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# FIN3385 / FIN3386

## Low-Voltage, 28-Bit, Flat-Panel Display Link Serializer / Deserializer

### Features

- Operation -40°C to +85°C
- Low Power Consumption
- 20MHz to 85MHz Shift Clock Support
- ±1V Common-Mode Range around 1.2V
- Narrow Bus Reduces Cable Size and Cost
- High Throughput (up to 2.38Gbps)
- Internal PLL with No External Component
- Compatible with TIA/EIA-644 Specification
- 56-Lead, TSSOP Package

### Description

The FIN3385 and FIN3386 transform 28-bit wide parallel Low-Voltage TTL (LVTTTL) data into four serial Low Voltage Differential Signaling (LVDS) data streams. A phase-locked transmit clock is transmitted in parallel with the data stream over a separate LVDS link. Every cycle of transmit clock, 28-bits of input LVTTTL data are sampled and transmitted.

The FIN3386 receives and converts the 4/3 serial LVDS data streams back into 28/21 bits of LVTTTL data, acting as the deserializer.

For the FIN3385, at a transmit clock frequency of 85MHz, 28-bits of LVTTTL data are transmitted at a rate of 595Mbps per LVDS channel.

This pair solves EMI and cable size problems associated with wide and high-speed TTL interfaces.

### Ordering Information

Part Number	Operating Temperature Range	Package	Packing Method
FIN3385MTDX	-40 to +85°C	56-Lead Thin-Shrink Small-Outline Package (TSSOP), JEDEC MO-153,6.1mm Wide	Tape and Reel
FIN3386MTDX			

Block Diagrams

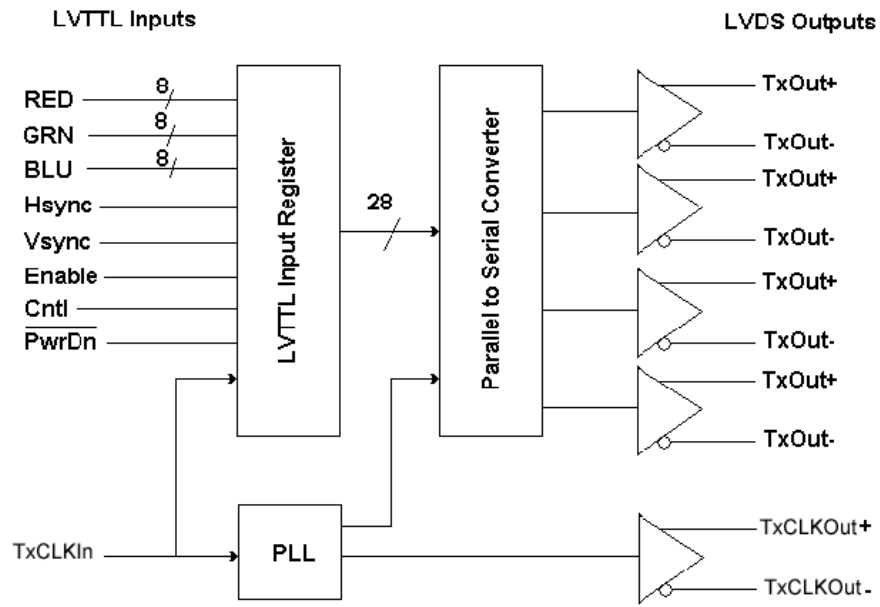


Figure 1. FIN3385 Transmitter Functional Diagram

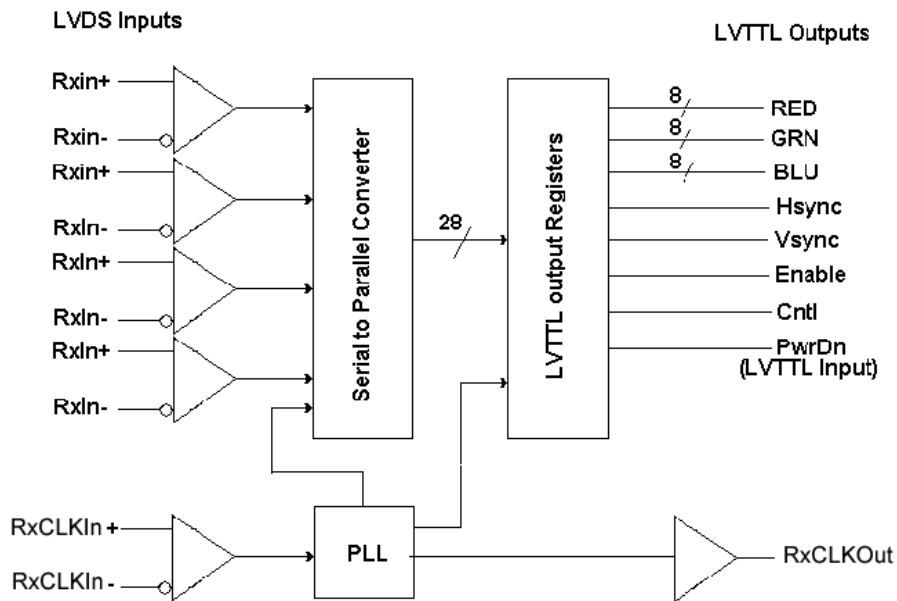


Figure 2. FIN3386 Receiver Functional Diagram

## Transmitter Pin Configuration

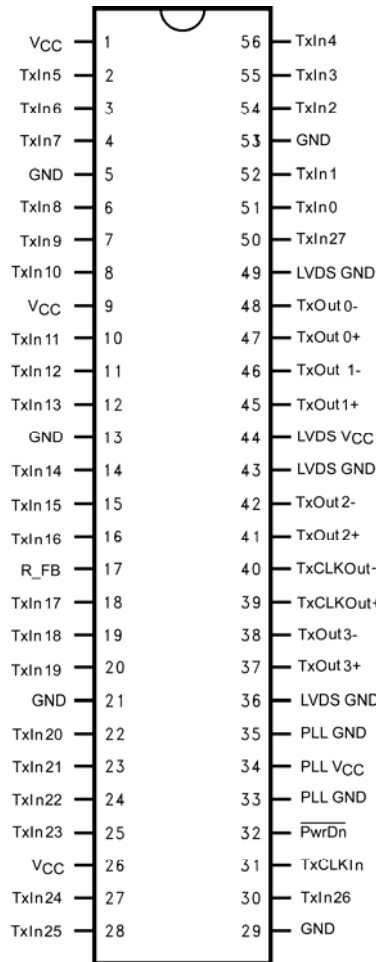


Figure 3. FIN3385 (28:4 Transmitter) Pin Assignments

### Pin Definitions

Pin Names	I/O Types	Number of Pins	Description of Signals
TxIn	I	28/21	LVTTTL Level Input
TxCLKIn	I	1	LVTTTL Level Clock Input, the rising edge is for data strobe
TxOut+	O	4/3	Positive LVDS Differential Data Output
TxOut-	O	4/3	Negative LVDS Differential Data Output
TxCLKOut+	O	1	Positive LVDS Differential Clock Output
TxCLKOut-	O	1	Negative LVDS Differential Clock Output
R_FB	I	1	Rising Edge Data Strobe: Assert HIGH ( $V_{CC}$ ) Falling Edge Data Strobe: Assert LOW (Ground)
/PwrDn	I	1	LVTTTL Level Power-Down Input Assertion (LOW) puts the outputs in High-Impedance state
PLL $V_{CC}$	I	1	Power Supply Pin for PLL
PLL GND	I	2	Ground Pins for PLL
LVDS $V_{CC}$	I	1	Power Supply Pin for LVDS Output
LVDS GND	I	3	Ground Pins for LVDS Output
$V_{CC}$	I	3	Power Supply Pins for LVTTTL Input
GND	I	5	Ground Pin for LVTTTL Input

