-V input for card power |
| 12V | NA | 7, 20 | 7, 24 | I | 12-V input for card power (xVPP). The two 12-V pins must be externally connected. |
| AVCC | 9, 10 | 9, 10 | 9, 10, 11 | O | Switched output that delivers 3.3 V, 5 V, ground or high impedance to card |
| AVPP | 8 | 8 | 8 | O | Switched output that delivers 3.3 V, 5 V, 12 V, ground or high impedance to card (12 V not applicable to TPS2223) |
| BVCC | 17, 18 | 17, 18 | 20, 21, 22 | O | Switched output that delivers 3.3 V, 5 V, ground or high impedance to card |
| BVPP | 19 | 19 | 23 | O | Switched output that delivers 3.3 V, 5 V, 12 V, ground or high impedance to card (12 V not applicable for TPS2223) |
| GND | 11 | 11 | 12 | | Ground |
| $\overline{\text{OC}}$ | 15 | 15 | 18 | O | Open-drain overcurrent reporting output that goes low when an overcurrent condition exists. An external pullup is required. |
| $\overline{\text{SHDN}}$ | 21 | 21 | 25 | I | Hi-Z (open) all switches. Identical function to serial D8. Asynchronous active-low command, internal pullup |
| $\overline{\text{RESET}}$ | 12 | 12 | 14 | I | Logic-level RESET input active low. Asynchronous active-low command, internal pullup |
| CLOCK | 4 | 4 | 4 | I | Logic-level clock for serial data word |
| DATA | 3 | 3 | 3 | I | Logic-level serial data word |
| LATCH | 5 | 5 | 5 | I | Logic-level latch for serial data word, internal pulldown |
| NC | 6, 7, 16, 20, 22, 23 | 6, 16, 22, 23 | 6, 13, 19, 26–29 | | No internal connection |

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absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

| | |
|---|------------------------------|
| Input voltage range for card power: $V_{I(3.3V)}$ | -0.3 V to 5.5 V |
| $V_{I(5V)}$ | -0.3 V to 5.5 V |
| $V_{I(12V)}^{\ddagger}$ | -0.3 V to 14 V |
| Logic input/output voltage | -0.3 V to 6 V |
| Output voltage: $V_{O(xVCC)}$ | -0.3 V to 6 V |
| $V_{O(xVPP)}$ | -0.3 V to 14 V |
| Continuous total power dissipation | See Dissipation Rating Table |
| Output current: $I_{O(xVCC)}$ | Internally Limited |
| $I_{O(xVPP)}$ | Internally Limited |
| Operating virtual junction temperature range, T_J | -40°C to 100°C |
| Storage temperature range, T_{STG} | -55°C to 150°C |
| Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds) | 260°C |
| \overline{OC} sink current | 10 mA |

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

‡ Not applicable for TPS2223

DISSIPATION RATING TABLE

| PACKAGE§ | | $T_A \leq 25^\circ\text{C}$ POWER RATING | DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$ | $T_A = 70^\circ\text{C}$ POWER RATING | $T_A = 85^\circ\text{C}$ POWER RATING |
|----------|----|---|---|--|--|
| DB | 24 | 890 mW | 8.9 mW/°C | 489 mW | 356 mW |
| | 30 | 1095 mW | 10.95 mW/°C | 602 mW | 438 mW |
| PWP | 24 | 3322 mW | 33.22 mW/°C | 1827 mW | 1329 mW |

§ These devices are mounted on an JEDEC low-k board (2-oz. traces on surface).

recommended operating conditions

| | | MIN | MAX | UNIT |
|---|--|-----|------|------|
| Input voltage, $V_{I(3.3V)}$ is required for all circuit operations. 5V and 12V are only required for their respective functions. | $V_{I(3.3V)}^{\text{¶}}$ | 3 | 3.6 | V |
| | $V_{I(5V)}$ | 3 | 5.5 | |
| | $V_{I(12V)}^{\ddagger}$ | 7 | 13.5 | |
| Output current, I_O | $I_{O(xVCC)}$ at $T_J = 100^\circ\text{C}$ | | 1 | A |
| | $I_{O(xVPP)}$ at $T_J = 100^\circ\text{C}$ | | 100 | mA |
| Clock frequency, $f_{(\text{clock})}$ | | | 2.5 | MHz |
| Pulse duration, t_w | Data | | 200 | ns |
| | Latch | | 250 | |
| | Clock | | 100 | |
| | Reset | | 100 | |
| Data-to-clock hold time (see Figure 2) | | | 100 | ns |
| Data-to-clock setup time, t_{su} (see Figure 2) | | | 100 | ns |
| Latch delay time, $t_{d(\text{latch})}$ (see Figure 2) | | | 100 | ns |
| Clock delay time, $t_{d(\text{clock})}$ (see Figure 2) | | | 250 | ns |
| Operating virtual junction temperature, T_J (maximum to be calculated at worst cast P_D at 85°C ambient) | | -40 | 100 | °C |

‡ Not applicable for TPS2223

¶ It is understood that for $V_{I(3.3V)} < 3$ V, voltages within the absolute maximum ratings applied to pin 5V or pin 12V will not damage the IC.



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electrical characteristics, $T_J = 25^\circ\text{C}$, $V_{I(5V)} = 5\text{ V}$, $V_{I(3.3V)} = 3.3\text{ V}$, $V_{I(12V)} = 12\text{ V}$ (not applicable for TPS2223), all outputs unloaded (unless otherwise noted)

power switch

| PARAMETER | | TEST CONDITIONS† | | MIN | TYP | MAX | UNIT |
|---|--|--|--|---------------------------|-----|----------|------------------|
| $r_{DS(on)}$ | Static drain-source on-state resistance | 3.3V to xVCC, with two switches on | $I_O = 750\text{ mA}$ each | | 85 | 110 | m Ω |
| | | | $I_O = 750\text{ mA}$ each, $T_J = 100^\circ\text{C}$ | | 110 | 140 | |
| | | 5V to xVCC, with two switches on | $I_O = 500\text{ mA}$ each | | 95 | 130 | m Ω |
| | | | $I_O = 500\text{ mA}$ each, $T_J = 100^\circ\text{C}$ | | 120 | 160 | |
| | | 3.3V or 5V to xVPP, with two switches on | $I_O = 50\text{ mA}$ each | | 0.8 | 1 | Ω |
| | | | $I_O = 50\text{ mA}$ each, $T_J = 100^\circ\text{C}$ | | 1 | 1.3 | |
| | 12V to xVPP, with two switches on | $I_O = 50\text{ mA}$ each | | 2 | 2.5 | Ω | |
| | | $I_O = 50\text{ mA}$ each, $T_J = 100^\circ\text{C}$ | | 2.5 | 3.4 | | |
| Output discharge resistance | Discharge at xVCC | $I_{O(disc)} = 1\text{ mA}$ | | 0.5 | 0.7 | 1 | k Ω |
| | Discharge at xVPP | $I_{O(disc)} = 1\text{ mA}$ | | 0.2 | 0.4 | 0.5 | |
| I_{OS} | Short-circuit output current | Limit (steady-state value), output powered into a short circuit | $I_{OS(xVCC)}$ | 1 | 1.4 | 2 | A |
| | | | $I_{OS(xVPP)}$ | 120 | 200 | 300 | mA |
| | | Limit (steady-state value), output powered into a short circuit, $T_J = 100^\circ\text{C}$ | $I_{OS(xVCC)}$ | 1 | 1.4 | 2 | A |
| | | | $I_{OS(xVPP)}$ | 120 | 200 | 300 | mA |
| Thermal shutdown temperature (see Note 1) | Thermal trip point, T_J | Rising temperature | | | 135 | | $^\circ\text{C}$ |
| | Hysteresis, T_J | | | | 10 | | |
| Current-limit response time (see Note 1) | | $V_{O(xVCC)}$ with 100-m Ω short, from short to OC signal falling edge | | | 10 | | μs |
| | | $V_{O(xVPP)}$ with 100-m Ω short, from short to OC signal falling edge | | | 3 | | |
| I_I | Input current, quiescent | Normal operation and reset mode | $I_I(3.3V)$ | | 140 | 200 | μA |
| | | | $I_I(5V)$ | | 8 | 12 | |
| | | | $I_I(12V)$ | | 100 | 180 | |
| | Shutdown mode, $V_{O(xVCC)} = \text{Hi-Z}$, $V_{O(xVPP)} = \text{Hi-Z}$ | $I_I(3.3V)$ | | 0.3 | 2 | | |
| | | $I_I(5V)$ | | 0.1 | 2 | | |
| | | $I_I(12V)$ | | 0.3 | 2 | | |
| I_{lkg} | Leakage current, output off state | Shutdown mode | $V_{O(xVCC)} = 5\text{ V}$, $V_{I(5V)} = V_{I(12V)} = 0$ | $T_J = 100^\circ\text{C}$ | | 10 | μA |
| | | | $V_{O(xVPP)} = 12\text{ V}$, $V_{I(5V)} = V_{I(12V)} = 0$ | $T_J = 100^\circ\text{C}$ | | 50 | |
| | | | $T_J = 100^\circ\text{C}$ | | 10 | | |
| | | | $T_J = 100^\circ\text{C}$ | | 50 | | |

† Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.
 NOTE 1: Specified by design; not tested in production.

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electrical characteristics, $T_J = 25^\circ\text{C}$, $V_{I(5V)} = 5\text{ V}$, $V_{I(3.3V)} = 3.3\text{ V}$, $V_{I(12V)} = 12\text{ V}$ (not applicable for TPS2223), all outputs unloaded (unless otherwise noted) (continued)

logic section (CLOCK, DATA, LATCH, $\overline{\text{RESET}}$, $\overline{\text{SHDN}}$, $\overline{\text{OC}}$)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|---------------------------|---|--|-----|------|-----|---------------|
| I_I | Input current, logic | $\overline{\text{RESET}} = 5.5\text{ V}$ | -1 | | 1 | μA |
| | | $\overline{\text{RESET}} = 0\text{ V}$ | -30 | -20 | -10 | |
| | $I_I(\overline{\text{SHDN}})$ (see Note 2) | $\overline{\text{SHDN}} = 5.5\text{ V}$ | -1 | | 1 | |
| | | $\overline{\text{SHDN}} = 0\text{ V}$ | -50 | | -3 | |
| | $I_I(\text{LATCH})$ (see Note 2) | LATCH = 5.5 V | | | 50 | |
| | | LATCH = 0 V | -1 | | 1 | |
| $I_I(\text{CLOCK, DATA})$ | 0 V to 5.5 V | -1 | | 1 | | |
| V_{IH} | High-level input voltage, logic | | 2 | | | V |
| V_{IL} | Low-level input voltage, logic | | | | 0.8 | V |
| $V_{O(\text{sat})}$ | Output saturation voltage at $\overline{\text{OC}}$ | $I_O = 2\text{ mA}$ | | 0.14 | 0.4 | V |
| I_{Ikg} | Leakage current at $\overline{\text{OC}}$ | $V_{O(\overline{\text{OC}})} = 5.5\text{ V}$ | | 0 | 1 | μA |

NOTE 2: LATCH has low current pulldown. $\overline{\text{RESET}}$ and $\overline{\text{SHDN}}$ have low-current pullup.

UVLO and POR (power-on reset)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|---------------------|--|---|-----|-----|-----|---------------|
| $V_{I(3.3V)}$ | Input voltage at 3.3V pin, UVLO | 3.3 V level below which all switches are Hi-Z | 2.4 | 2.7 | 2.9 | V |
| $V_{hys(3.3V)}$ | UVLO hysteresis voltage at VA (see Note 1) | | 70 | 100 | | mV |
| $V_{I(5V)}$ | Input voltage at 5V pin, UVLO | 5 V level below which only 5V switches are Hi-Z | 2.3 | 2.5 | 2.9 | V |
| $V_{hys(5V)}$ | UVLO hysteresis voltage at 5V (see Note 1) | | 70 | 100 | | mV |
| t_{df} | Delay time for falling response, UVLO (see Note 1) | Delay from voltage hit (step from 3 V to 2.3 V) to Hi-Z control (90% V_G to GND) | | 4 | | μs |
| $V_{I(\text{POR})}$ | Input voltage, power-on reset (see Note 1) | 3.3 V voltage below which POR is asserted causing a RESET internally with all line switches open and all discharge switches closed. | | | 1.7 | V |

NOTE 1: Specified by design; not tested in production.



