

54A, 600V, Rugged UFS Series N-Channel IGBT with Anti-Parallel Ultrafast Diode

January 1997

Features

- 54A, 600V, $T_C = 25^\circ\text{C}$
- 600V Switching SOA Capability
- Typical Fall Time at $T_J = 150^\circ\text{C}$ 200ns
- Short Circuit Rating at $T_J = 150^\circ\text{C}$ 10 μs
- Low Conduction Loss

Ordering Information

PART NUMBER	PACKAGE	BRAND
HGTG27N60C3DR	TO-247	27N60C3DR

NOTE: When ordering, use the entire part number.

Description

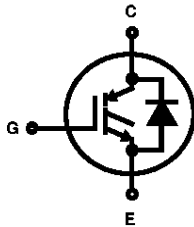
This IGBT was designed for optimum performance in the demanding world of motor control operation as well as other high voltage switching applications. This device demonstrates RUGGED performance capability when subjected to harsh SHORT CIRCUIT WITHSTAND TIME (SCWT) conditions. The parts have ULTRAFast (UFS) switching speed while the on-state conduction losses have been kept at a low level.

The electrical specifications include typical Turn-On and Turn-Off dv/dt ratings. These ratings and the turn-on ratings include the effect of the diode in the test circuit (Figure 18). The data was obtained with the diode at the same T_J as the IGBT under test. The IGBT was formerly developmental type 49048. The diode used in anti-parallel with the IGBT was formerly developmental type 49214.

The IGBT diode combination was formerly developmental type TA49013.

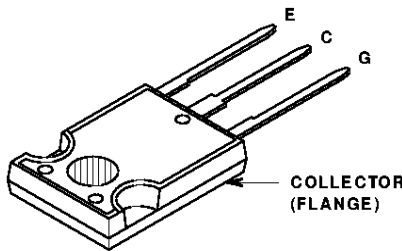
Symbol

N-CHANNEL ENHANCEMENT MODE



Package

JEDEC STYLE TO-247



HARRIS SEMICONDUCTOR IGBT PRODUCT IS COVERED BY ONE OR MORE OF THE FOLLOWING U.S. PATENTS:

4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,567,641
4,587,713	4,598,461	4,605,948	4,618,872	4,620,211	4,631,564	4,639,754	4,639,762
4,641,162	4,644,637	4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690
4,794,432	4,801,986	4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606
4,860,080	4,883,767	4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951
4,969,027							

HGTG27N60C3DR

Absolute Maximum Ratings $T_C = 25^\circ$ Unless Otherwise Specified

	HGTG27N60C3DR	UNITS
Collector-Emitter Voltage	600	V
Collector Current Continuous		
At $T_C = 25^\circ\text{C}$	54	A
At $T_C = 110^\circ\text{C}$	27	A
Collector Current Pulsed (Note 1)	108	A
Gate-Emitter Voltage Continuous	± 20	V
Gate-Emitter Voltage Pulsed	± 30	V
Switching Safe Operating Area at $T_J = 150^\circ\text{C}$, (Figure 12)	108A at 600V	
Power Dissipation Total at $T_C = 25^\circ\text{C}$	208	W
Power Dissipation Derating $T_C > 25^\circ\text{C}$	1.67	W/ $^\circ\text{C}$
Reverse Voltage Avalanche Energy	100	mJ
Operating and Storage Junction Temperature Range	-40 to 150	$^\circ\text{C}$
Maximum Lead Temperature for Soldering	260	$^\circ\text{C}$
Short Circuit Withstand Time (Note 2) at $V_{GE} = 15\text{V}$	10	μs

NOTES:

- Pulse width limited by maximum junction temperature.
- $V_{CE(PK)} = 440\text{V}$, $T_J = 150^\circ\text{C}$, $R_{GE} = 3\Omega$.