## Receiver Pin Configuration

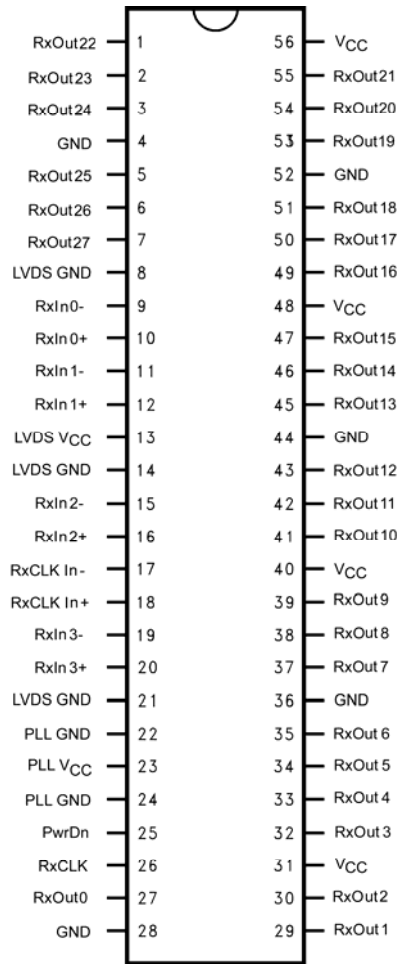


Figure 4. FIN3386 (28:4 Receiver) Pin Assignments

## Pin Definitions

Pin Names	I/O Types	Number of Pins	Description of Signals
RxIn	I	4/3	Negative LVDS Differential Data Output
RxIn+	I	4/3	Positive LVDS Differential Data Output
RxCLKIn-	I	1	Negative LVDS Differential Data Input
RxCLKIn+	I	1	Positive LVDS Differential Clock Input
RxOut	O	28/21	LVTTL Level Data Output, goes HIGH for /PwrDn LOW
RxCLKOut-	O	1	LVTTL Clock Output
/PwrDn	I	1	LVTTL Level Input. Refer to Table 2
PLL V <sub>CC</sub>	I	1	Power Supply Pin for PLL
PLL GND	I	2	Ground Pins for PLL
LVDS V <sub>CC</sub>	I	1	Power Supply Pin for LVDS Input
LVDS GND	I	3	Ground Pins for LVDS Input
V <sub>CC</sub>	I	4	Power Supply for LVTTL Output
GND	I	5	Ground Pins for LVTTL Output

## Truth Tables

**Table 1. Input / Output Truth Table**

Inputs			Outputs	
TxIn	TxCLKIn	/PwrDn <sup>(1)</sup>	TxOut±	TxCLKOut±
Active	Active	HIGH	LOW / HIGH	LOW / HIGH
Active	LOW / HIGH / High Impedance	HIGH	LOW / HIGH	Don't Care <sup>(2)</sup>
Floating	Active	HIGH	LOW	LOW / HIGH
Floating	Floating	HIGH	LOW	Don't Care <sup>(2)</sup>
Don't Care	Don't Care	LOW	High Impedance	High Impedance

**Notes:**

1. The outputs of the transmitter or receiver remain in a high-impedance state until  $V_{CC}$  reaches 2V.
2. TxCLKOut± settles at a free-running frequency when the part is powered up, /PwrDn is HIGH, and the TxCLKIn is a steady logic level (LOW / HIGH / High-Impedance).

### Power-Up / Power-Down Operation Truth Tables

The outputs of the transmitter remain in the High-Impedance state until the power supply reaches 2V. Table 2 shows the operation of the transmitter during power-up and power-down and operation of the /PwrDn pin.

**Table 2. Transmitter Power-Up / Power-Down Operation Truth Table**

		PwrDn	Normal
$V_{CC}$	<2V	>2V	>2V
TxIN	Don't Care	Don't Care	Active
TxOUT	High Impedance	High Impedance	Active
TxCLKIn	Don't Care	Don't Care	Active
TxCLKOut±	High Impedance	High Impedance	Active
/PwrDn	LOW	LOW	HIGH

**Table 3. Receiver Power-Up / Power-Down Operation Truth Table**

		/PwrDn				
RxIn±	Don't Care	Don't Care	Active	Active	Note 3	Note 3
RxOut	High Impedance	LOW	LOW/HIGH	Last Valid State	HIGH	Last Valid State
RxCLKIn±	Don't Care	Don't Care	Active	Note 3	Note 3	Note 3
RxCLKOut	High Impedance	Note 4	Active	Note 4	Note 4	Note 4
/PwrDn	LOW	LOW	HIGH	HIGH	HIGH	HIGH
$V_{CC}$	<2V	<2V	<2V	<2V	<2V	<2V

**Notes:**

3. If the input is terminated and un-driven (high-impedance) or shorted or open (fail-safe condition).
4. For /PwrDn or fail-safe condition, the RxCLKOut pin goes LOW for panel link devices and HIGH for channel link devices.
5. Shorted means (± inputs are shorted to each other, or ± inputs are shorted to each other and ground or  $V_{CC}$ , or either ± inputs are shorted to ground or  $V_{CC}$ ) with no other current/voltage sources (noise) applied. If the  $V_{ID}$  is still in the valid range (greater than 100mV) and  $V_{CM}$  is in the valid range (0V to 2.4V), the input signal is still recognized and the part responds normally.

## Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter	Min.	Max.	Unit
$V_{CC}$	Power Supply Voltage	-0.3	+4.6	V
$V_{ID\_TTL}$	TTL/CMOS Input/Output Voltage	-0.5	+4.6	V
$V_{IO\_LVDS}$	LVDS Input/Output Voltage	-0.3	+4.6	V
$I_{OSD}$	LVDS Output Short-Circuit Current	Continuous		
$T_{STG}$	Storage Temperature Range	-65	+150	°C
$T_J$	Maximum Junction Temperature		+150	°C
$T_L$	Lead Temperature, Soldering, 4 Seconds		+260	°C
ESD	Human Body Model, JESD22-A114 (1.5k $\Omega$ , 100pF)	I/O to GND	>10.0	kV
		All Pins	>6.5	
	Machine Model, JESD22-A115 (0 $\Omega$ , 200pF)	>400	V	

### Note:

- Absolute maximum ratings are DC values beyond which the device may be damaged or have its useful life impaired. The datasheet specifications should be met, without exception, to ensure that the system design is reliable over its power supply, temperature, and output/input loading variables.

## Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter	Min.	Max.	Unit
$V_{CC}$	Supply Voltage	3.0	3.6	V
$T_A$	Operating Temperature	-40	+85	°C
$V_{CCNPP}$	Maximum Supply Noise Voltage <sup>(7)</sup>		100	mV <sub>PP</sub>

### Note:

- 100mV  $V_{CC}$  noise should be tested for frequency at least up to 2MHz. All the specifications should be met under such noise.