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switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$, $V_{I(3.3\text{V})} = 3.3\text{ V}$, $V_{I(5\text{V})} = 5\text{ V}$, $V_{I(12)} = 12\text{ V}$ (not applicable for TPS2223) all outputs unloaded (unless otherwise noted)

| PARAMETER† | LOAD CONDITION | TEST CONDITIONS‡ | MIN | TYP | MAX | UNIT |
|---|--|---|-------------|------|-----|------|
| t_r Output rise times (see Note 1) | $C_L(xVCC) = 0.1\ \mu\text{F}$, $C_L(xVPP) = 0.1\ \mu\text{F}$, $I_O(xVCC) = 0\ \text{A}$, $I_O(xVPP) = 0\ \text{A}$ | $V_O(xVCC) = 5\ \text{V}$ | | 0.9 | | ms |
| | | $V_O(xVPP) = 12\ \text{V}$ | | 0.26 | | |
| | $C_L(xVCC) = 150\ \mu\text{F}$, $C_L(xVPP) = 10\ \mu\text{F}$, $I_O(xVCC) = 0.75\ \text{A}$, $I_O(xVPP) = 50\ \text{mA}$ | $V_O(xVCC) = 5\ \text{V}$ | | 1.1 | | |
| | | $V_O(xVPP) = 12\ \text{V}$ | | 0.6 | | |
| t_f Output fall times (see Note 1) | $C_L(xVCC) = 0.1\ \mu\text{F}$, $C_L(xVPP) = 0.1\ \mu\text{F}$, $I_O(xVCC) = 0\ \text{A}$, $I_O(xVPP) = 0\ \text{A}$ | $V_O(xVCC) = 5\ \text{V}$, Discharge switches ON | | 0.5 | | ms |
| | | $V_O(xVPP) = 12\ \text{V}$, Discharge switches ON | | 0.2 | | |
| | $C_L(xVCC) = 150\ \mu\text{F}$, $C_L(xVPP) = 10\ \mu\text{F}$, $I_O(xVCC) = 0.75\ \text{A}$, $I_O(xVPP) = 50\ \text{mA}$ | $V_O(xVCC) = 5\ \text{V}$ | | 2.35 | | |
| | | $V_O(xVPP) = 12\ \text{V}$ | | 3.9 | | |
| t_{pd} Propagation delay (see Note 1) | $C_L(xVCC) = 0.1\ \mu\text{F}$, $C_L(xVPP) = 0.1\ \mu\text{F}$, $I_O(xVCC) = 0\ \text{A}$, $I_O(xVPP) = 0\ \text{A}$ | Latch↑ to xVPP (12 V)§ | t_{pdon} | 2 | | ms |
| | | | t_{pdoff} | 0.62 | | |
| | | Latch↑ to xVPP (5 V) | t_{pdon} | 0.77 | | |
| | | | t_{pdoff} | 0.51 | | |
| | | Latch↑ to xVPP (3.3 V) | t_{pdon} | 0.75 | | |
| | | | t_{pdoff} | 0.52 | | |
| | | Latch↑ to xVCC (5 V) | t_{pdon} | 0.3 | | |
| | | | t_{pdoff} | 2.5 | | |
| | $C_L(xVCC) = 150\ \mu\text{F}$, $C_L(xVPP) = 10\ \mu\text{F}$, $I_O(xVCC) = 0.75\ \text{A}$, $I_O(xVPP) = 50\ \text{mA}$ | Latch↑ to xVPP (12 V)§ | t_{pdon} | 2.2 | | ms |
| | | | t_{pdoff} | 0.8 | | |
| | | Latch↑ to xVPP (5 V) | t_{pdon} | 0.8 | | |
| | | | t_{pdoff} | 0.6 | | |
| | | Latch↑ to xVPP (3.3 V) | t_{pdon} | 0.8 | | |
| | | | t_{pdoff} | 0.6 | | |
| Latch↑ to xVCC (5 V) | t_{pdon} | 0.6 | | | | |
| | t_{pdoff} | 2.5 | | | | |
| Latch↑ to xVCC (3.3V) | t_{pdon} | 0.5 | | | | |
| | t_{pdoff} | 2.6 | | | | |

† Refer to Parameter Measurement Information in Figure 1.

‡ No card inserted, assumes a 0.1- μF output capacitor (see Figure 1).

§ Not applicable for TPS2223

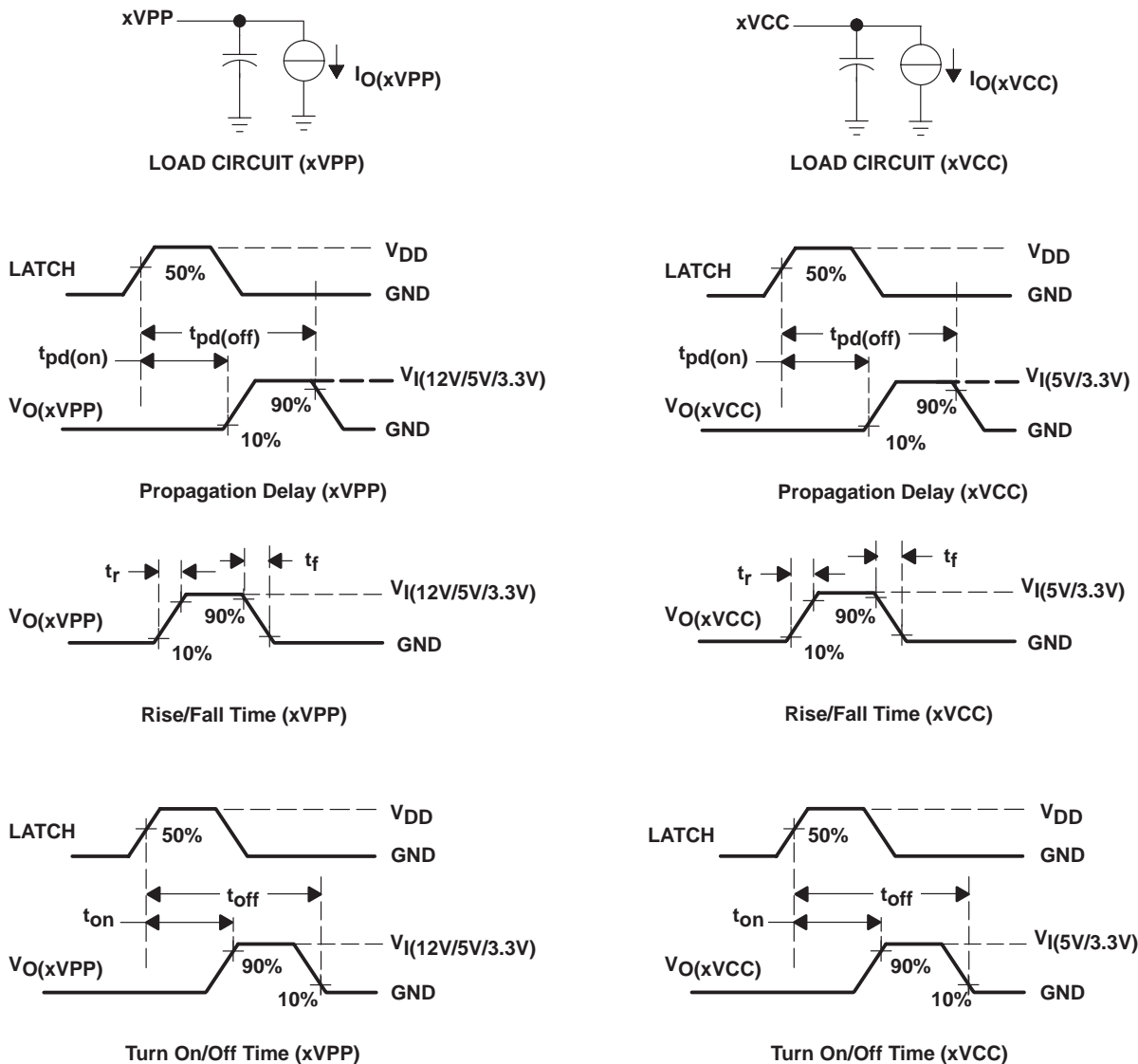
NOTE 1: Specified by design; not tested in production.



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PARAMETER MEASUREMENT INFORMATION



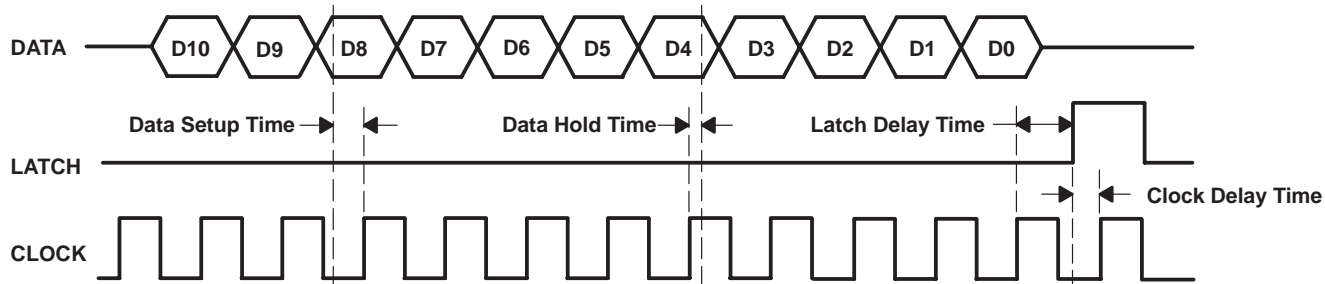
VOLTAGE WAVEFORMS

Figure 1. Test Circuits and Voltage Waveforms

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PARAMETER MEASUREMENT INFORMATION



NOTE: Data is clocked in on the positive edge of the clock. The positive edge of the latch signal should occur before the next positive edge of the clock. For definition of D0 to D10, see the control logic table.

Figure 2. Serial-Interface Timing for TPS2226

Table of Graphs

| | | FIGURE |
|--|-------------------------|--------|
| Short-circuit response, short applied to powered-on 5-V xVCC-switch output | vs Time | 3 |
| Short-circuit response, short applied to powered-on 12-V xVPP-switch output | vs Time | 4 |
| \overline{OC} response with ramped overcurrent-limit load on 5-V xVCC-switch output | vs Time | 5 |
| \overline{OC} response with ramped overcurrent-limit load on 12-V xVPP-switch output | vs Time | 6 |
| xVCC Turnon propagation delay time ($C_L = 150 \mu\text{F}$) | vs Junction temperature | 7 |
| xVCC Turnoff propagation delay time ($C_L = 150 \mu\text{F}$) | vs Junction temperature | 8 |
| xVPP Turnon propagation delay time ($C_L = 10 \mu\text{F}$) | vs Junction temperature | 9 |
| xVPP Turnoff propagation delay time ($C_L = 10 \mu\text{F}$) | vs Junction temperature | 10 |
| xVCC Turnon propagation delay time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 11 |
| xVCC Turnoff propagation delay time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 12 |
| xVPP Turnon propagation delay time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 13 |
| xVPP Turnoff propagation delay time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 14 |
| xVCC Rise time ($C_L = 150 \mu\text{F}$) | vs Junction temperature | 15 |
| xVCC Fall time ($C_L = 150 \mu\text{F}$) | vs Junction temperature | 16 |
| xVPP Rise time ($C_L = 10 \mu\text{F}$) | vs Junction temperature | 17 |
| xVPP Fall time ($C_L = 10 \mu\text{F}$) | vs Junction temperature | 18 |
| xVCC Rise time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 19 |
| xVCC Fall time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 20 |
| xVPP Rise time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 21 |
| xVPP Fall time ($T_J = 25^\circ\text{C}$) | vs Load capacitance | 22 |

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PARAMETER MEASUREMENT INFORMATION

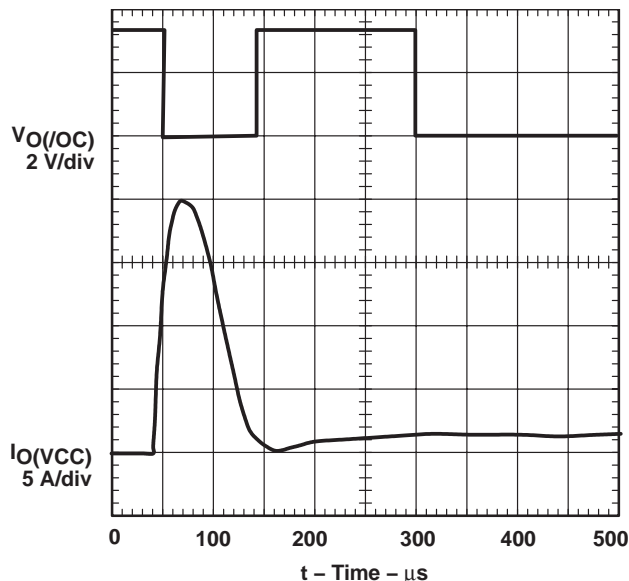


Figure 3. Short-Circuit Response, Short Applied to Powered-on 5-V xVCC-Switch Output

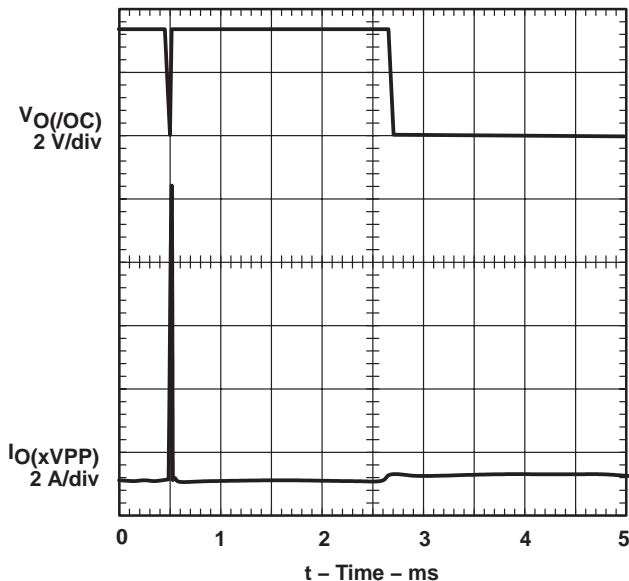


Figure 4. Short-Circuit Response, Short Applied to Powered-on 12-V xVPP-Switch Output

