

MAX2680/MAX2681/ MAX2682

400MHz to 2.5GHz, Low-Noise, SiGe Downconverter Mixers

General Description

The MAX2680/MAX2681/MAX2682 miniature, low-cost, low-noise downconverter mixers are designed for low-voltage operation and are ideal for use in portable communications equipment. Signals at the RF input port are mixed with signals at the local oscillator (LO) port using a double-balanced mixer. These downconverter mixers operate with RF input frequencies between 400MHz and 2500MHz, and downconvert to IF output frequencies between 10MHz and 500MHz.

The MAX2680/MAX2681/MAX2682 operate from a single +2.7V to +5.5V supply, allowing them to be powered directly from a 3-cell NiCd or a 1-cell Lithium battery. These devices offer a wide range of supply currents and input intercept (IIP3) levels to optimize system performance. Additionally, each device features a low-power shutdown mode in which it typically draws less than 0.1µA of supply current. Consult the *Selector Guide* for various combinations of IIP3 and supply current.

The MAX2680/MAX2681/MAX2682 are manufactured on a high-frequency, low-noise, advanced silicon-germanium process and are offered in the space-saving 6-pin SOT23 package.

Applications

- 400MHz/900MHz/2.4GHz ISM-Band Radios
- Personal Communications Systems (PCS)
- Cellular and Cordless Phones
- Wireless Local Loop
- IEEE-802.11 and Wireless Data

Features

- 400MHz to 2.5GHz Operation
- +2.7V to +5.5V Single-Supply Operation
- Low Noise Figure: 6.3dB at 900MHz (MAX2680)
- High Input Third-Order Intercept Point (IIP3 at 2450MHz)
 - -6.9dBm at 5.0mA (MAX2680)
 - +1.0dBm at 8.7mA (MAX2681)
 - +3.2dBm at 15.0mA (MAX2682)
- < 0.1µA Low-Power Shutdown Mode
- Ultra-Small Surface-Mount Packaging

Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE | SOT TOP MARK |
|--------------|----------------|-------------|--------------|
| MAX2680EUT-T | -40°C to +85°C | 6 SOT23 | AAAR |
| MAX2681EUT-T | -40°C to +85°C | 6 SOT23 | AAAS |
| MAX2682EUT-T | -40°C to +85°C | 6 SOT23 | AAAT |

Selector Guide

| PART | I _{CC} (mA) | FREQUENCY | | | | | | | | |
|---------|----------------------|------------|---------|-----------|------------|---------|-----------|------------|---------|-----------|
| | | 900MHz | | | 1950MHz | | | 2450MHz | | |
| | | IIP3 (dBm) | NF (dB) | GAIN (dB) | IIP3 (dBm) | NF (dB) | GAIN (dB) | IIP3 (dBm) | NF (dB) | GAIN (dB) |
| MAX2680 | 5.0 | -12.9 | 6.3 | 11.6 | -8.2 | 8.3 | 7.6 | -6.9 | 11.7 | 7.0 |
| MAX2681 | 8.7 | -6.1 | 7.0 | 14.2 | +0.5 | 11.1 | 8.4 | +1.0 | 12.7 | 7.7 |
| MAX2682 | 15.0 | -1.8 | 6.5 | 14.7 | +4.4 | 10.2 | 10.4 | +3.2 | 13.4 | 7.9 |

Typical Operating Circuit appears at end of data sheet.

MAX2680/MAX2681/ MAX2682

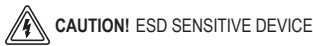
400MHz to 2.5GHz, Low-Noise, SiGe Downconverter Mixers

Absolute Maximum Ratings

| | |
|------------------------------------|--|
| V _{CC} to GND | -0.3V to +6.0V |
| RFIN Input Power (50Ω source)..... | +10dBm |
| LO Input Power (50Ω source) | +10dBm |
| SHDN, IFOUT, RFIN to GND | -0.3V to (V _{CC} + 0.3V) |
| LO to GND..... | (V _{CC} - 1V) to (V _{CC} + 0.3V) |

| | |
|---|-----------------|
| Continuous Power Dissipation (T _A = +70°C) | |
| SOT23 (derate 8.7mW/°C above +70°C)..... | 696mW |
| Operating Temperature Range..... | -40°C to +85°C |
| Junction Temperature..... | +150°C |
| Storage Temperature Range..... | -65°C to +160°C |
| Lead Temperature (soldering, 10s) | +300°C |

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.