## Transmitter DC Electrical Characteristics

Typical values are at  $T_A=25^\circ\text{C}$  and with  $V_{CC}=3.3\text{V}$ ; minimum and maximum are at over supply voltages and operating temperatures ranges, unless otherwise specified.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit	
<b>Transmitter LVTTTL Input Characteristics</b>							
$V_{IH}$	Input HIGH Voltage		2.0		$V_{CC}$	V	
$V_{IL}$	Input LOW Voltage		GND		0.8	V	
$V_{IK}$	Input Clamp Voltage	$I_{IK}=-18\text{mA}$		-0.79	-1.50	V	
$I_{IN}$	Input Current	$V_{IN}=0.4\text{V to }4.6\text{V}$		1.8	10.0	$\mu\text{A}$	
		$V_{IN}=\text{GND}$	-10	0			
<b>Transmitter LVDS Output Characteristics<sup>(8)</sup></b>							
$V_{OD}$	Output Differential Voltage	$R_L=100\Omega$ , Figure 5	250		450	mV	
$\Delta V_{OD}$	$V_{OD}$ Magnitude Change from Differential LOW-to-HIGH				35	mV	
$V_{OS}$	Offset Voltage		1.125	1.250	1.375	V	
$\Delta V_{OS}$	Offset Magnitude Change from Differential LOW-to-HIGH			25		mV	
$I_{OS}$	Short-Circuit Output Current	$V_{OUT}=0\text{V}$		-3.5	-5.0	mA	
$I_{OZ}$	Disabled Output Leakage Current	$DO=0\text{V to }4.6\text{V}$ , $/PwrDn=0\text{V}$		$\pm 1$	$\pm 10$	$\mu\text{A}$	
<b>Transmitter Supply Current</b>							
$I_{CCWT}$	28:4 Transmitter Power Supply Current for Worst-Case Pattern (with Load) <sup>(9)</sup>	$R_L=100\Omega$ Figure 8	32.5MHz		31.0	49.5	mA
			40MHz		32.0	55.0	
			66MHz		37.0	60.5	
			85MHz		42.0	66.0	
$I_{CCPDT}$	Powered-Down Supply Current	$/PwrDn=0.8\text{V}$		10.0	55.0	$\mu\text{A}$	
$I_{CCGT}$	28:4 Transmitter Supply Current for 16 Grayscale <sup>(9)</sup>	Figure 23 <sup>(10)</sup>	32.5MHz		29.0	41.8	mA
			40MHz		30.0	44.0	
			66MHz		35.0	49.5	
			85MHz		39.0	55.0	

### Notes:

- Positive current values refer to the current flowing into device and negative values refer to current flowing out of pins. Voltages are referenced to ground unless otherwise specified (except  $\Delta V_{OD}$  and  $V_{OD}$ ).
- The power supply current for both transmitter and receiver can vary with the number of active I/O channels.
- The 16-grayscale test pattern tests device power consumption for a "typical" LCD display pattern. The test pattern approximates signal switching needed to produce groups of 16 vertical strips across the display.



## Transmitter AC Electrical Characteristics

Typical values are at  $T_A=25^\circ\text{C}$  and with  $V_{CC}=3.3\text{V}$ ; minimum and maximum are at over supply voltages and operating temperatures ranges, unless otherwise specified.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
$t_{TCP}$	Transmit Clock Period	Figure 9	11.76	T	50.00	ns
$t_{TCH}$	Transmit Clock (TxCLKIn) HIGH Time		0.35	0.50	0.65	T
$t_{TCL}$	Transmit Clock LOW Time		0.35	0.50	0.65	T
$t_{CLKT}$	TxCLKIn Transition Time (Rising and Falling)	(10% to 90%) Figure 10	1.0		6.0	ns
$t_{JIT}$	TxCLKIn Cycle-to-Cycle Jitter				3.0	
$t_{XIT}$	TxIn Transition Time		1.5		6.0	ns
<b>LVDS Transmitter Timing Characteristics</b>						
$t_{TLH}$	Differential Output Rise Time (20% to 80%)	Figure 8		0.75	1.50	ns
$t_{THL}$	Differential Output Fall Time (20% to 80%)			0.75	1.50	ns
$t_{STC}$	TxIn Setup to TxCLNIn	Figure 9 $f=85\text{MHz}$	2.5			ns
$t_{HTC}$	TxIn Holds to TxCLNIn		0			ns
$t_{TPDD}$	Transmitter Power-Down Delay	Figure 14 <sup>(11)</sup>			100	ns
$t_{TCCD}$	Transmitter Clock Input to Clock Output Delay	( $T_A=25^\circ\text{C}$ and with $V_{CC}=3.3\text{V}$ ) Figure 13	2.8	5.5	6.8	ns
<b>Transmitter Output Data Jitter (<math>f=40\text{MHz}</math>)<sup>(12)</sup></b>						
$t_{TPPB0}$	Transmitter Output Pulse Position of Bit 0	Figure 20 $a = \frac{1}{f \times 7}$	-0.25	0	0.25	ns
$t_{TPPB1}$	Transmitter Output Pulse Position of Bit 1		$a-0.25$	$a$	$a+0.25$	ns
$t_{TPPB2}$	Transmitter Output Pulse Position of Bit 2		$2a-0.25$	$2a$	$2a+0.25$	ns
$t_{TPPB3}$	Transmitter Output Pulse Position of Bit 3		$3a-0.25$	$3a$	$3a+0.25$	ns
$t_{TPPB4}$	Transmitter Output Pulse Position of Bit 4		$4a-0.25$	$4a$	$4a+0.25$	ns
$t_{TPPB5}$	Transmitter Output Pulse Position of Bit 5		$5a-0.25$	$5a$	$5a+0.25$	ns
$t_{TPPB6}$	Transmitter Output Pulse Position of Bit 6		$6a-0.25$	$6a$	$6a+0.25$	ns
<b>Transmitter Output Data Jitter (<math>f=65\text{MHz}</math>)<sup>(12)</sup></b>						
$t_{TPPB0}$	Transmitter Output Pulse Position of Bit 0	Figure 20 $a = \frac{1}{f \times 7}$	-0.2	0	0.2	ns
$t_{TPPB1}$	Transmitter Output Pulse Position of Bit 1		$a-0.2$	$a$	$a+0.2$	ns
$t_{TPPB2}$	Transmitter Output Pulse Position of Bit 2		$2a-0.2$	$2a$	$2a+0.2$	ns
$t_{TPPB3}$	Transmitter Output Pulse Position of Bit 3		$3a-0.2$	$3a$	$3a+0.2$	ns
$t_{TPPB4}$	Transmitter Output Pulse Position of Bit 4		$4a-0.2$	$4a$	$4a+0.2$	ns
$t_{TPPB5}$	Transmitter Output Pulse Position of Bit 5		$5a-0.2$	$5a$	$5a+0.2$	ns
$t_{TPPB6}$	Transmitter Output Pulse Position of Bit 6		$6a-0.2$	$6a$	$6a+0.2$	ns

Continued on the following page...