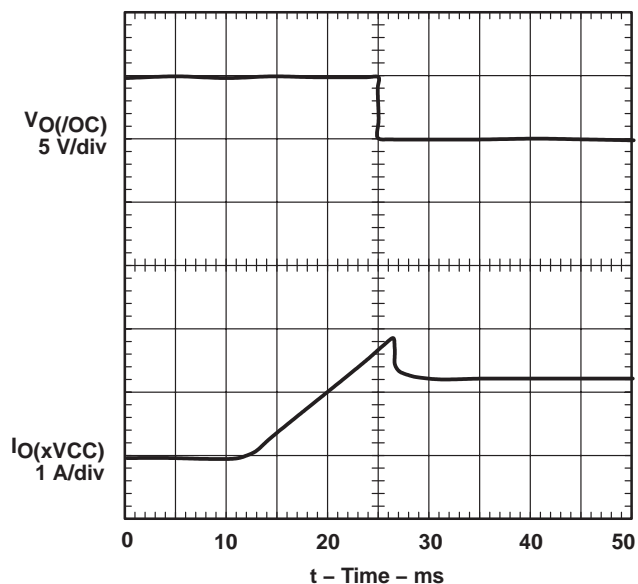


Figure 5. \overline{OC} Response With Ramped Overcurrent-Limit Load on 5-V xVCC-Switch Output

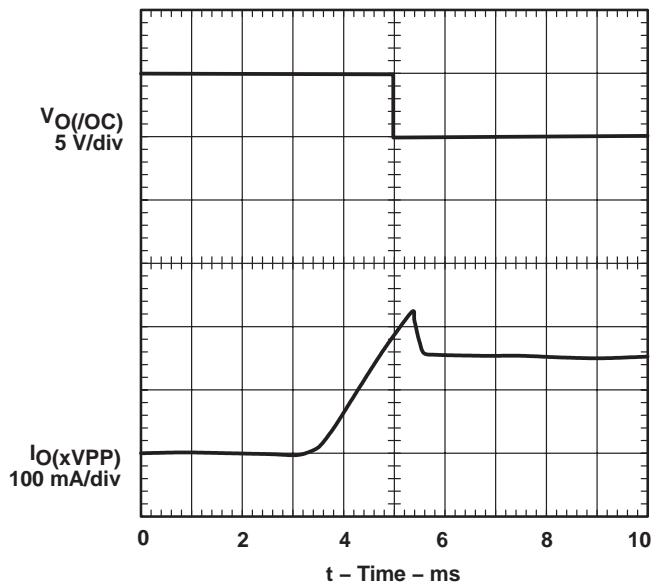


Figure 6. \overline{OC} Response With Ramped Overcurrent-Limit Load on 12-V xVPP-Switch Output

PARAMETER MEASUREMENT INFORMATION

TURNON PROPAGATION DELAY TIME, xVCC
 vs
 JUNCTION TEMPERATURE

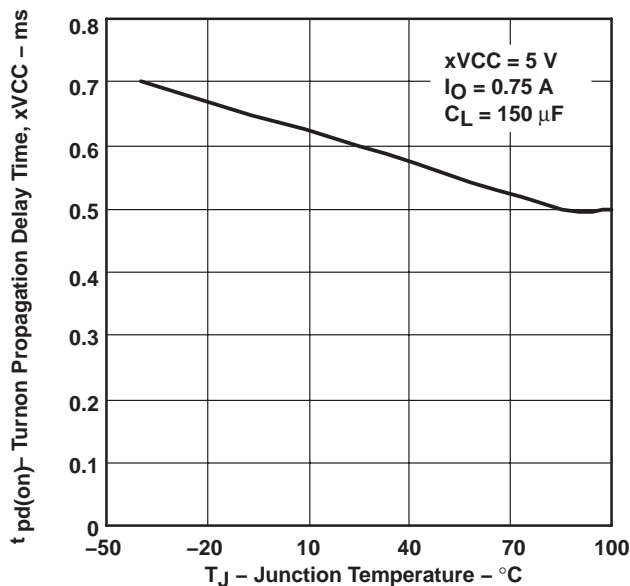


Figure 7

TURNOFF PROPAGATION DELAY TIME, xVCC
 vs
 JUNCTION TEMPERATURE

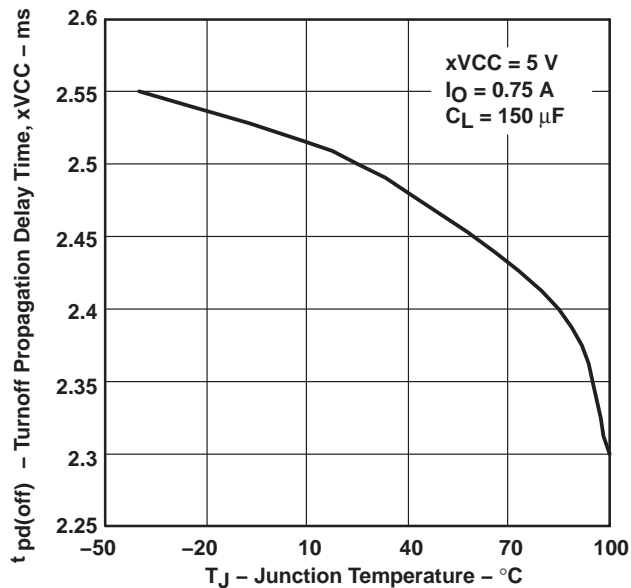


Figure 8

TURNON PROPAGATION DELAY TIME, xVPP
 vs
 JUNCTION TEMPERATURE

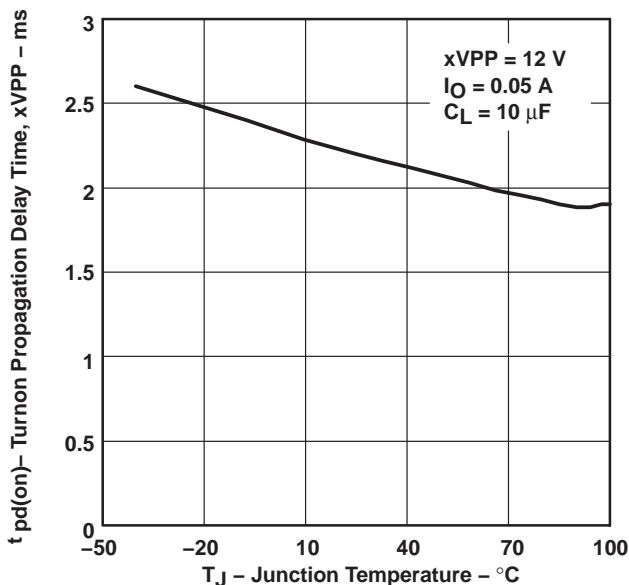


Figure 9

TURNOFF PROPAGATION DELAY TIME, xVPP
 vs
 JUNCTION TEMPERATURE

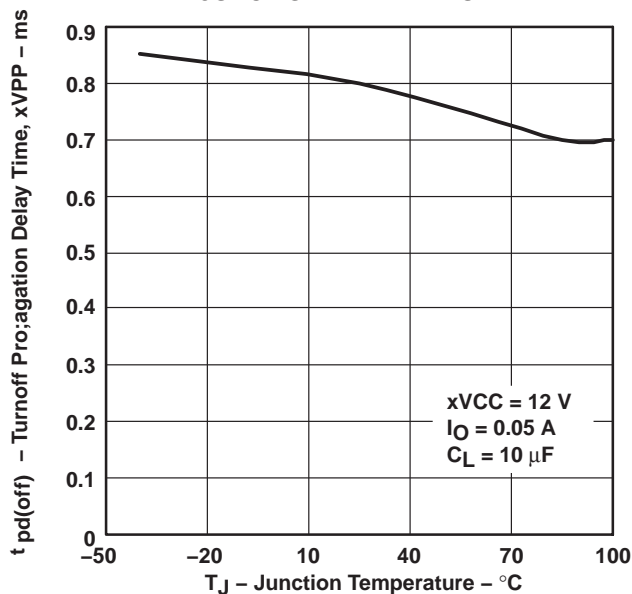


Figure 10

**TPS2223, TPS2224, TPS2226
DUAL-SLOT CARDBUS POWER-INTERFACE SWITCHES
FOR SERIAL PCMCIA CONTROLLERS**

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PARAMETER MEASUREMENT INFORMATION