Electrical Specifications $T_C = 25^\circ\text{C}$, Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNITS	
			MIN	TYP	MAX		
Collector-Emitter Breakdown Voltage	BV_{CES}	$I_C = 250\mu\text{A}$, $V_{GE} = 0\text{V}$	600	-	-	V	
Emitter-Collector Breakdown Voltage	BV_{ECS}	$I_C = 10\text{mA}$, $V_{GE} = 0\text{V}$	15	-	-	V	
Collector-Emitter Leakage Current	I_{CES}	$V_{CE} = BV_{CES}$	$T_C = 25^\circ\text{C}$	-	-	250	μA
			$T_C = 150^\circ\text{C}$	-	-	3.0	mA
Collector-Emitter Saturation Voltage	$V_{CE(SAT)}$	$I_C = I_{C110}$, $V_{GE} = 15\text{V}$	$T_C = 25^\circ\text{C}$	-	1.8	2.2	V
			$T_C = 150^\circ\text{C}$	-	2.1	2.5	V
Gate-Emitter Threshold Voltage	$V_{GE(TH)}$	$I_C = 250\mu\text{A}$, $V_{CE} = V_{GE}$	3.5	5.7	7.5	V	
Gate-Emitter Leakage Current	I_{GES}	$V_{GE} = \pm 20\text{V}$	-	-	± 100	nA	
Switching SOA (See Figure 12)	SSOA	$T_J = 150^\circ\text{C}$, $R_G = 3\Omega$, $L = 50\mu\text{H}$ $V_{GE} = 15\text{V}$, $V_{CE(PK)} = 600\text{V}$	108	-	-	A	
Gate-Emitter Plateau Voltage	V_{GEP}	$I_C = I_{C110}$, $V_{CE} = 0.5 BV_{CES}$	-	9.0	-	V	
On-State Gate Charge	$Q_{G(ON)}$	$I_C = I_{C110}$, $V_{CE} = 0.5 BV_{ES}$	$V_{GE} = 15\text{V}$	-	156	203	nC
			$V_{GE} = 20\text{V}$	-	212	277	nC
Current Turn-On Delay Time	$t_{D(ON)I}$	$T_J = 150^\circ\text{C}$ $I_{CE} = I_{C110}$ $V_{CE(PK)} = 0.8 BV_{CES}$ $V_{GE} = 15\text{V}$ $R_G = 3\Omega$ $L = 1\text{mH}$ Diode used in test circuit RURP3060 at 150°C	-	38	-	ns	
Current Rise Time	t_{RI}		-	30	-	ns	
Current Turn-Off Delay Time	$t_{D(OFF)I}$		-	250	500	ns	
Current Fall Time	t_{FI}		-	200	400	ns	
Turn-Off Voltage dv/dt (Note 3)	dV_{CE}/dt		-	2	-	V/ns	
Turn-On Voltage dv/dt (Note 3)	dV_{CE}/dt		-	7	-	V/ns	
Turn-On Energy (Note 4)	E_{ON}		-	2.3	-	mJ	
Turn-Off Energy (Note 5)	E_{OFF}		-	2.0	-	mJ	
Diode Forward Voltage	V_{EC}	$I_{EC} = 27\text{A}$	-	-	1.5	V	

HGTG27N60C3DR

Electrical Specifications $T_C = 25^\circ\text{C}$, Unless Otherwise Specified (Continued)

PARAMETER	SYMBOL	TEST CONDITIONS	LIMITS			UNITS
			MIN	TYP	MAX	
Diode Reverse Recovery Time	t_{rr}	$I_{EC} = 1\text{A}, dI_{EC}/dt = 200\text{A}/\mu\text{s}$	-	-	55	ns
		$I_{EC} = 27\text{A}, dI_{EC}/dt = 200\text{A}/\mu\text{s}$	-	-	60	ns
Thermal Resistance	$R_{\theta JC}$	IGBT	-	-	0.6	$^\circ\text{C}/\text{W}$
		Diode	-	-	1.25	$^\circ\text{C}/\text{W}$

NOTES:

- dV_{CE}/dt depends on the diode used and the temperature of the diode.
- Turn-On Energy Loss (E_{ON}) includes losses due to the diode recovery and is defined as the integral of the instantaneous power loss starting at the leading edge of the input pulse and ending at the point where the collector voltage equals $V_{CE(SAT)}$. This value of E_{ON} was obtained with a RURP3060 diode at $T_J = 150^\circ\text{C}$. A different diode or temperature will result in a different E_{ON} . For example, with diode at $T_J = 25^\circ\text{C}$ E_{ON} is about one half the value at 150°C .
- Turn-Off Energy Loss (E_{OFF}) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ($I_{CE} = 0\text{A}$). All devices were tested per JEDEC standard No. 24-1 Method for measurement of power device turn-off switching loss. This test method produces the true total Turn-Off Energy Loss.

Typical Performance Curves

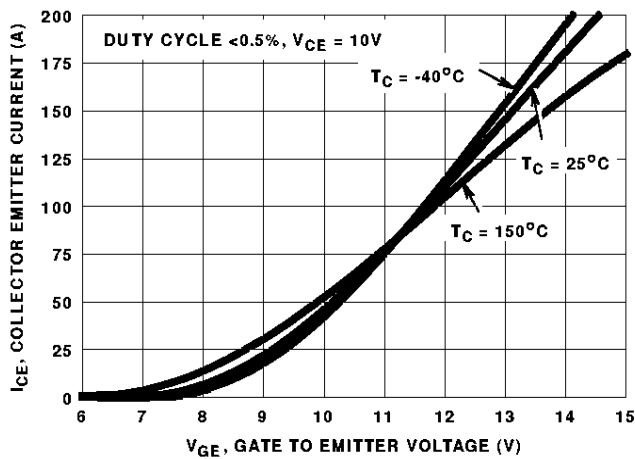


FIGURE 1. TRANSFER CHARACTERISTICS

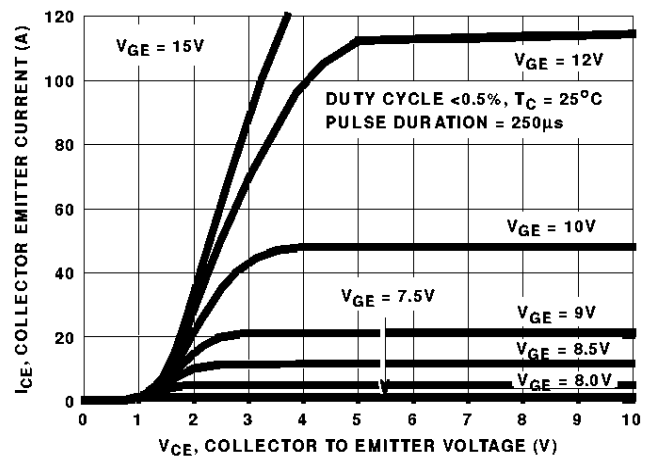


FIGURE 2. SATURATION CHARACTERISTICS

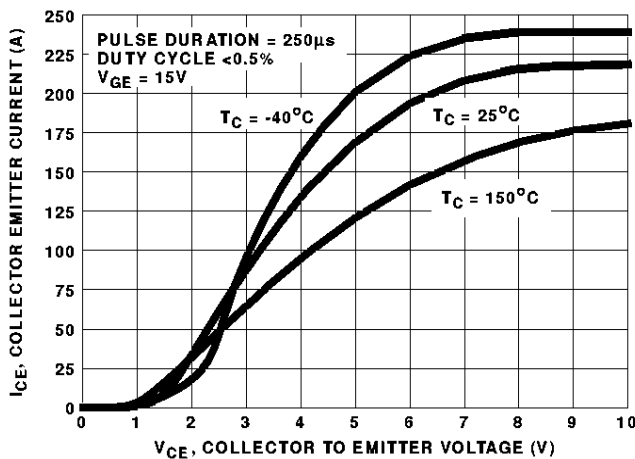


FIGURE 3. COLLECTOR EMITTER ON STATE VOLTAGE

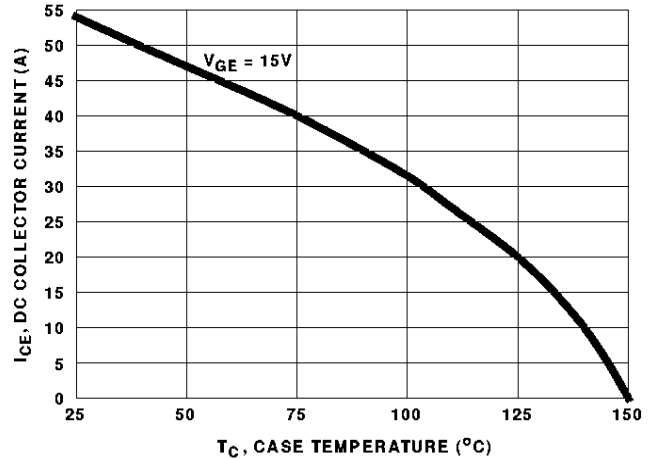


FIGURE 4. DC COLLECTOR CURRENT AS A FUNCTION OF CASE TEMPERATURE

Typical Performance Curves (Continued)