### Transmitter AC Electrical Characteristics (Continued)

Over supply voltage and operating temperature ranges, unless otherwise specified.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
<b>Transmitter Output Data Jitter (f=85MHz)<sup>(12)</sup></b>						
t <sub>TPPB0</sub>	Transmitter Output Pulse Position of Bit 0	Figure 20 $a = \frac{1}{f \times 7}$	-0.2	0	0.2	ns
t <sub>TPPB1</sub>	Transmitter Output Pulse Position of Bit 1		a-0.2	a	a+0.2	ns
t <sub>TPPB2</sub>	Transmitter Output Pulse Position of Bit 2		2a-0.2	2a	2a+0.2	ns
t <sub>TPPB3</sub>	Transmitter Output Pulse Position of Bit 3		3a-0.2	3a	3a+0.2	ns
t <sub>TPPB4</sub>	Transmitter Output Pulse Position of Bit 4		4a-0.2	4a	4a+0.2	ns
t <sub>TPPB5</sub>	Transmitter Output Pulse Position of Bit 5		5a-0.2	5a	5a+0.2	ns
t <sub>TPPB6</sub>	Transmitter Output Pulse Position of Bit 6		6a-0.2	6a	6a+0.2	ns
t <sub>JCC</sub>	FIN3385 Transmitter Clock Out Jitter, Cycle-to-Cycle, Figure 20	f=40MHz		350	370	ps
		f=65MHz		210	230	
		f=85MHz		110	150	
t <sub>TPLLS</sub>	Transmitter Phase Lock Loop Set Time <sup>(13)</sup>	Figure 26 <sup>(12)</sup>			10	ms

**Notes:**

11. Outputs of all transmitters stay in 3-STATE until power reaches 2V. Clock and data output begins to toggle 10ms after V<sub>CC</sub> reaches 3.0V and /PwrDn pin is above 1.5V.
12. This output data pulse position works for both transmitters for TTL inputs, except the LVDS output bit mapping difference (see Figure 18). Figure 20 shows the skew between the first data bit and clock output. A two-bit cycle delay is guaranteed when the MSB is output from transmitter.
13. This jitter specification is based on the assumption that PLL has a reference clock with cycle-to-cycle input jitter of less than 2ns.

## Receiver DC Characteristics

Typical values are at  $T_A=25^\circ\text{C}$  and with  $V_{CC}=3.3\text{V}$ . Minimum and maximum values are over supply voltage and operating temperature ranges unless otherwise specified. Positive current values refer to the current flowing into device and negative values refer to current flowing out of pins. Voltages are referenced to ground unless otherwise specified (except  $\Delta V_{OD}$  and  $V_{OD}$ ).

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit	
<b>LVTTTL/CMOS DC Characteristics</b>							
$V_{IH}$	Input High Voltage		2.0		$V_{CC}$	V	
$V_{IL}$	Input Low Voltage		GND		0.8	V	
$V_{OH}$	Output High Voltage	$I_{OH}=-0.4\text{mA}$	2.7	3.3		V	
$V_{OL}$	Output Low Voltage	$I_{OL}=2\text{mA}$		0.06	0.30	V	
$V_{IK}$	Input Clamp Voltage	$I_{IK}=-18\text{mA}$		-0.79	-1.50	V	
$I_{IN}$	Input Current	$V_{IN}=0\text{V to }4.6\text{V}$	-10		10	$\mu\text{A}$	
$I_{OFF}$	Input/Output Power-Off Leakage Current	$V_{CC}=0\text{V}$ , All LVTTTL Inputs / Outputs $0\text{V to }4.6\text{V}$			$\pm 10$	$\mu\text{A}$	
$I_{OS}$	Output Short-Circuit Current	$V_{OUT}=0\text{V}$		-60	-120	$\text{mA}$	
<b>Receiver LVDS Input Characteristics</b>							
$V_{TH}$	Differential Input Threshold HIGH	Figure 6, Table 4			100	mV	
$V_{TL}$	Differential Input Threshold LOW	Figure 6, Table 4	-100			mV	
$V_{ICM}$	Input Common Mode Range	Figure 6, Table 4	0.05		2.35	V	
$I_{IN}$	Input Current	$V_{IN}=2.4\text{V}$ , $V_{CC}=3.6\text{V}$ or $0\text{V}$			$\pm 10$	$\mu\text{A}$	
		$V_{IN}=0\text{V}$ , $V_{CC}=3.6\text{V}$ or $0\text{V}$			$\pm 10$		
<b>Receiver Supply Current</b>							
$I_{CCWR}$	4:28 Receiver Power Supply Current for Worst-Case Pattern with Load <sup>(14)</sup>	$C_L=8\text{pF}$ , Figure 7	32.5MHz			70	mA
			40.0MHz			75	
			66.0MHz			114	
			85.0MHz			135	
	3:21 Receiver Power Supply Current for Worst-Case Pattern with Load <sup>(14)</sup>		32.5MHz		49	60	
			40.0MHz		53	65	
			66.0MHz		78	100	
			85.0MHz		90	115	
$I_{CCPDT}$	Powered-Down Supply Current	/PwrDn=0.8V (RxOut Stays LOW)		NA	55	$\mu\text{A}$	

## Receiver AC Characteristics

Typical values are at  $T_A=25^\circ\text{C}$  and with  $V_{CC}=3.3\text{V}$ ; minimum and maximum are at over supply voltages and operating temperatures ranges, unless otherwise specified.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
$t_{RCOP}$	Receiver Clock Output (RxCLKOut) Period		11.76	T	50.00	
$t_{RCOL}$	RxCLKOut LOW Time	Figure 12 Rising Edge Strobe $f=85\text{MHz}$	4.0	5.0	6.0	ns
$t_{RCOH}$	RxCLKOut HIGH Time		4.5	5.0	6.5	ns
$t_{RSRC}$	RxOut Valid Prior to RxCLKOut		3.5			ns
$t_{RHRC}$	RxOut Valid After RxCLKOut		3.5			ns
$t_{ROLH}$	Output Rise Time (20% to 80%)	$C_L=8\text{pF}$ , Figure 8		2.0	3.5	ns
$t_{ROHL}$	Output Fall Time (80% to 20%)			1.8	3.5	
$t_{RCCD}$	Receiver Clock Input to Clock Output Delay <sup>(15)</sup>	$T_A=25^\circ\text{C}$ , $V_{CC}=3.3\text{V}$ , Figure 24	3.5	5.0	7.5	ns
$t_{RPPD}$	Receiver Power-Down Delay	Figure 17			1.0	$\mu\text{s}$
$t_{RSPB0}$	Receiver Input Strobe Position of Bit 0	Figure 21 $f=85\text{MHz}$	0.49	0.84	1.19	ns
$t_{RSPB1}$	Receiver Input Strobe Position of Bit 1		2.17	2.52	2.87	ns
$t_{RSPB2}$	Receiver Input Strobe Position of Bit 2		3.85	4.20	4.55	ns
$t_{RSPB3}$	Receiver Input Strobe Position of Bit 3		5.53	5.88	6.23	ns
$t_{RSPB4}$	Receiver Input Strobe Position of Bit 4		7.21	7.56	7.91	ns
$t_{RSPB5}$	Receiver Input Strobe Position of Bit 5		8.89	9.24	9.59	ns
$t_{RSPB6}$	Receiver Input Strobe Position of Bit 6		10.57	10.92	11.27	ns
$t_{RSKM}$	RxIN Skew Margin <sup>(16)</sup>	Figure 21	290			ps
$t_{RPLLS}$	Receiver Phase Lock Loop Set Time	Figure 21		10		ms
$t_{RCOP}$	Receiver Clock Output (RxCLKOut) Period	Figure 12	15	T	50	ns
$t_{RCOL}$	RxCLKOut LOW Time	Figure 12 Rising Edge Strobe $f=40\text{MHz}$	10.0	11.0		ns
$t_{RCOH}$	RxCLKOut HIGH Time		10.0	12.2		
$t_{RSRC}$	RxOUT Valid Prior to RxCLKOut		6.5	11.6		
$t_{RHRC}$	RxOUT Valid After RxCLKOut		6.0	11.6		
$t_{RCOL}$	RxCLKOut LOW Time	Figure 12, Rising Edge Strobe <sup>(17)</sup> $f=66\text{MHz}$	5.0	6.3	9.0	ns
$t_{RCOH}$	RxCLKOut HIGH Time		5.0	7.6	9.0	
$t_{RSRC}$	RxOUT Valid Prior to RxCLKOut		4.5	7.3		
$t_{RHRC}$	RxOUT Valid After RxCLKOut		4.0	6.3		
$t_{ROLH}$	Output Rise Time (20% to 80%)	$C_L=8\text{pF}$ <sup>(17)</sup> , Figure 12		2.0	5.0	ns
$t_{ROHL}$	Output Fall Time (20% to 80%)			1.8	5.0	
$t_{RCCD}$	Receiver Clock Input to Clock Output Delay <sup>(18)</sup>	Figure 14, $T_A=25^\circ\text{C}$ and $V_{CC}=3.3\text{V}$	3.5	5.0	7.5	ns
$t_{RPPD}$	Receiver Power-Down Delay	Figure 17			1.0	$\mu\text{s}$
$t_{RSPB0}$	Receiver Input Strobe Position of Bit 0	Figure 21, $f=40\text{MHz}$	1.00	1.40	2.15	ns
$t_{RSPB1}$	Receiver Input Strobe Position of Bit 1		4.50	5.00	5.80	
$t_{RSPB2}$	Receiver Input Strobe Position of Bit 2		8.10	8.50	9.15	
$t_{RSPB3}$	Receiver Input Strobe Position of Bit 3		11.6	11.9	12.6	
$t_{RSPB4}$	Receiver Input Strobe Position of Bit 4		15.1	15.6	16.3	
$t_{RSPB5}$	Receiver Input Strobe Position of Bit 5		18.8	19.2	19.9	
$t_{RSPB6}$	Receiver Input Strobe Position of Bit 6		22.5	22.9	23.6	