**TURNON PROPAGATION DELAY TIME, xVCC
VS
LOAD CAPACITANCE**

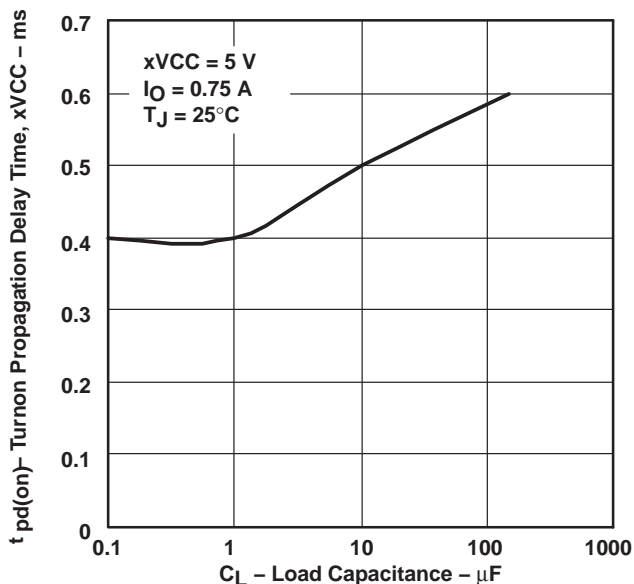


Figure 11

**TURNOFF PROPAGATION DELAY TIME, xVCC
VS
LOAD CAPACITANCE**

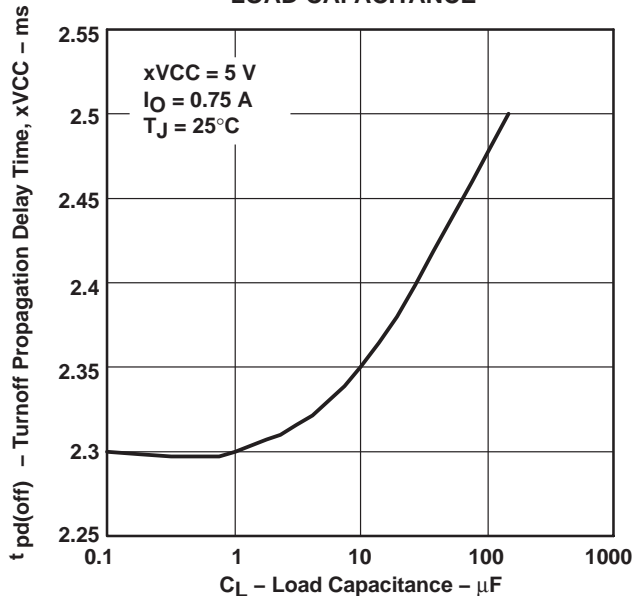


Figure 12

**TURNON PROPAGATION DELAY TIME, xVPP
VS
LOAD CAPACITANCE**

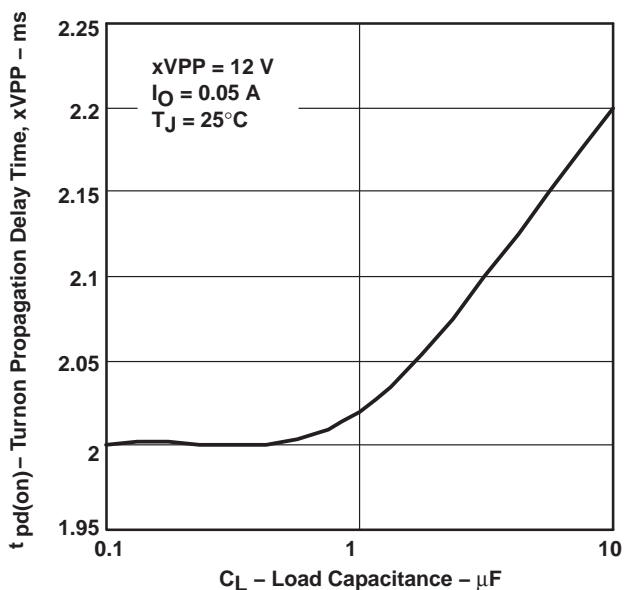


Figure 13

**TURNOFF PROPAGATION DELAY TIME, xVPP
VS
LOAD CAPACITANCE**

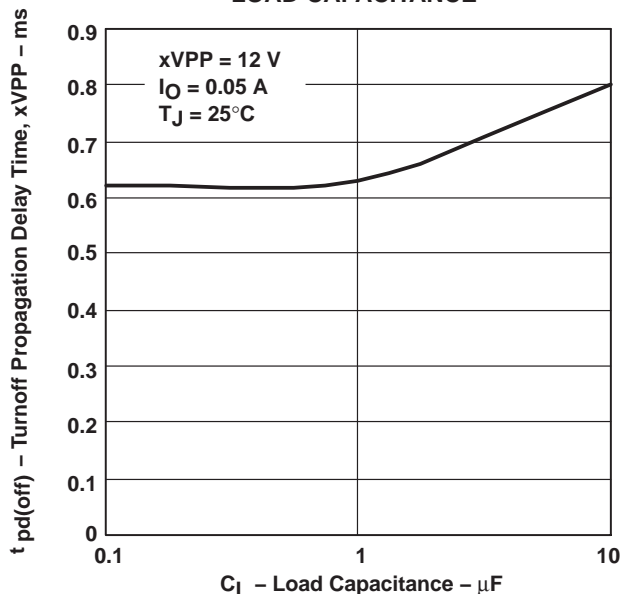
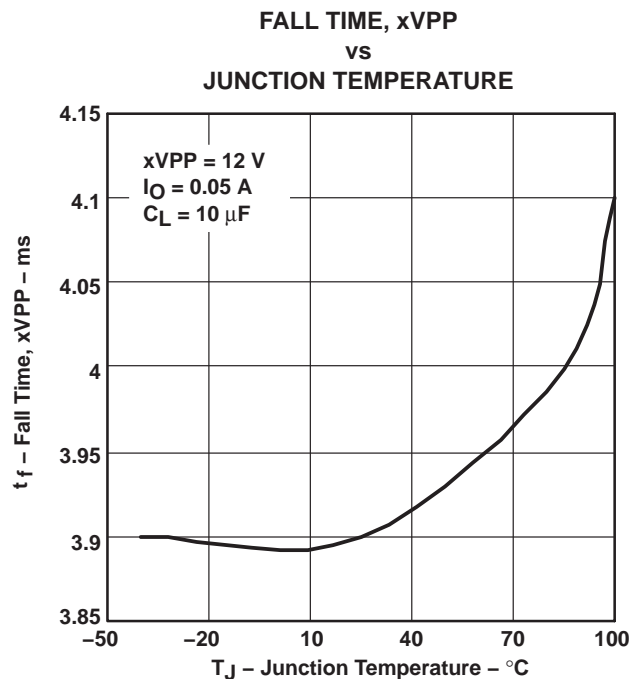
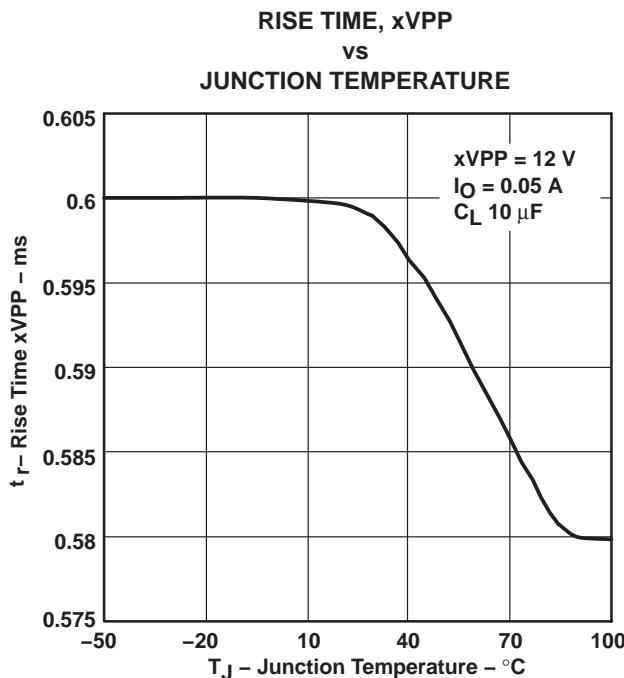
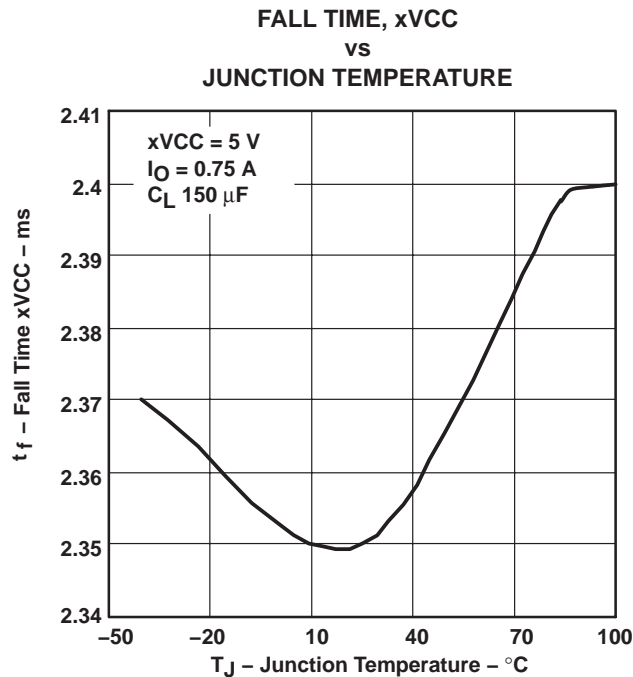
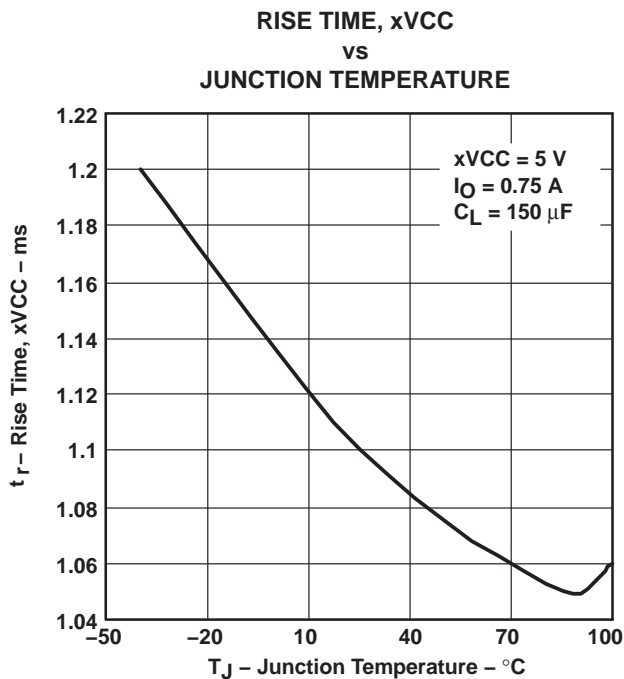


Figure 14



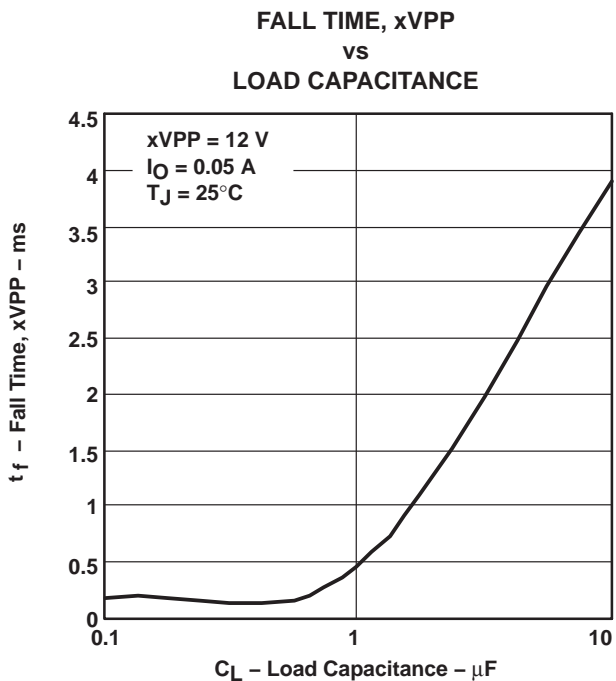
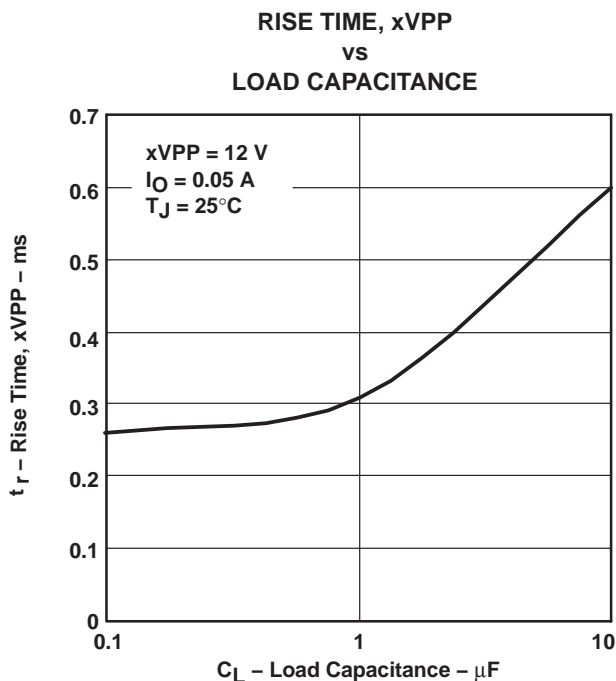
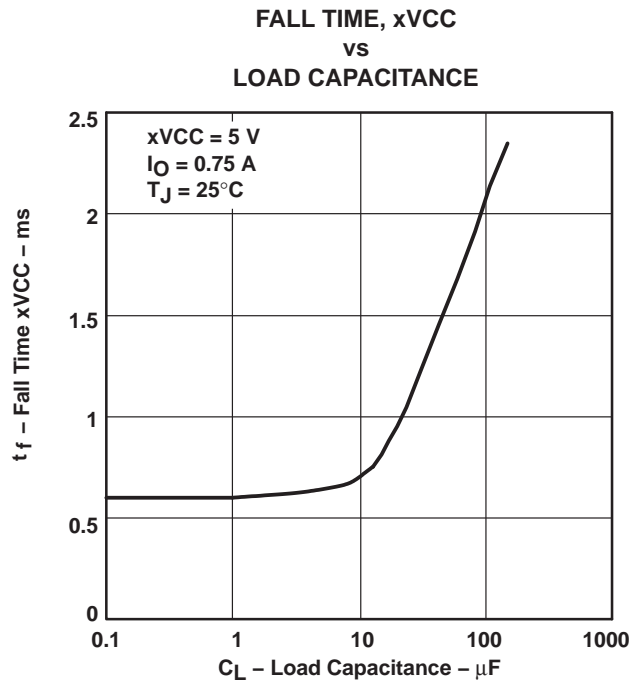
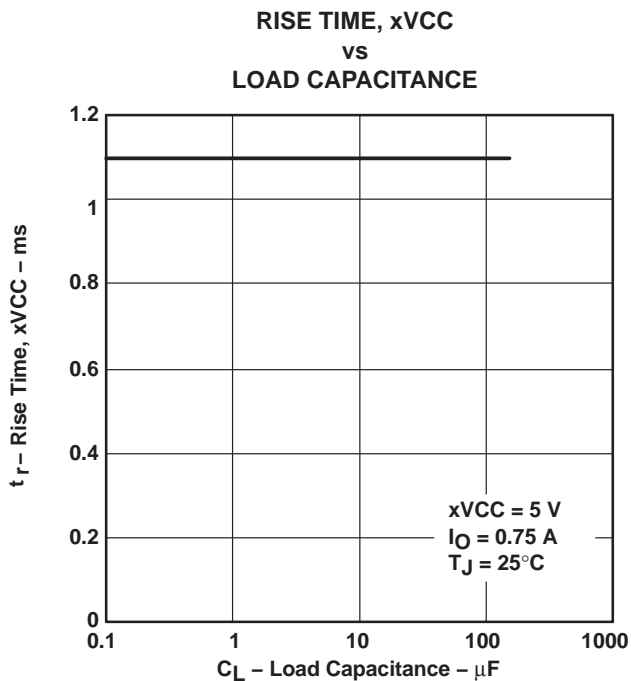
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PARAMETER MEASUREMENT INFORMATION



TYPICAL CHARACTERISTICS

Table of Graphs

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**INPUT CURRENT, $xVCC = 3.3\text{ V}$
vs
JUNCTION TEMPERATURE**

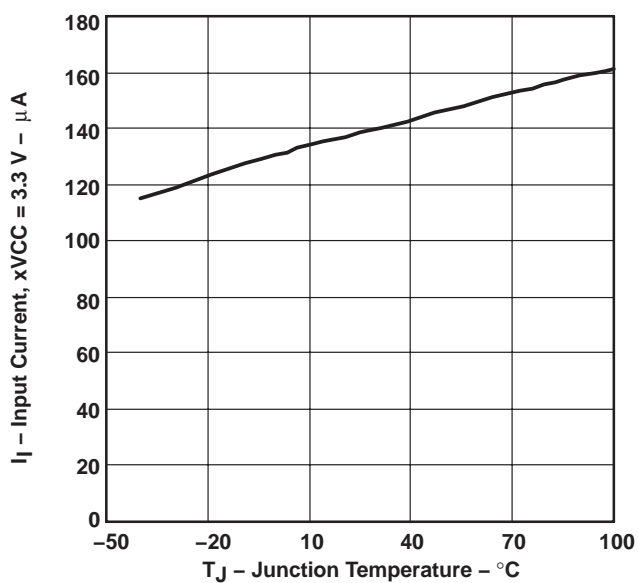


Figure 23

**INPUT CURRENT, $xVCC = 5\text{ V}$
vs
JUNCTION TEMPERATURE**

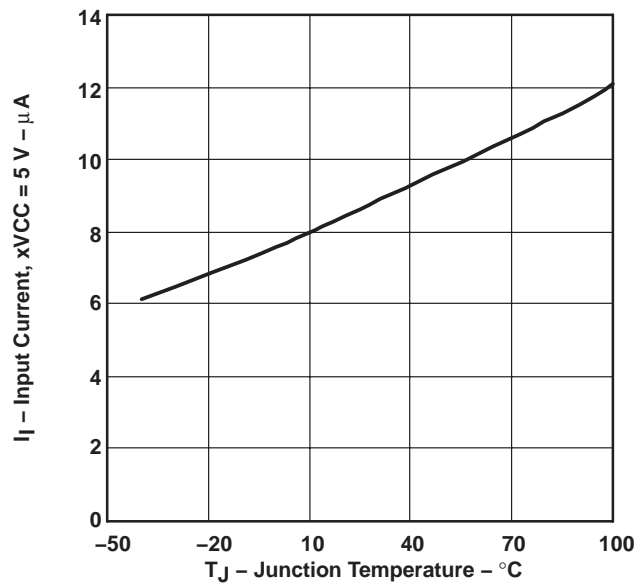


Figure 24

TPS2223, TPS2224, TPS2226
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TYPICAL CHARACTERISTICS