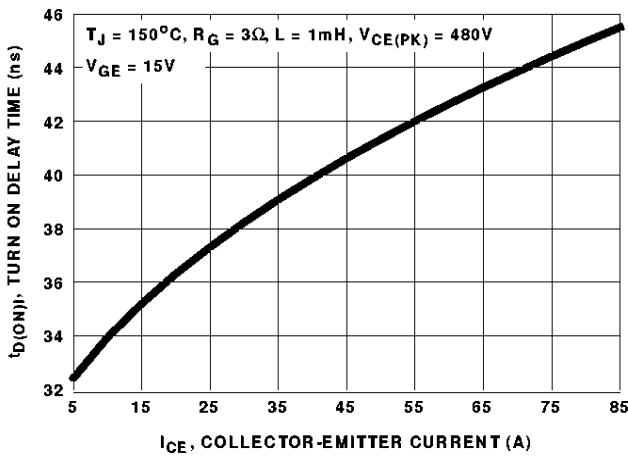


FIGURE 5. TURN ON DELAY TIME AS A FUNCTION OF COLLECTOR EMITTER CURRENT

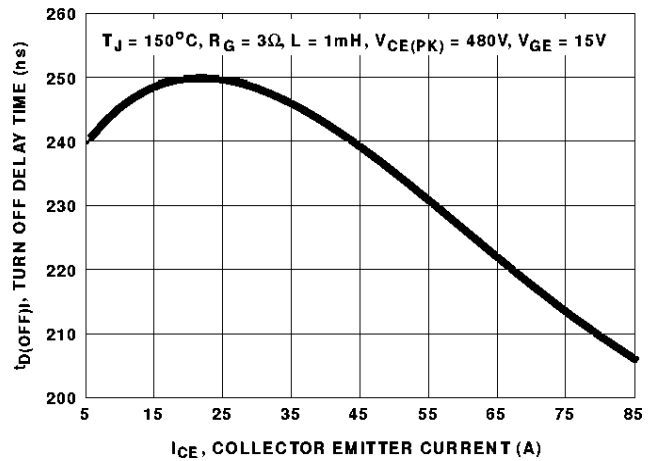


FIGURE 6. TURN OFF DELAY TIME AS A FUNCTION OF COLLECTOR EMITTER CURRENT

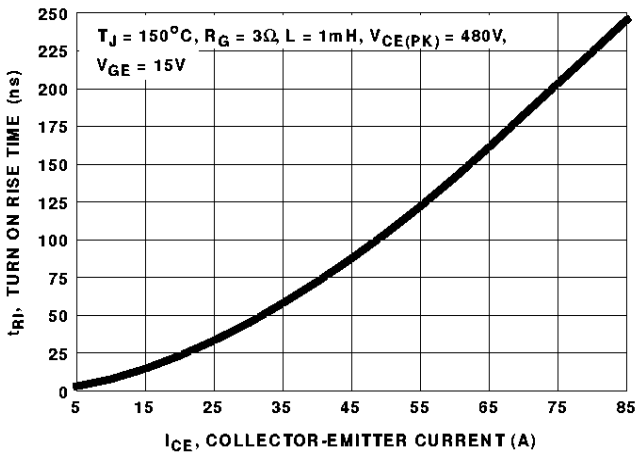


FIGURE 7. TURN ON RISE TIME AS A FUNCTION OF COLLECTOR EMITTER CURRENT

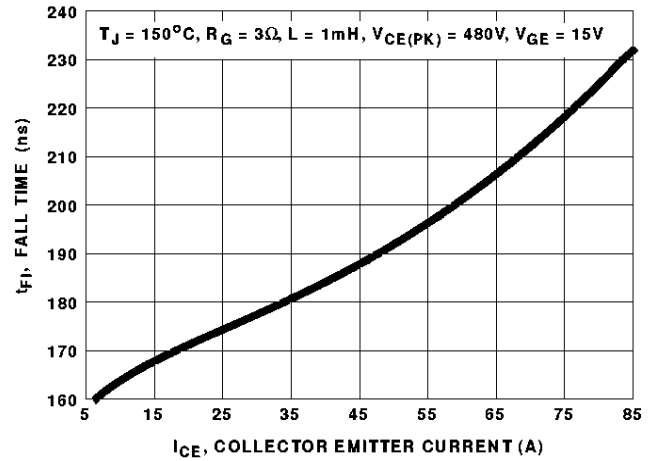