## Receiver AC Characteristics

Typical values are at  $T_A=25^{\circ}\text{C}$  and with  $V_{CC}=3.3\text{V}$ ; minimum and maximum are at over supply voltages and operating temperatures ranges, unless otherwise specified.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
$t_{RSP0}$	Receiver Input Strobe Position of Bit 0	Figure 21, $f=66\text{MHz}$	0.7	1.1	1.4	ns
$t_{RSP1}$	Receiver Input Strobe Position of Bit 1		2.9	3.3	3.6	
$t_{RSP2}$	Receiver Input Strobe Position of Bit 2		5.1	5.5	5.8	
$t_{RSP3}$	Receiver Input Strobe Position of Bit 3		7.3	7.7	8.0	
$t_{RSP4}$	Receiver Input Strobe Position of Bit 4		9.5	9.9	10.2	
$t_{RSP5}$	Receiver Input Strobe Position of Bit 5		11.7	12.1	12.4	
$t_{RSP6}$	Receiver Input Strobe Position of Bit 6		13.9	14.3	14.6	
$t_{RSKM}$	RxIn Skew Margin <sup>(19)</sup>	$f=40\text{MHz}$ , Figure 21	490			ps
		$f=66\text{MHz}$ , Figure 21	400			
$t_{RPLLS}$	Receiver Phase Lock Loop Set Time	Figure 15			10.0	ms

### Notes:

14. The power supply current for the receiver can vary with the number of I/O channels.
15. Total channel latency from serializer to deserializer is  $(t + t_{TCCD})$  where  $t$  is a clock period.
16. Receiver skew margin is defined as the valid sampling window after considering potential setup/hold time and minimum/maximum bit position.
17. For the receiver with falling-edge strobe, the definition of setup/hold time is slightly different from the one with rising-edge strobe. The clock reference point is the time when the clock falling edge passes through 2V. For hold time  $t_{RHRC}$ , the clock reference point is the time when falling edge passes through +0.8V.
18. Total channel latency from serializer to deserializer is  $(t + t_{CCD}) (2 \cdot t + t_{RCCD})$  where  $t$  is the clock period.
19. Receiver skew margin is defined as the valid sampling window after considering potential setup/hold time and minimum / maximum bit position.

## Test Circuits

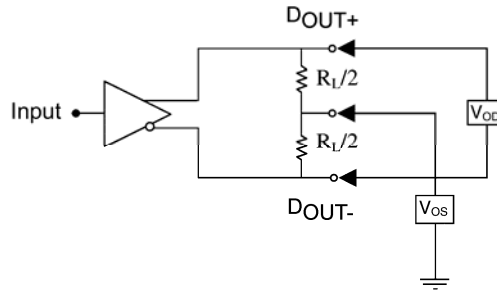
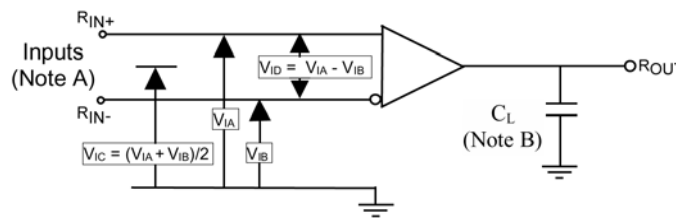


Figure 5. Differential LVDS Output DC Test Circuit



**Notes:**

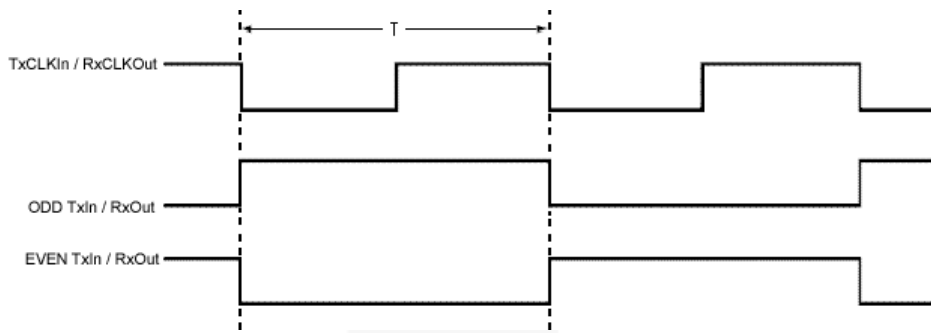
- A: For all input pulses,  $t_r$  or  $t_f \leq 1\text{ns}$ .
- B:  $C_L$  includes all probe and jig capacitance.

Figure 6. Differential Receiver Voltage Definitions, Propagation Delay, and Transition Time Test Circuit

Table 4. Receiver Minimum and Maximum Input Threshold Test Voltages

Applied Voltages (V)		Resulting Differential Input Voltage (mV)	Resulting Common Mode Input Voltage (V)
$V_{IA}$	$V_{IB}$	$V_{ID}$	$V_{ICM}$
1.25	1.15	100	1.20
1.15	1.25	-100	1.20
2.40	2.30	100	2.35
2.30	2.40	-100	2.35
0.10	0	100	0.05
0	0.10	-100	0.05
1.50	0.90	600	1.20
0.90	1.50	-600	1.20
2.40	1.80	600	2.10
1.80	2.40	-600	2.10
0.60	0	600	0.30
0	0.60	-600	0.30