INPUT CURRENT, xVPP = 12 V
vs
JUNCTION TEMPERATURE

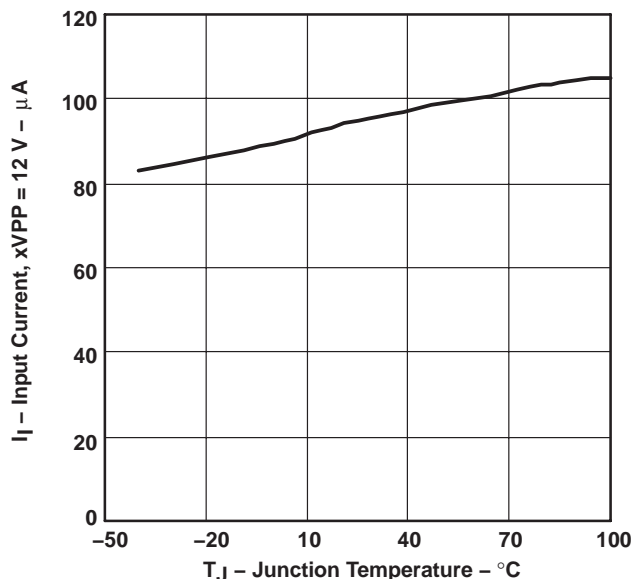


Figure 25

STATIC DRAIN-SOURCE ON-STATE RESISTANCE,
3.3 V TO xVCC SWITCH
vs
JUNCTION TEMPERATURE

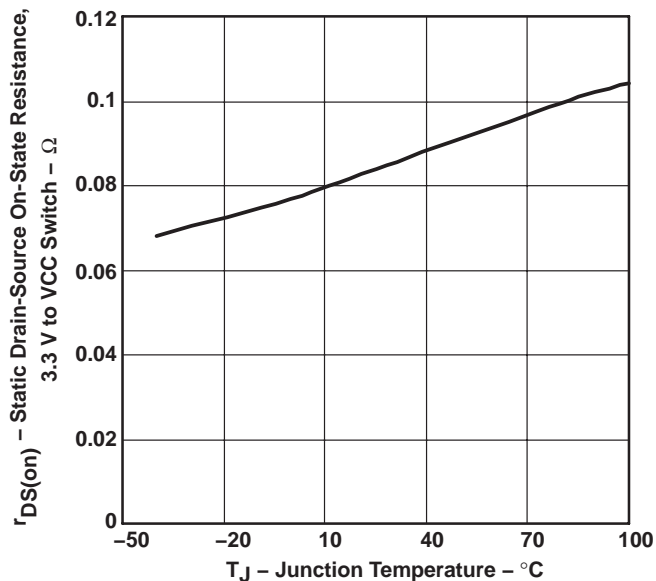


Figure 26

STATIC DRAIN-SOURCE ON-STATE RESISTANCE,
5 V TO xVCC SWITCH
vs
JUNCTION TEMPERATURE

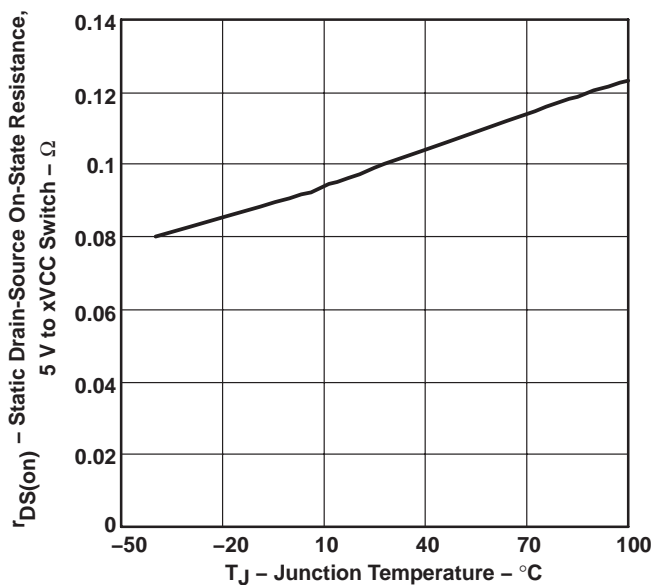


Figure 27

STATIC DRAIN-SOURCE ON-STATE RESISTANCE,
12 V TO xVPP SWITCH
vs
JUNCTION TEMPERATURE

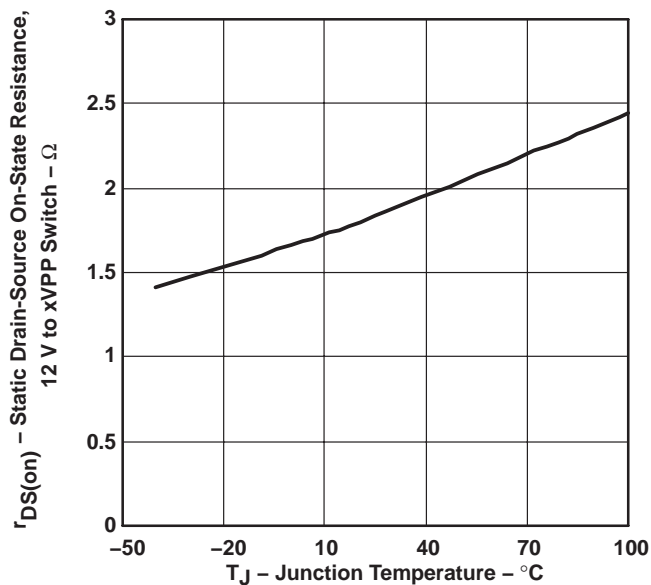


Figure 28



TYPICAL CHARACTERISTICS

xVCC SWITCH VOLTAGE DROP, 3.3-V INPUT
 VS
 LOAD CURRENT

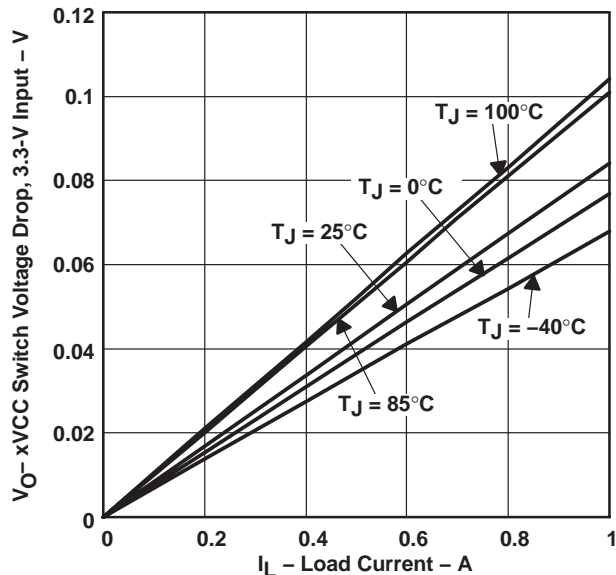


Figure 29

xVCC SWITCH VOLTAGE DROP, 5-V INPUT
 VS
 LOAD CURRENT

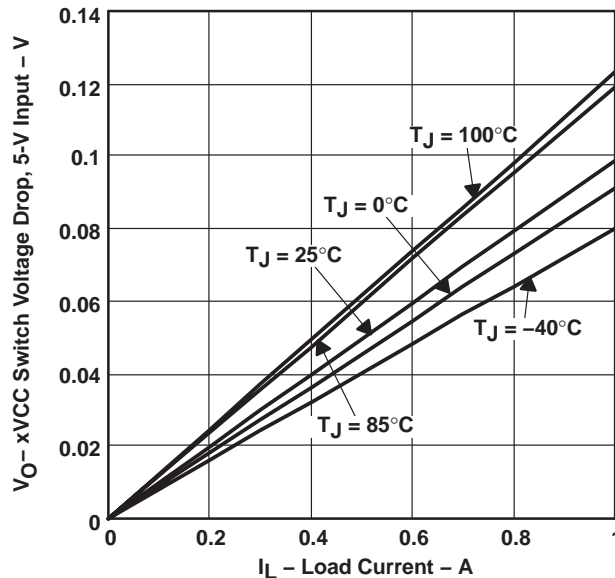


Figure 30

xVPP SWITCH VOLTAGE DROP, 12-V INPUT
 VS
 LOAD CURRENT

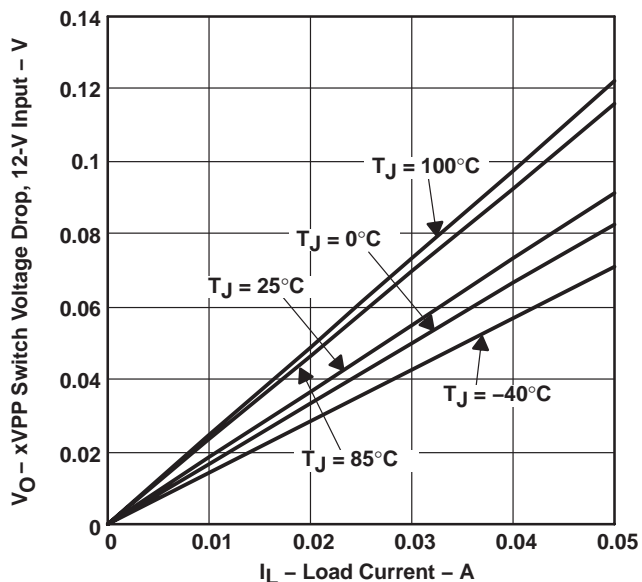


Figure 31

SHORT-CIRCUIT CURRENT LIMIT, 3.3 V TO xVCC
 VS
 JUNCTION TEMPERATURE

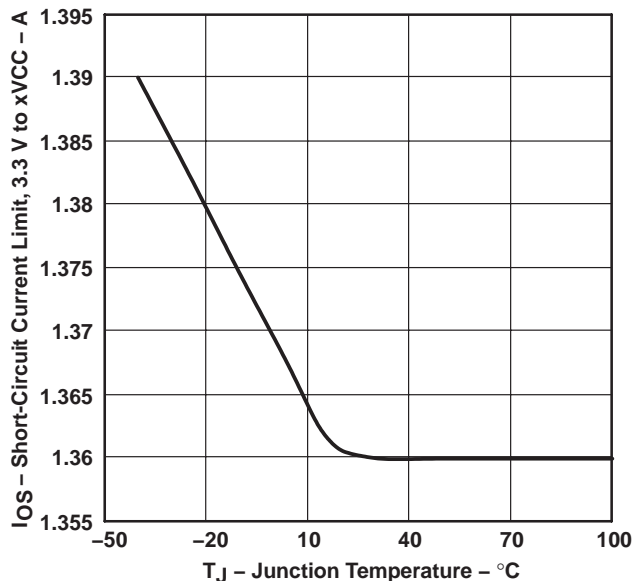


Figure 32

**TPS2223, TPS2224, TPS2226
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TYPICAL CHARACTERISTICS

**SHORT-CIRCUIT CURRENT LIMIT, 5 V TO xVCC
vs
JUNCTION TEMPERATURE**

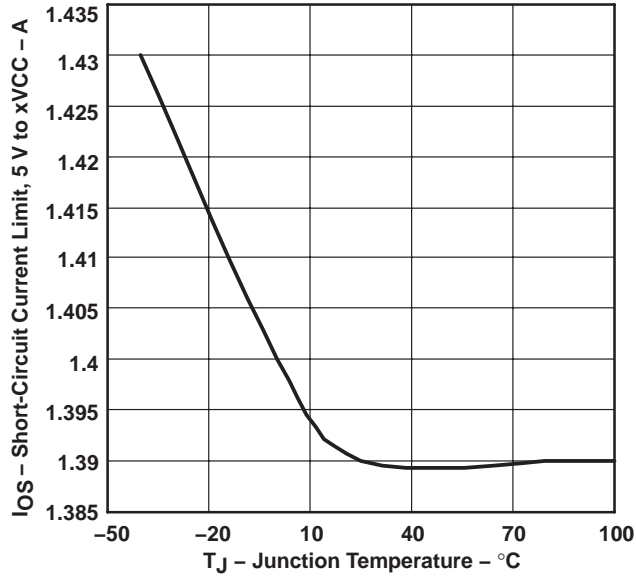


Figure 33

**SHORT-CIRCUIT CURRENT LIMIT, 12 V TO xVPP
vs
JUNCTION TEMPERATURE**

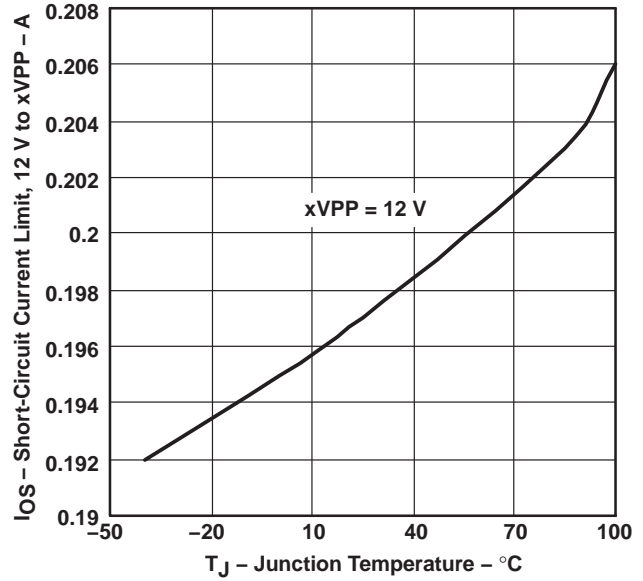


Figure 34

APPLICATION INFORMATION

overview

PC Cards were initially introduced as a means to add flash memory to portable computers. The idea of add-in cards quickly took hold, and modems, wireless LANs, global positioning satellite System (GPS), multimedia, and hard-disk versions were soon available. As the number of PC Card applications grew, the engineering community quickly recognized the need for a standard to ensure compatibility across platforms. To this end, the PCMCIA (Personal Computer Memory Card International Association) was established, comprising members from leading computer, software, PC Card, and semiconductor manufacturers. One key goal was to realize the *plug-and-play* concept, so that cards and hosts from different vendors would be transparently compatible.