FIGURE 8. TURN OFF FALL TIME AS A FUNCTION OF COLLECTOR EMITTER CURRENT

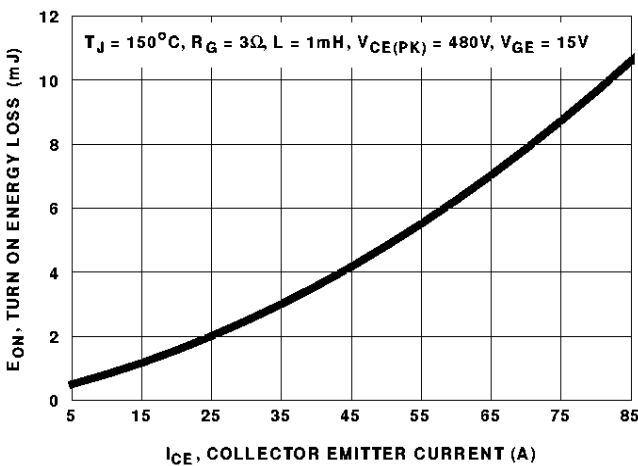


FIGURE 9. TURN ON ENERGY LOSS AS A FUNCTION OF COLLECTOR EMITTER CURRENT

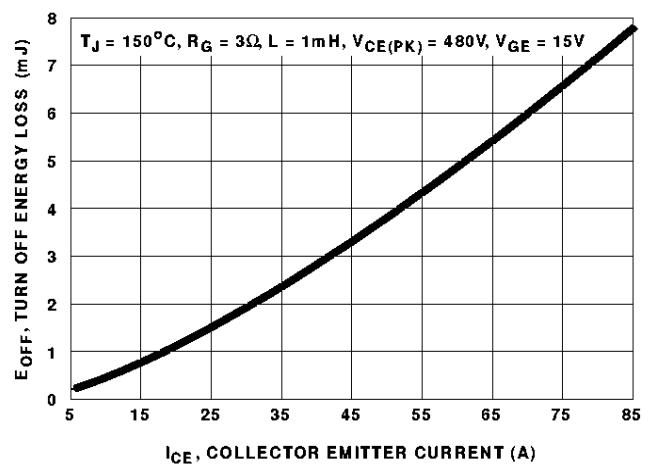


FIGURE 10. TURN OFF ENERGY LOSS AS A FUNCTION OF COLLECTOR EMITTER CURRENT

Typical Performance Curves (Continued)