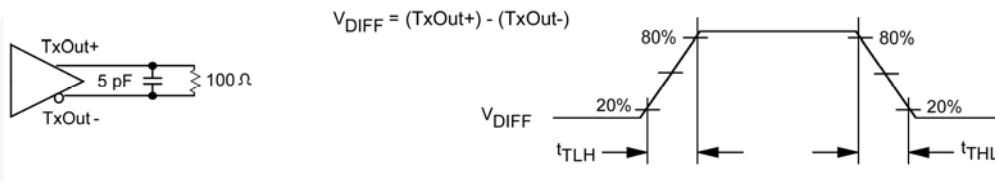
## AC Loadings and Waveforms



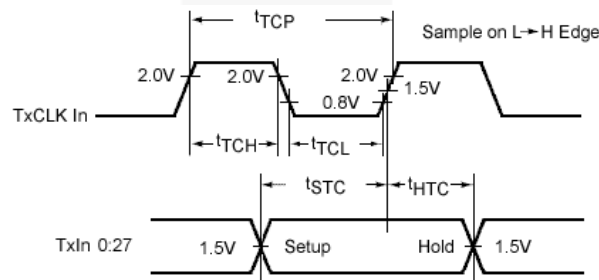
**Figure 7. Worst-Case Test Pattern**

**Note:**

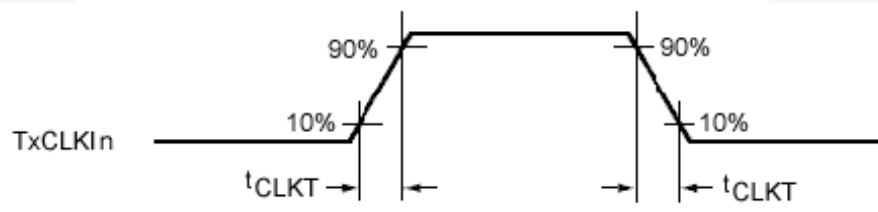
20. The worst-case test pattern produces a maximum toggling of digital circuits, LVDS I/O, and LVTTTL/CMOS I/O. Depending on the valid strobe edge of the transmitter, the TxCLKIn can be rising or falling edge data strobe.



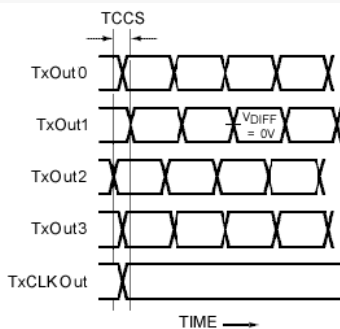
**Figure 8. Transmitter LVDS Output Load and Transition Times**



**Figure 9. Transmitter Setup/Hold and HIGH/LOW Times (Rising-Edge Strobe)**

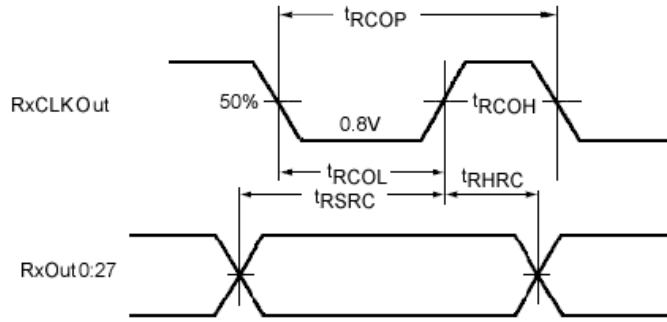


**Figure 10. Transmitter Input Clock Transition Time**



**Figure 11. Transmitter Outputs Channel-to-Channel Skew**

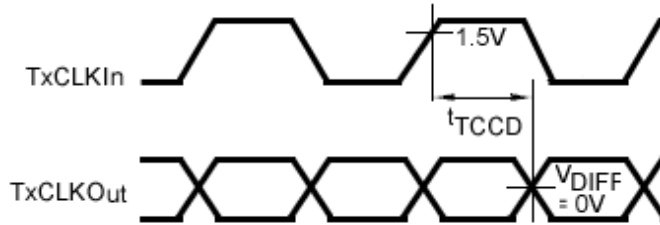
**AC Loadings and Waveforms (Continued)**



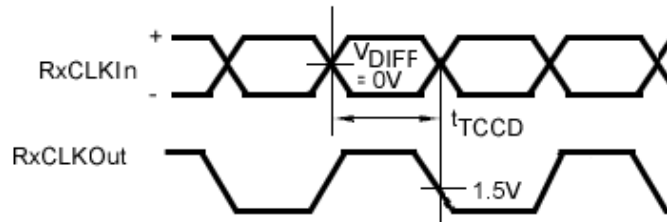
**Figure 12. Receiver Setup/Hold and HIGH/LOW Times**

**Note:**

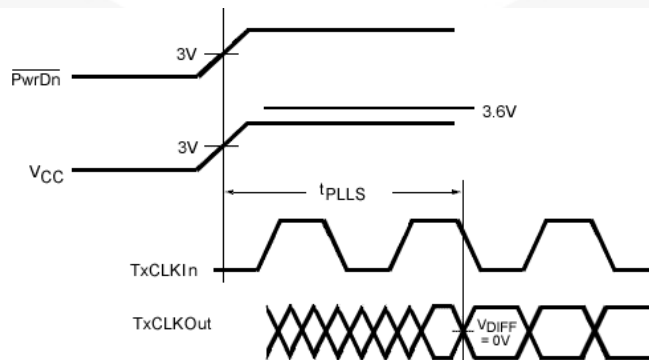
21. For the receiver with falling-edge strobe, the definition of setup/hold time is slightly different from the one with rising-edge strobe. The clock reference point is the time when the clock falling edge passes through 2V. For hold time  $t_{RHRC}$ , the clock reference point is the time when falling edge passes through +0.8V.



**Figure 13. Transmitter Clock-In to Clock-Out Delay (Rising-Edge Strobe)**



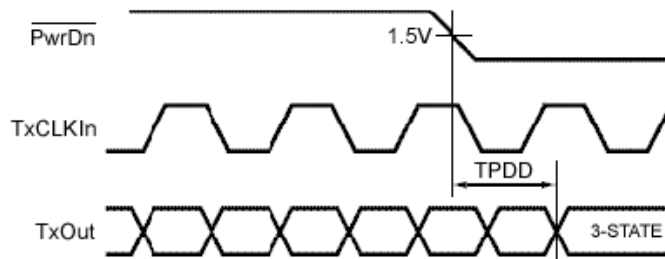
**Figure 14. Receiver Clock-In to Clock-Out Delay (Falling-Edge Strobe)**



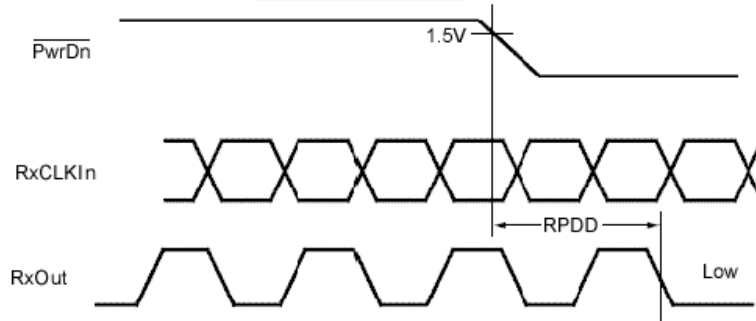
**Figure 15. Receiver Phase-Lock-Loop Set Time**