PC Card power specification

System compatibility also means power compatibility. The most current set of specifications (PC Card Standard) set forth by the PCMCIA committee states that power is to be transferred between the host and the card through eight of the 68 terminals of the PC Card connector. This power interface consists of two V_{CC} , two V_{pp} , and four ground terminals. Multiple V_{CC} and ground terminals minimize connector-terminal and line resistance. The two V_{pp} terminals were originally specified as separate signals, but are normally tied together in the host to form a single node to minimize voltage losses. Card primary power is supplied through the V_{CC} terminals; flash-memory programming and erase voltage is supplied through the V_{pp} terminals. Cardbus cards of today typically do not use 12 V, which is now more of an optional requirement in the host.

designing for voltage regulation

The current PCMCIA specification for output voltage regulation, $V_{O(reg)}$, of the 5-V output is 5% (250 mV). In a typical PC power-system design, the power supply has an output-voltage regulation, $V_{PS(reg)}$, of 2% (100 mV). Also, a voltage drop from the power supply to the PC Card will result from resistive losses, V_{PCB} , in the PCB traces and the PCMCIA connector. A typical design would limit the total of these resistive losses to less than 1% (50 mV) of the output voltage. Therefore, the allowable voltage drop, V_{DS} , for the TPS2223, TPS2224 and TPS2226 would be the PCMCIA voltage regulation less the power supply regulation and less the PCB and connector resistive drops:

$$V_{DS} = V_{O(reg)} - V_{PS(reg)} - V_{PCB}$$

Typically, this would leave 100 mV for the allowable voltage drop across the 5-V switch. The specification for output voltage regulation of the 3.3-V output is 300 mV; therefore, using the same equation by deducting the voltage drop percentages (2%) for power-supply regulation and PCB resistive loss (1%), the allowable voltage drop for the 3.3-V switch is 200 mV. The voltage drop is the output current multiplied by the switch resistance of the TPS2223, TPS2224, and TPS2226. Therefore, the maximum output current, I_{Omax} , that can be delivered to the PC Card in regulation is the allowable voltage drop across the IC, divided by the output-switch resistance.

$$I_{Omax} = \frac{V_{DS}}{r_{DS(on)}}$$

The xVCC outputs have been designed to deliver the peak and average currents defined by the PC Card specification within regulation over the operating temperature range. The xVPP outputs of the TPS2226 have been designed to deliver 100 mA continuously.

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APPLICATION INFORMATION

overcurrent and overtemperature protection

PC Cards are inherently subject to damage that can result from mishandling. Host systems require protection against short-circuited cards that could lead to power-supply or PCB trace damage. Even extremely robust systems could undergo rapid battery discharge into a damaged PC Card, resulting in the rather sudden and unacceptable loss of system power. The reliability of fused systems is poor, in comparison, as blown fuses require troubleshooting and repair, usually by the manufacturer.

The TPS2223, TPS2224 and TPS2226 take a two-pronged approach to overcurrent protection, which is designed to activate if an output is shorted or when an overcurrent condition is present when switches are powered up. First, instead of fuses, sense FETs monitor each of the xVCC and xVPP power outputs. Unlike sense resistors or polyfuses, these FETs do not add to the series resistance of the switch; therefore voltage and power losses are reduced. Overcurrent sensing is applied to each output separately. Excessive current generates an error signal that limits the output current of only the affected output, preventing damage to the host. Each xVCC output overcurrent limits from 1 A to 2.2 A, typically around 1.6 A; the xVPP outputs limit from 100 mA to 250 mA, typically around 200 mA.

Second, when an overcurrent condition is detected, the TPS2223, TPS2224 and TPS2226 assert an active low \overline{OC} signal that can be monitored by the microprocessor or controller to initiate diagnostics and/or send the user a warning message. In the event that an overcurrent condition persists, causing the IC to exceed its maximum junction temperature, thermal-protection circuitry activates, shutting down all power outputs until the device cools to within a safe operating region, which is ensured by a thermal shutdown hysteresis. Thermal limiting prevents destruction of the IC from overheating beyond the package power-dissipation ratings.

During power up, the devices control the rise times of the xVCC and xVPP outputs and limit the inrush current into a large load capacitance, faulty card, or connector.

12-V supply not required

A few PC Card switches use the externally supplied 12 V to power gate drive and other chip functions, which requires that power be present at all times. The TPS2224 and TPS2226 offer considerable power savings by using an internal charge pump to generate the required higher gate drive voltages from the VA input (3.3 V). Therefore, the external 12-V supply can be disabled except when needed by the PC Card in the slot, thereby extending battery lifetime. A special feature in the 12-V circuitry actually helps to reduce the supply current demanded from the 3.3 V input. When 12 V is supplied and requested at the VPP output, a voltage selection circuit will draw the charge-pump drive current for the 12-V FETs from the 12-V input. This selection is automatic and effectively reduces demand fluctuations on the normal 3.3-V VCC rail. For proper operation of this feature, a minimum 3.3-V input capacitance of 4.7 μ F is recommended, and a minimum 12-V input ramp-up rate of 12 V/50 ms (240 V/s) is required. Additional power savings are realized by the TPS2226 during a software shutdown in which quiescent current drops to a maximum of 1 μ A.

voltage-transitioning requirement

PC Cards, like portables, are migrating from 5 V to 3.3 V to minimize power consumption, optimize board space, and increase logic speeds. The TPS2223, TPS2224 and TPS2226 meet all combinations of power delivery as currently defined in the PCMCIA standard. The latest protocol accommodates mixed 3.3-V/5-V systems by first powering the card with 5 V, then polling it to determine its 3.3-V compatibility. The PCMCIA specification requires that the capacitors on 3.3-V-compatible cards be discharged to below 0.8 V before applying 3.3-V power. This action ensures that sensitive 3.3-V circuitry is not subjected to any residual 5-V charge and functions as a power reset. PC Card specification requires that V_{CC} be discharged within 100 ms. PC Card resistance cannot be relied on to provide a discharge path for voltages stored on PC Card capacitance because of possible high-impedance isolation by power-management schemes. The devices include discharge transistors on all xVCC and xVPP outputs to meet the specification requirement.



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APPLICATION INFORMATION

shutdown mode

In the shutdown mode, which can be controlled by $\overline{\text{SHDN}}$ or bit D8 of the input serial DATA word, each of the xVCC and xVPP outputs is forced to a high-impedance state. In this mode, the chip quiescent current is reduced to 1 μA or less to conserve battery power.

power-supply considerations

These switches have multiple pins for each 3.3-V and 5-V power input and for the switched xVCC outputs. Any individual pin can conduct the rated input or output current. Unless all pins are connected in parallel, the series resistance is higher than that specified, resulting in increased voltage drops and power loss. It is recommended that all input and output power pins be paralleled for optimum operation.

To increase the noise immunity of the TPS2223, TPS2224 and TPS2226, the power-supply inputs should be bypassed with at least a 4.7 μF electrolytic or tantalum capacitor paralleled by a 0.047- μF to 0.1- μF ceramic capacitor. It is strongly recommended that the switched outputs be bypassed with a 0.1- μF (or larger) ceramic capacitor; doing so improves the immunity of the IC to electrostatic discharge (ESD). Care should be taken to minimize the inductance of PCB traces between the devices and the load. High switching currents can produce large negative voltage transients, which forward biases substrate diodes, resulting in unpredictable performance. Similarly, no pin should be taken below -0.3 V .

$\overline{\text{RESET}}$ input

To ensure that cards are in a known state after power brownouts or system initialization, the PC Cards should be reset at the same time as the host by applying low-impedance paths from xVCC and xVPP terminals to ground. A low-impedance output state allows discharging of residual voltage remaining on PC Card filter capacitance, permitting the system (host and PC Cards) to be powered up concurrently. The active low $\overline{\text{RESET}}$ input will close internal switches S1, S4, S7, and S11 with all other switches left open. The TPS2223, TPS2224 and TPS2226 remain in the low-impedance output state until the signal is deasserted and further data is clocked in and latched. The input serial data cannot be latched during reset mode. $\overline{\text{RESET}}$ is provided for direct compatibility with systems that use an active-low reset voltage supervisor. The $\overline{\text{RESET}}$ pin has an internal 150-k Ω pullup resistor.

calculating junction temperature

The switch resistance, $r_{\text{DS(on)}}$, is dependent on the junction temperature, T_J , of the die. The junction temperature is dependent on both $r_{\text{DS(on)}}$ and the current through the switch. To calculate T_J , first find $r_{\text{DS(on)}}$ from Figures 26 through 28, using an initial temperature estimate about 30°C above ambient. Then calculate the power dissipation for each switch, using the formula:

$$P_D = r_{\text{DS(on)}} \times I^2$$

Next, sum the power dissipation of all switches and calculate the junction temperature:

$$T_J = \left(\sum P_D \times R_{\theta\text{JA}} \right) + T_A$$

Where:

$R_{\theta\text{JA}}$ is the inverse of the derating factor given in the dissipation rating table.

Compare the calculated junction temperature with the initial temperature estimate. If the temperatures are not within a few degrees of each other, recalculate using the calculated temperature as the initial estimate.

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logic inputs and outputs

The serial interface consists of DATA, CLOCK, and LATCH leads. The data is clocked in on the positive edge of the clock (see Figure 2). The 11-bit (D0–D10) serial data word is loaded during the positive edge of the latch signal. The positive edge of the latch signal should occur before the next positive edge of the clock occurs.

The serial interface of the device is compatible with serial-interface PCMCIA controllers.

An overcurrent output (\overline{OC}) is provided to indicate an overcurrent or overtemperature condition in any of the xVCC and xVPP outputs as previously discussed.

TPS2223, TPS2224 and TPS2226 control logic

xVPP

| D8 (\overline{SHDN}) | AVPP CONTROL SIGNALS | | | OUTPUT V_AVPP | BVPP CONTROL SIGNALS | | | | OUTPUT V_BVPP |
|--------------------------|----------------------|----|----|-------------------|--------------------------|----|----|-----|-------------------|
| | D0 | D1 | D9 | | D8 (\overline{SHDN}) | D4 | D5 | D10 | |
| 1 | 0 | 0 | X | 0 V | 1 | 0 | 0 | X | 0 V |
| 1 | 0 | 1 | 0 | 3.3 V | 1 | 0 | 1 | 0 | 3.3 V |
| 1 | 0 | 1 | 1 | 5 V | 1 | 0 | 1 | 1 | 5 V |
| 1 | 1 | 0 | X | 12 V [†] | 1 | 1 | 0 | X | 12 V [†] |
| 1 | 1 | 1 | X | Hi-Z | 1 | 1 | 1 | X | Hi-Z |
| 0 | X | X | X | Hi-Z | 0 | X | X | X | Hi-Z |

[†] The output V_xVPP is Hi-Z for TPS2223.