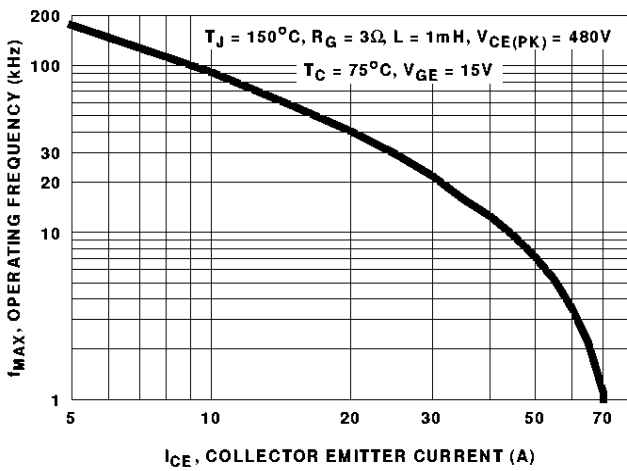


FIGURE 11. OPERATING FREQUENCY AS A FUNCTION OF COLLECTOR EMITTER CURRENT

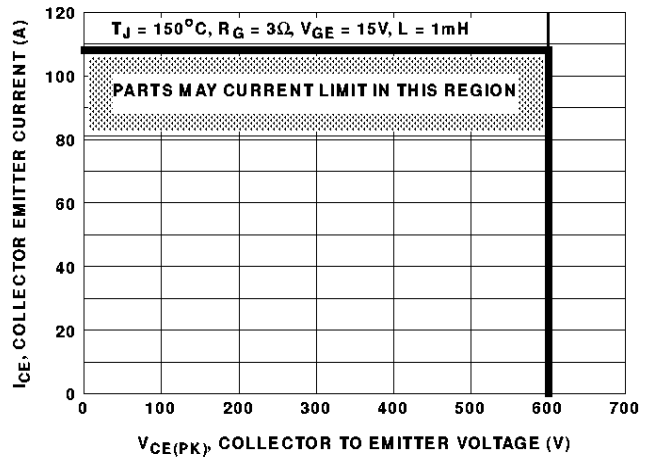


FIGURE 12. SWITCHING SAFE OPERATING AREA

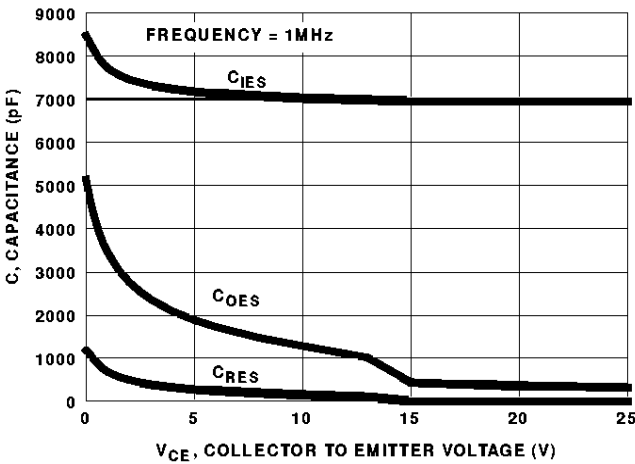


FIGURE 13. CAPACITANCE AS A FUNCTION OF COLLECTOR-EMITTER VOLTAGE