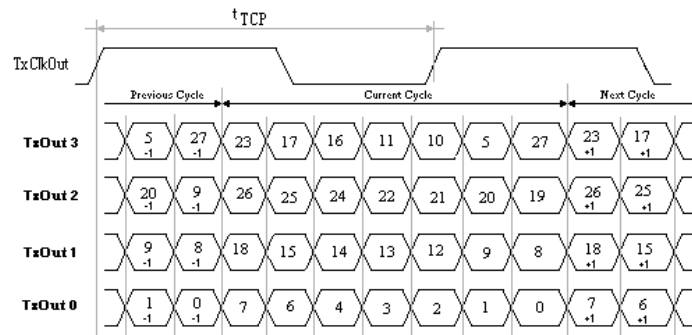
**AC Loadings and Waveforms (Continued)**



**Figure 16. Transmitter Power-Down Delay**



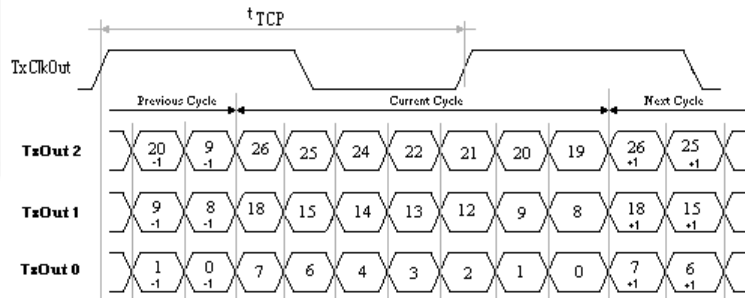
**Figure 17. Receiver Power-Down Delay**



**Figure 18. 28 Parallel LVTTTL Inputs Mapped to Four Serial LVDS Outputs**

**Note:**

22. The information in this diagram shows the difference between clock out and the first data bit. A 2-bit cycle delay is guaranteed when the MSB is output from the transmitter.

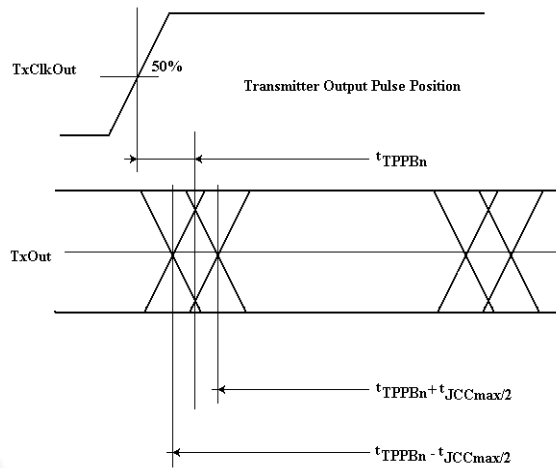


**Figure 19. 21 Parallel LVTTTL Inputs Mapped to Three Serial Outputs**

**Note:**

23. This output date pulse position works for both transmitters with 21 TTL inputs, except the LVDS output bit mapping difference. Two-bit cycle delay is guaranteed with the MSB is output from transmitter.

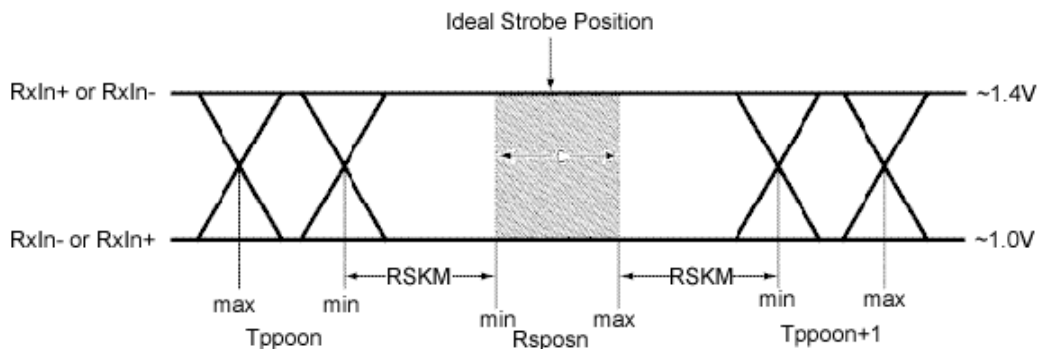
### AC Loadings and Waveforms (Continued)



**Figure 20. Transmitter Output Pulse Bit Position**



**Figure 21. Receiver Input Bit Position**

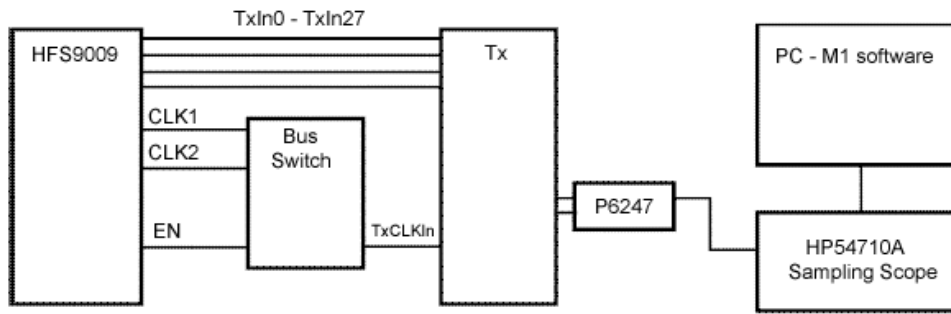


**Figure 22. Receiver LVDS Input Skew Margin**

**Note:**

24.  $t_{RSKM}$  is the budget for the cable skew and source clock skew plus Inter-Symbol Interference (ISI). The minimum and maximum pulse position values are based on the bit position of each of the seven bits within the LVDS data stream across PVT (Process, Voltage Supply, and Temperature).

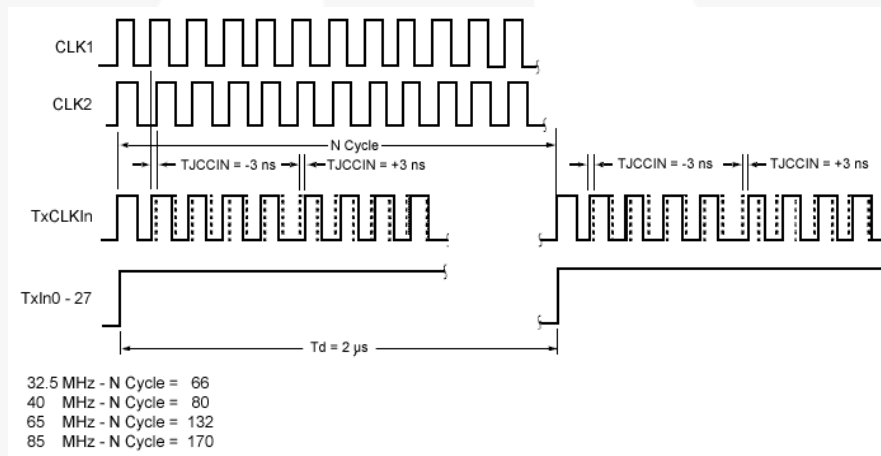
## AC Loadings and Waveforms (Continued)



**Figure 23. Transmitter Clock Out Jitter Measurement Setup**

**Note:**

25. Test setup considers no requirement for separation of RMS and deterministic jitter. Other hardware setups, such as Wavecrest boxes, can be used if no M1 software is available, but the test methodology in Figure 24 should be followed.