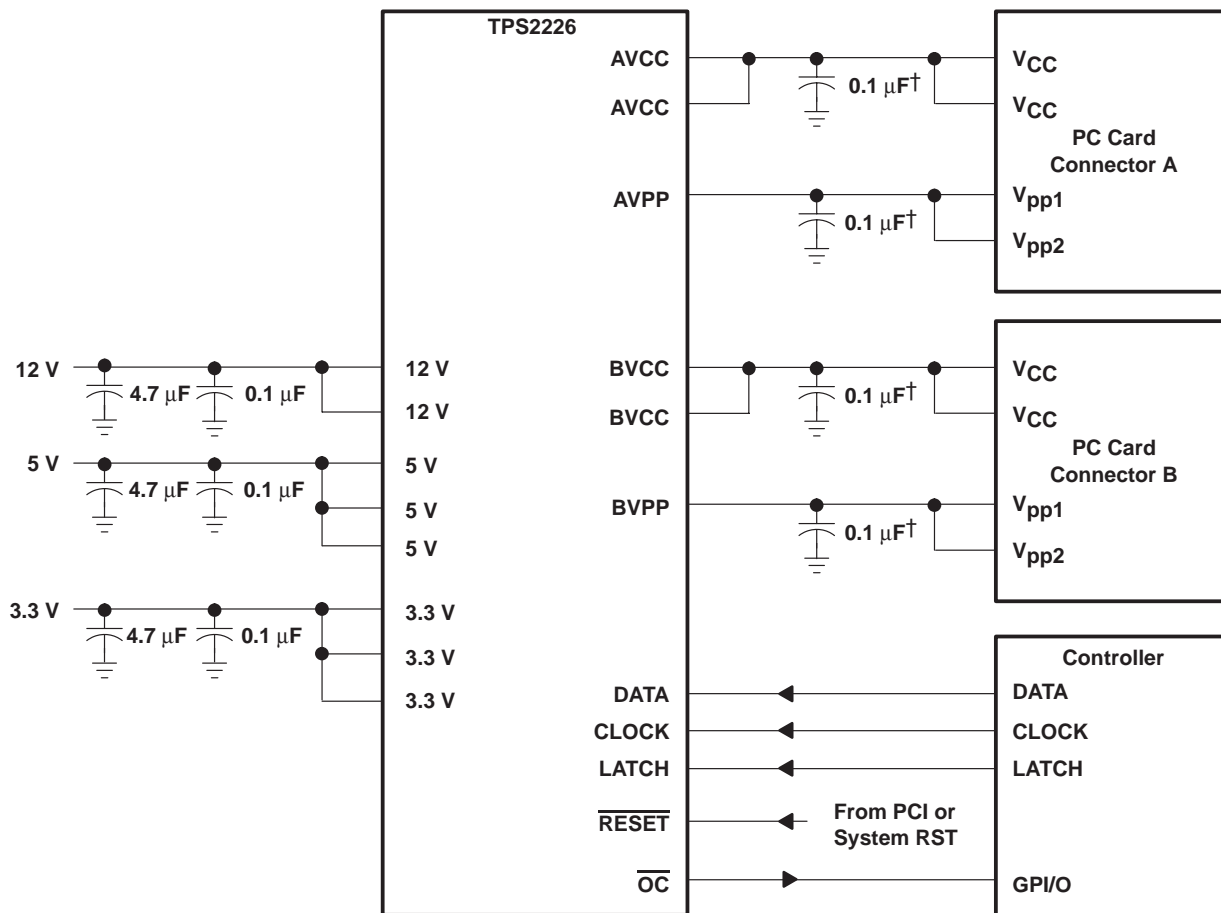
xVCC

| D8 (\overline{SHDN}) | AVCC CONTROL SIGNALS | | OUTPUT V_AVCC | BVCC CONTROL SIGNALS | | | OUTPUT V_BVCC |
|--------------------------|----------------------|----|------------------|--------------------------|----|----|------------------|
| | D3 | D2 | | D8 (\overline{SHDN}) | D6 | D7 | |
| 1 | 0 | 0 | 0 V | 1 | 0 | 0 | 0 V |
| 1 | 0 | 1 | 3.3 V | 1 | 0 | 1 | 3.3 V |
| 1 | 1 | 0 | 5 V | 1 | 1 | 0 | 5 V |
| 1 | 1 | 1 | 0 V | 1 | 1 | 1 | 0 V |
| 0 | X | X | Hi-Z | 0 | X | X | Hi-Z |

APPLICATION INFORMATION

ESD protections (see Figure 35)

All inputs and outputs of these devices incorporate ESD-protection circuitry designed to withstand a 2-kV human-body-model discharge as defined in MIL-STD-883C, Method 3015. The xVCC and xVPP outputs can be exposed to potentially higher discharges from the external environment through the PC Card connector. Bypassing the outputs with 0.1- μ F capacitors protects the devices from discharges up to 10 kV.



† Maximum recommended output capacitance for xVCC is 220 μ F including card capacitance, and for xVPP is 10 μ F, without \overline{OC} glitch when switches are powered on.

Figure 35. Detailed Interconnections and Capacitor Recommendations

TPS2223, TPS2224, TPS2226 DUAL-SLOT CARDBUS POWER-INTERFACE SWITCHES FOR SERIAL PCMCIA CONTROLLERS

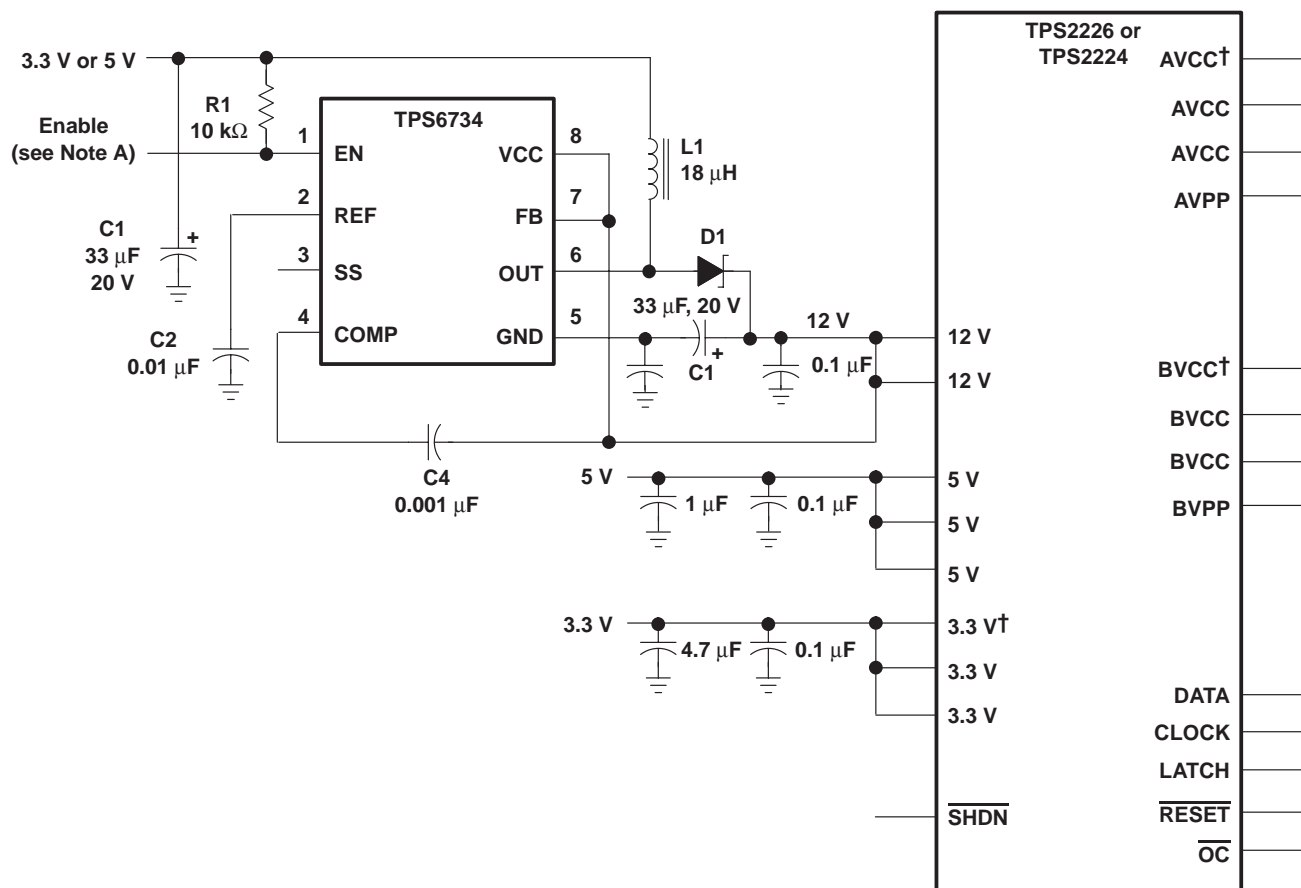
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APPLICATION INFORMATION

12-V flash memory supply

The TPS6734 is a fixed 12-V output boost converter capable of delivering 120 mA from inputs as low as 2.7 V. The device is pin-for-pin compatible with the MAX734 regulator and offers the following advantages: lower supply current, wider operating input-voltage range, and higher output currents. As shown in Figure 36, the only external components required are: an inductor, a Schottky rectifier, an output filter capacitor, an input filter capacitor, and a small capacitor for loop compensation. The entire converter occupies less than 0.7 in² of PCB space when implemented with surface-mount components. An enable input is provided to shut the converter down and reduce the supply current to 3 μ A when 12 V is not needed.

The TPS6734 is a 170-kHz current-mode PWM (pulse-width modulation) controller with an n-channel MOSFET power switch. Gate drive for the switch is derived from the 12-V output after start-up to minimize the die area needed to realize the 0.7- Ω MOSFET and improve efficiency at input voltages below 5 V. Soft start is accomplished with the addition of one small capacitor. A 1.22-V reference, pin 2 of TPS6734, is brought out for external use. For additional information, see the TPS6734 data sheet (SLVS127).



† Not on TPS2224

NOTE A: The enable terminal can be tied to a general-purpose I/O terminal on the PCMCIA controller or tied high.

Figure 36. TPS2224 and TPS2226 with TPS6734 12-V, 120-mA Supply

PACKAGING INFORMATION

| Orderable Device | Status (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan (2) | Lead finish/ Ball material (6) | MSL Peak Temp (3) | Op Temp (°C) | Device Marking (4/5) | Samples |
|------------------|---------------|--------------|-----------------|------|-------------|-----------------|--------------------------------------|----------------------|--------------|-------------------------|-------------------------|
| TPS2223DB | ACTIVE | SSOP | DB | 24 | 60 | RoHS & Green | NIPDAU | Level-1-260C-UNLIM | -40 to 85 | TPS2223 | Samples |
| TPS2224DB | ACTIVE | SSOP | DB | 24 | 60 | RoHS & Green | NIPDAU | Level-1-260C-UNLIM | -40 to 85 | TPS2224 | Samples |
| TPS2224PWP | ACTIVE | HTSSOP | PWP | 24 | 60 | RoHS & Green | NIPDAU | Level-2-260C-1 YEAR | -40 to 85 | TPS2224 | Samples |
| TPS2226DB | ACTIVE | SSOP | DB | 30 | 50 | RoHS & Green | NIPDAU | Level-1-260C-UNLIM | -40 to 85 | TPS2226 | Samples |
| TPS2226DBR | ACTIVE | SSOP | DB | 30 | 2000 | RoHS & Green | NIPDAU | Level-1-260C-UNLIM | -40 to 85 | TPS2226 | Samples |

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| TPS2226DBR | SSOP | DB | 30 | 2000 | 330.0 | 16.4 | 8.2 | 10.5 | 2.5 | 12.0 | 16.0 | Q1 |

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|------------|--------------|-----------------|------|------|-------------|------------|-------------|
| TPS2226DBR | SSOP | DB | 30 | 2000 | 356.0 | 356.0 | 35.0 |

TUBE


*All dimensions are nominal

| Device | Package Name | Package Type | Pins | SPQ | L (mm) | W (mm) | T (μm) | B (mm) |
|------------|--------------|--------------|------|-----|--------|--------|--------|--------|
| TPS2223DB | DB | SSOP | 24 | 60 | 530 | 10.5 | 4000 | 4.1 |
| TPS2224DB | DB | SSOP | 24 | 60 | 530 | 10.5 | 4000 | 4.1 |
| TPS2224PWP | PWP | HTSSOP | 24 | 60 | 530 | 10.2 | 3600 | 3.5 |
| TPS2226DB | DB | SSOP | 30 | 50 | 530 | 10.5 | 4000 | 4.1 |