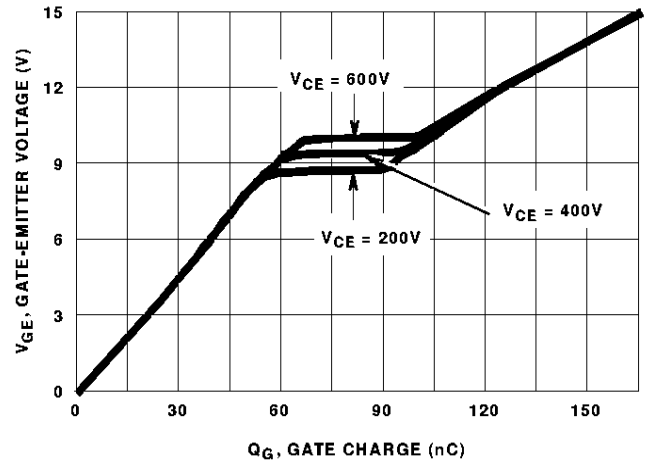


FIGURE 14. GATE CHARGE WAVEFORMS

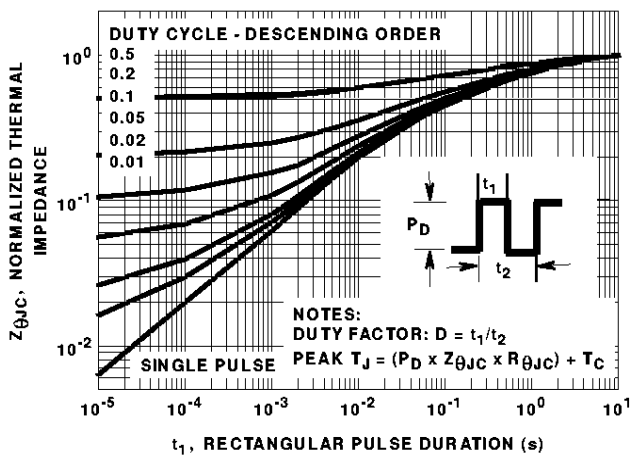


FIGURE 15. IGBT NORMALIZED TRANSIENT THERMAL IMPEDANCE, JUNCTION TO CASE

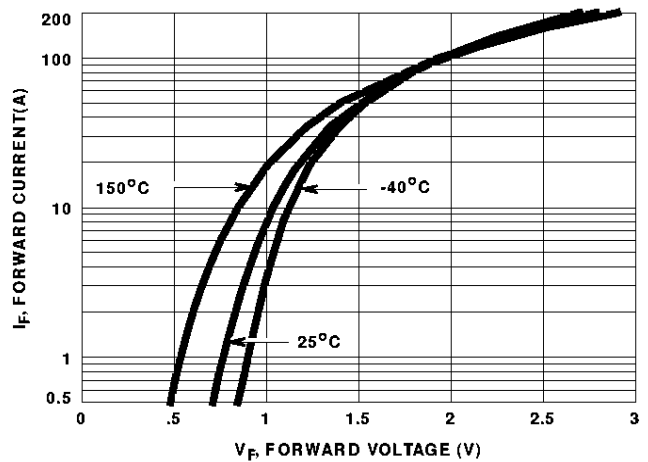


FIGURE 16. DIODE FORWARD CURRENT AS A FUNCTION OF FORWARD VOLTAGE DROP

Typical Performance Curves (Continued)