**Figure 24. Timing Diagram of Transmitter Clock Input with Jitter**

**Note:**

26. This jitter pattern is used to test the jitter response (clock out) of the device over the power supply range with worst jitter  $\pm 3\text{ns}$  (cycle-to-cycle) clock input. The specific test methodology is as follows:
27. Switching input data TxIn0 to TxIn20 at 0.5MHz and the input clock is shifted to left -3ns and to the right +3ns when data is HIGH.
28. The  $\pm 3\text{ns}$  cycle-to-cycle input jitter is the static phase error between the two clock sources. Jumping between two clock sources to simulate the worst-case of clock-edge jump (3ns) from graphical controllers. Cycle-to-cycle jitter at TxCLKOut pin should be measured cross  $V_{CC}$  range with 100mV noise ( $V_{CC}$  noise frequency  $< 2\text{MHz}$ ).

## AC Loadings and Waveforms (Continued)

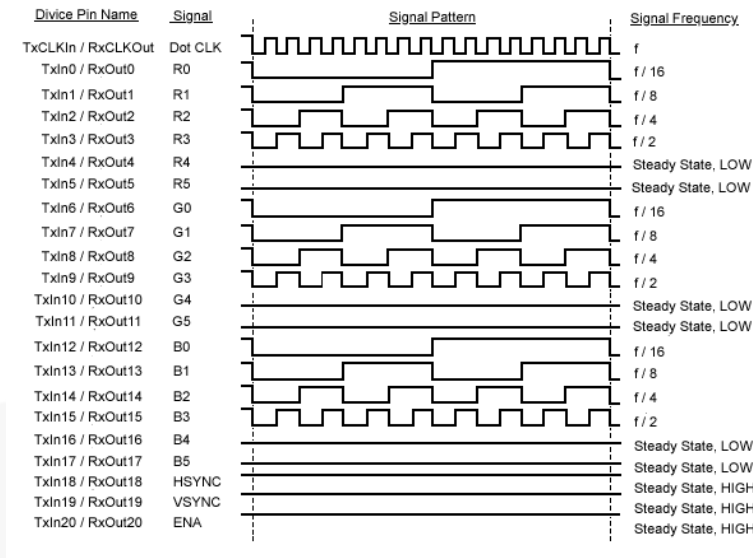


Figure 25. “16-Grayscale” Test Pattern

**Note:**

29. The 16-grayscale test pattern tests device power consumption for a “typical” LCD display pattern. The test pattern approximates signal switching needed to produce groups of 16 vertical strips across the display.

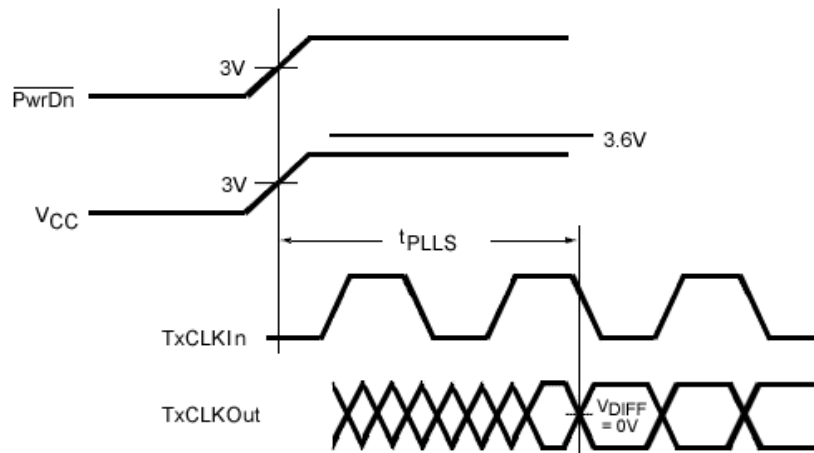


Figure 26. Transmitter Phase-Lock-Loop Time

### Physical Dimensions

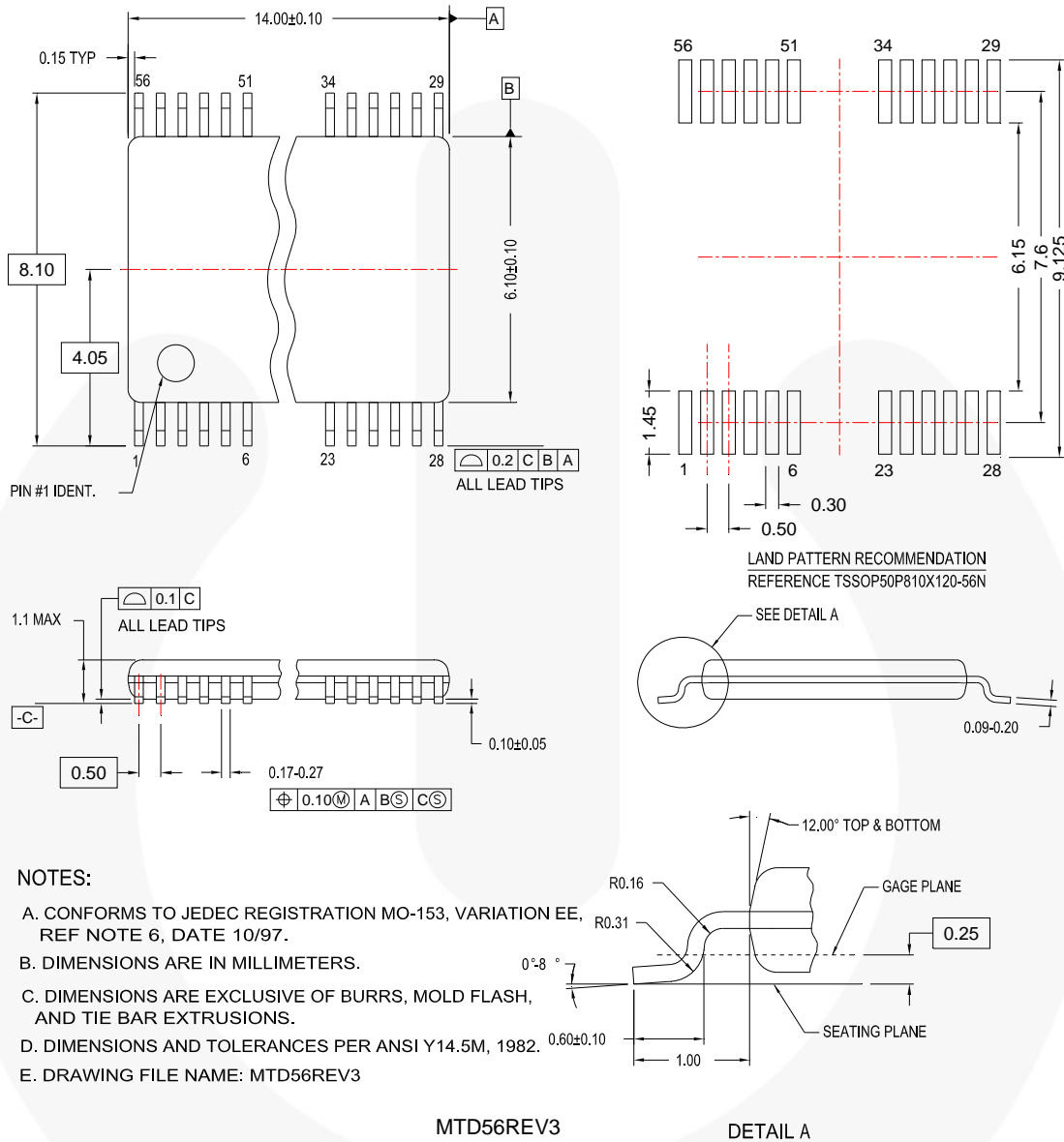


Figure 27. 56-Lead Thin Shrink Small Outline Package (TSSOP), JEDEC MO-153, 6.1mm Wide

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