GENERIC PACKAGE VIEW

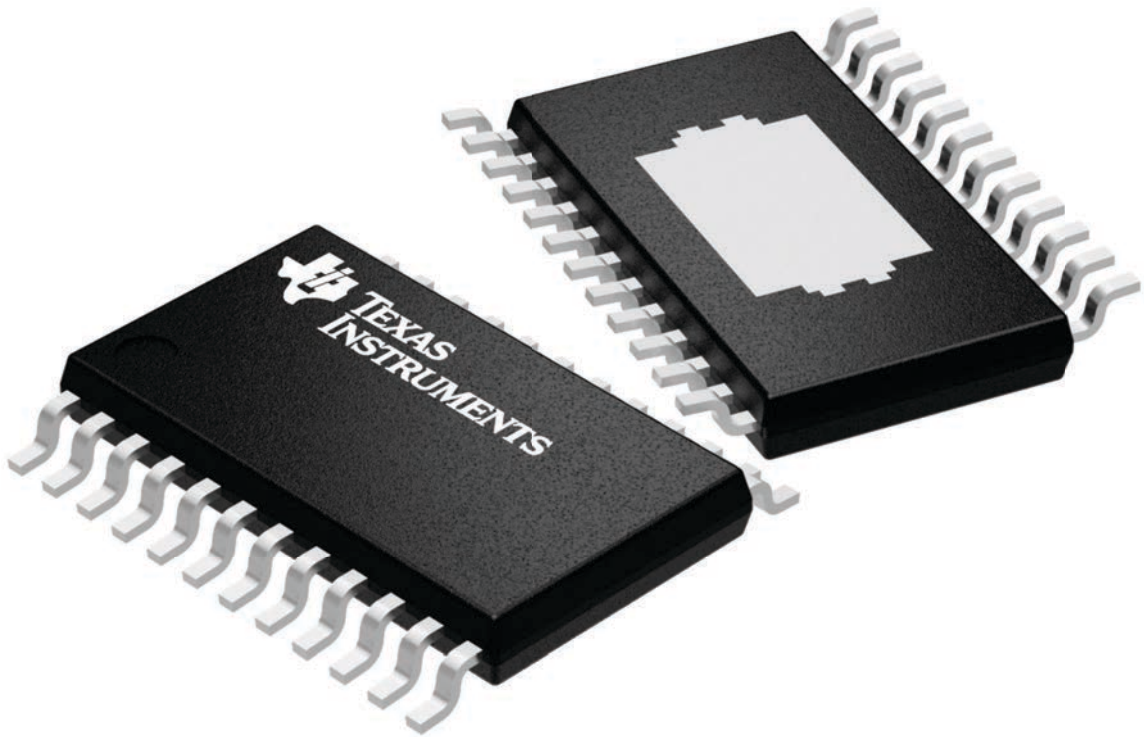
PWP 24

PowerPAD™ TSSOP - 1.2 mm max height

4.4 x 7.6, 0.65 mm pitch

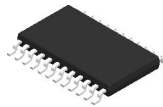
PLASTIC SMALL OUTLINE

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4224742/B

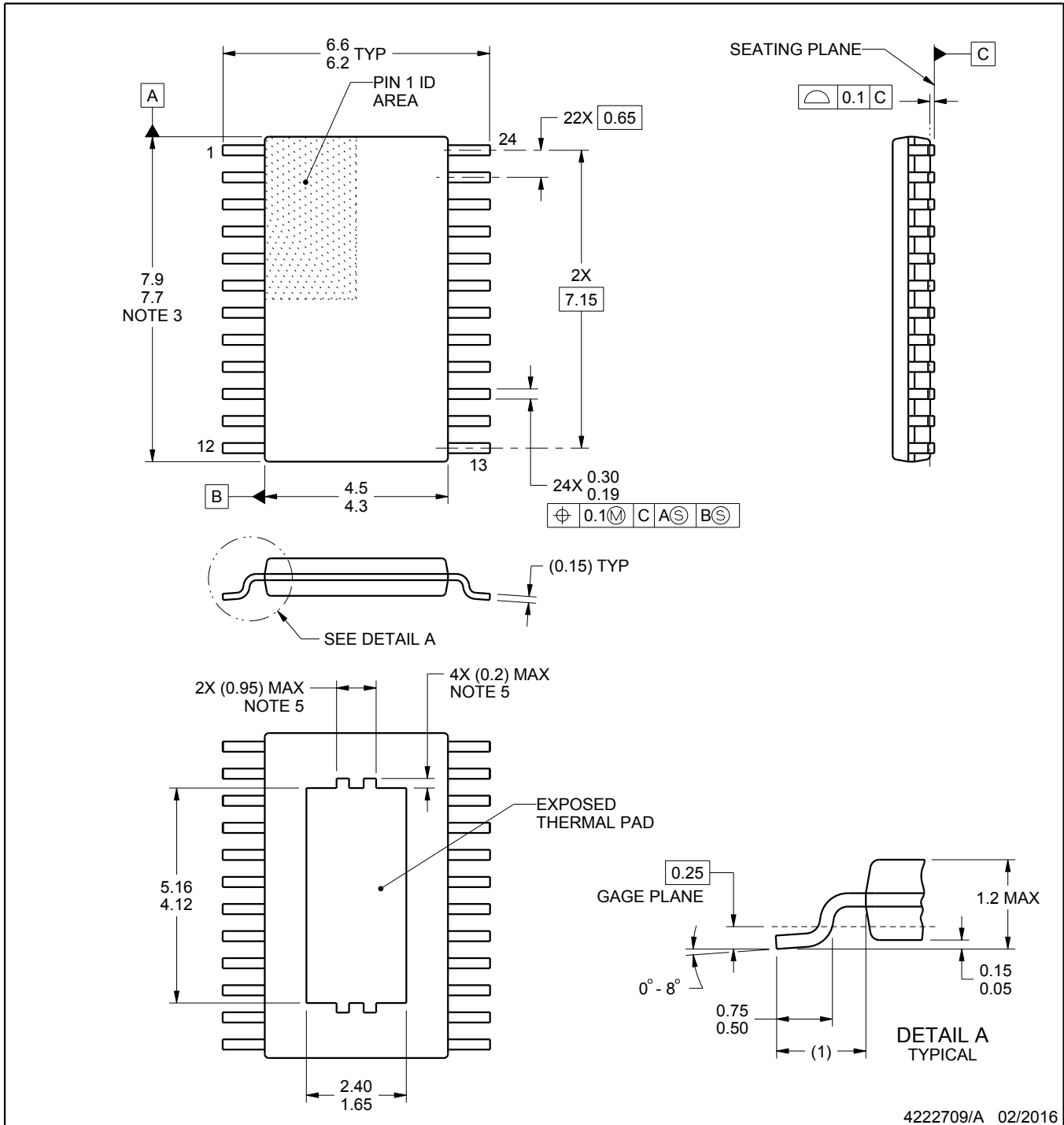
PWP0024B



PACKAGE OUTLINE

PowerPAD™ TSSOP - 1.2 mm max height

PLASTIC SMALL OUTLINE



NOTES:

PowerPAD is a trademark of Texas Instruments.

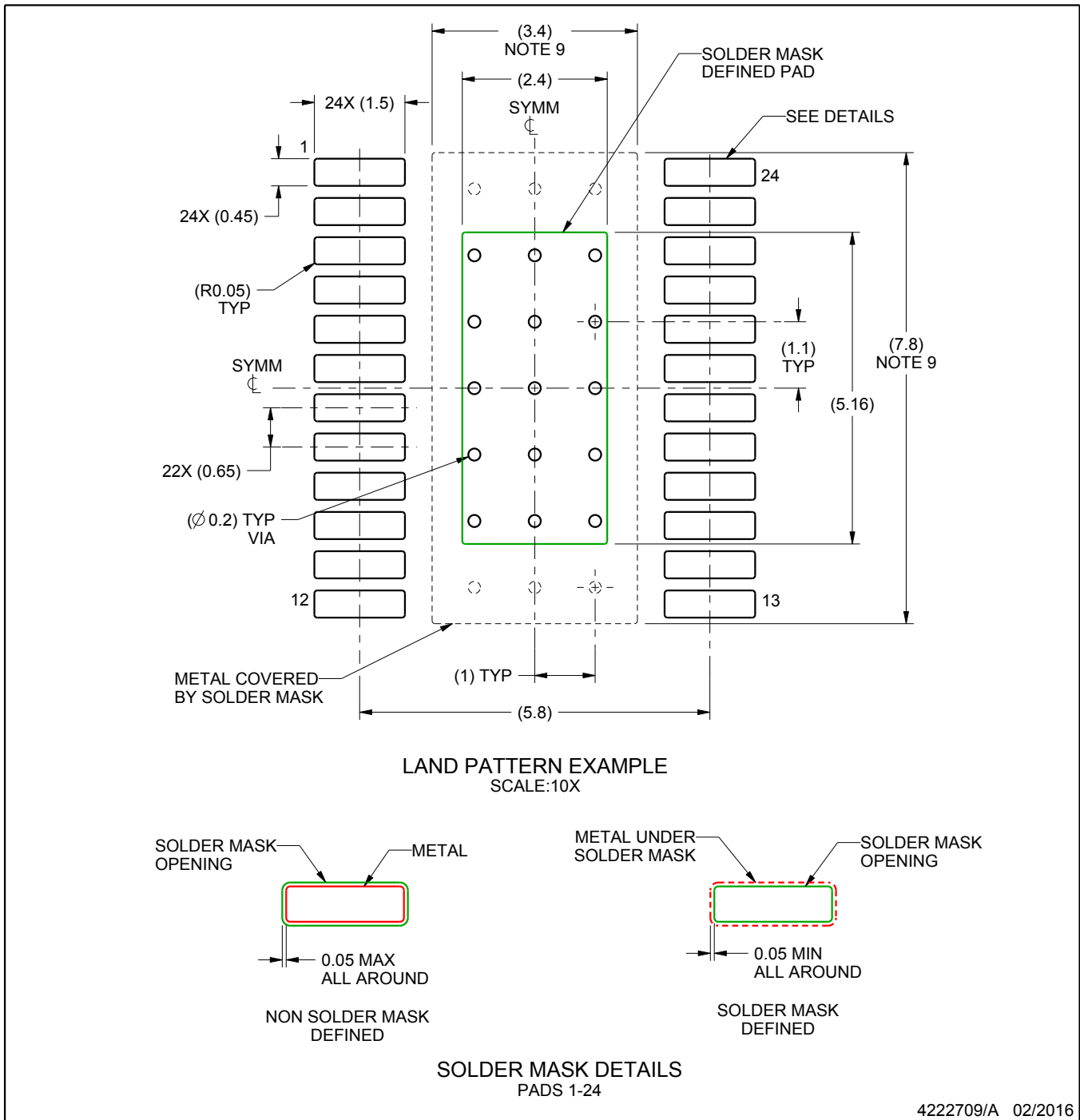
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. Reference JEDEC registration MO-153.
5. Features may not be present and may vary.

EXAMPLE BOARD LAYOUT

PWP0024B

PowerPAD™ TSSOP - 1.2 mm max height

PLASTIC SMALL OUTLINE



NOTES: (continued)

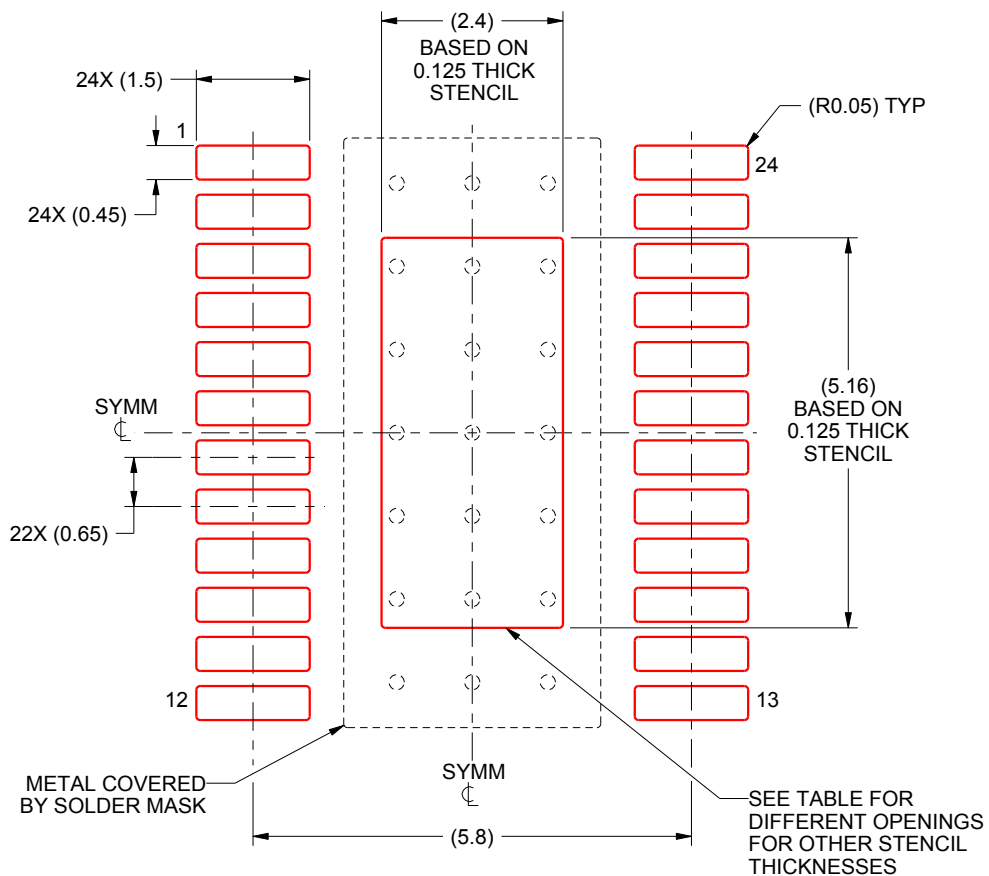
6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature numbers SLMA002 (www.ti.com/lit/slma002) and SLMA004 (www.ti.com/lit/slma004).
9. Size of metal pad may vary due to creepage requirement.

EXAMPLE STENCIL DESIGN

PWP0024B

PowerPAD™ TSSOP - 1.2 mm max height

PLASTIC SMALL OUTLINE



SOLDER PASTE EXAMPLE
 EXPOSED PAD
 100% PRINTED SOLDER COVERAGE BY AREA
 SCALE:10X

| STENCIL THICKNESS | SOLDER STENCIL OPENING |
|-------------------|------------------------|
| 0.1 | 2.68 X 5.77 |
| 0.125 | 2.4 X 5.16 (SHOWN) |
| 0.15 | 2.19 X 4.71 |
| 0.175 | 2.03 X 4.36 |

4222709/A 02/2016

NOTES: (continued)

- 10. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 11. Board assembly site may have different recommendations for stencil design.

DB (R-PDSO-G**)

PLASTIC SMALL-OUTLINE

28 PINS SHOWN



- NOTES: A. All linear dimensions are in millimeters.
 B. This drawing is subject to change without notice.
 C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.
 D. Falls within JEDEC MO-150

DB (R-PDSO-G**)

PLASTIC SMALL-OUTLINE

28 PINS SHOWN



- NOTES: A. All linear dimensions are in millimeters.
 B. This drawing is subject to change without notice.
 C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.
 D. Falls within JEDEC MO-150

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