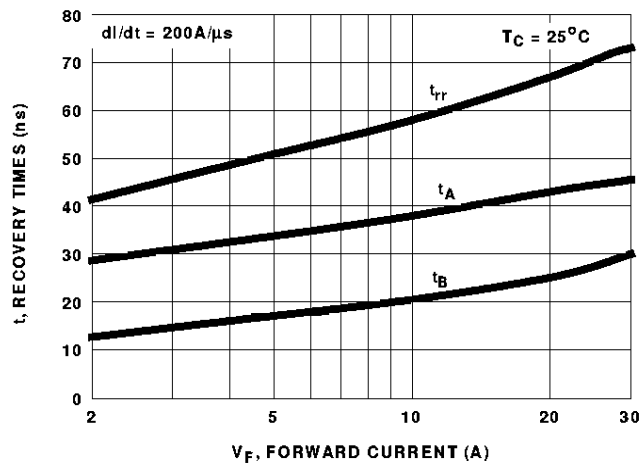


FIGURE 17. RECOVERY TIMES AS A FUNCTION OF FORWARD CURRENT

Test Circuit and Waveform

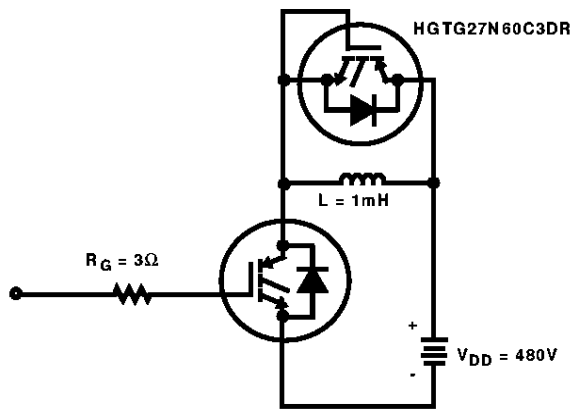


FIGURE 18. INDUCTIVE SWITCHING TEST CIRCUIT

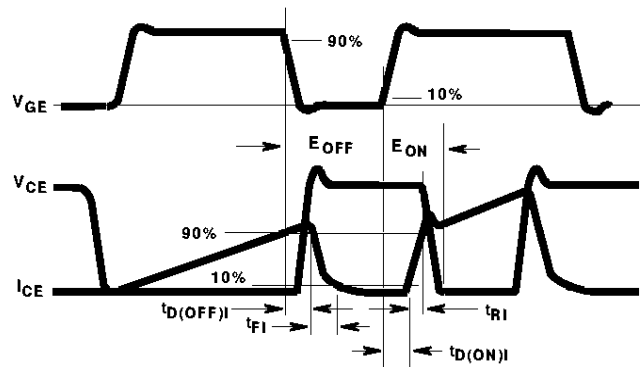


FIGURE 19. SWITCHING TEST WAVEFORMS

Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
3. Tips of soldering irons should be grounded.
4. Devices should never be inserted into or removed from circuits with power on.
5. **Gate Voltage Rating** - Never exceed the gate-voltage rating of V_{GEM} . Exceeding the rated V_{GE} can result in permanent damage to the oxide layer in the gate region.
6. **Gate Termination** - The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
7. **Gate Protection** - These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

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Operating Frequency Information

Operating frequency information for a typical device (Figure 11) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (I_{CE}) plots are possible using the information shown for a typical unit in Figures 3, 5, 6, 9 and 10. The operating frequency plot (Figure 11) of a typical device shows f_{MAX1} or f_{MAX2} whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

f_{MAX1} is defined by $f_{MAX1} = 0.05/(t_{D(OFF)} + t_{D(ON)})$. Dead-time (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible. $t_{D(OFF)}$ and $t_{D(ON)}$ are defined in Figure 17. Device turn-off delay can establish an additional frequency limiting condition for an application other than T_{JMAX} . $t_{D(OFF)}$ is important when controlling output ripple under a lightly loaded condition.

f_{MAX2} is defined by $f_{MAX2} = (P_D - P_C)/(E_{OFF} + E_{ON})$. The allowable dissipation (P_D) is defined by $P_D = (T_{JMAX} - T_C)/R_{\theta JC}$. The sum of device switching and conduction losses must not exceed P_D . A 50% duty factor was used (Figure 11) and the conduction losses (P_C) are approximated by $P_C = (V_{CE} \times I_{CE})/2$.

E_{ON} and E_{OFF} are defined in the switching waveforms shown in Figure 17. E_{ON} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-on and E_{OFF} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-off. All tail losses are included in the calculation for E_{OFF} ; i.e., the collector current equals zero ($I_{CE} = 0$).

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