DC Electrical Characteristics

(V_{CC} = +2.7V to +5.5V, SHDN = +2V, T_A = T_{MIN} to T_{MAX} unless otherwise noted. Typical values are at V_{CC} = +3V and T_A = +25°C. Minimum and maximum values are guaranteed over temperature by design and characterization.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|-----------------------------|-------------------|----------------------------|-----|------|------|-------|
| Operating Supply Current | I _{CC} | MAX2680 | | 5.0 | 7.7 | mA |
| | | MAX2681 | | 8.7 | 12.7 | |
| | | MAX2682 | | 15.0 | 21.8 | |
| Shutdown Supply Current | I _{CC} | SHDN = 0.5V | | 0.05 | | μA |
| Shutdown Input Voltage High | V _{IH} | | 2.0 | | | V |
| Shutdown Input Voltage Low | V _{IL} | | | | 0.5 | V |
| Shutdown Input Bias Current | I _{SHDN} | 0 < SHDN < V _{CC} | | 0.2 | | μA |

AC Electrical Characteristics

(MAX2680/1/2 EV Kit, V_{CC} = SHDN = +3.0V, T_A = +25°C, unless otherwise noted. RFIN and IFOUT matched to 50Ω. P_{LO} = -5dBm, P_{RFIN} = -25dBm.)

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|--|---|-----|-------|------|-------|
| MAX2680 | | | | | |
| RF Frequency Range | (Notes 1, 2) | 400 | | 2500 | MHz |
| LO Frequency Range | (Notes 1, 2) | 400 | | 2500 | MHz |
| IF Frequency Range | (Notes 1, 2) | 10 | | 500 | MHz |
| Conversion Power Gain | f _{RF} = 400MHz, f _{LO} = 445MHz, f _{IF} = 45MHz | | 7.3 | | dB |
| | f _{RF} = 900MHz, f _{LO} = 970MHz, f _{IF} = 70MHz | | 11.6 | | |
| | f _{RF} = 1950MHz, f _{LO} = 1880MHz, f _{IF} = 70MHz (Note 1) | 5.7 | 7.6 | 8.6 | |
| | f _{RF} = 2450MHz, f _{LO} = 2210MHz, f _{IF} = 240MHz | | 7.0 | | |
| Gain Variation Over Temperature | f _{RF} = 1950MHz, f _{LO} = 1880MHz, f _{IF} = 70MHz, T _A = T _{MIN} to T _{MAX} (Note 1) | | 1.9 | 2.4 | dB |
| Input Third-Order Intercept Point (Note 3) | f _{RF} = 900MHz, 901MHz, f _{LO} = 970MHz, f _{IF} = 70MHz | | -12.9 | | dBm |
| | f _{RF} = 1950MHz, 1951MHz, f _{LO} = 1880MHz, f _{IF} = 70MHz | | -8.2 | | |
| | f _{RF} = 2450MHz, 2451MHz, f _{LO} = 2210MHz, f _{IF} = 240MHz | | -6.9 | | |
| Noise Figure (Single Sideband) | f _{RF} = 900MHz, f _{LO} = 970MHz, f _{IF} = 70MHz | | 6.3 | | dB |
| | f _{RF} = 1950MHz, f _{LO} = 2020MHz, f _{IF} = 70MHz | | 8.3 | | |
| | f _{RF} = 2450MHz, f _{LO} = 2210MHz, f _{IF} = 240MHz | | 11.7 | | |
| LO Input VSWR | 50Ω source impedance | | 1.5:1 | | |
| LO Leakage at IFOUT Port | f _{LO} = 1880MHz | | -22 | | dBm |

AC Electrical Characteristics (continued)

(MAX2680/1/2 EV Kit, $V_{CC} = \overline{SHDN} = +3.0V$, $T_A = +25^\circ C$, unless otherwise noted. RFIN and IFOUT matched to 50Ω. $P_{LO} = -5dBm$, $P_{RFIN} = -25dBm$.)

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|--|---|-----|-------|------|-------|
| LO Leakage at RFIN Port | $f_{LO} = 1880MHz$ | | -26 | | dBm |
| IF/2 Spurious Response | $f_{RF} = 1915MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$ (Note 4) | | -51 | | dBm |
| MAX2681 | | | | | |
| RF Frequency Range | (Notes 1, 2) | 400 | | 2500 | MHz |
| LO Frequency Range | (Notes 1, 2) | 400 | | 2500 | MHz |
| IF Frequency Range | (Notes 1, 2) | 10 | | 500 | MHz |
| Conversion Power Gain | $f_{RF} = 400MHz$, $f_{LO} = 445MHz$, $f_{IF} = 45MHz$ | | 11.0 | | dB |
| | $f_{RF} = 900MHz$, $f_{LO} = 970MHz$, $f_{IF} = 70MHz$ | | 14.2 | | |
| | $f_{RF} = 1950MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$ (Note 1) | 6.7 | 8.4 | 9.4 | |
| | $f_{RF} = 2450MHz$, $f_{LO} = 2210MHz$, $f_{IF} = 240MHz$ | | 7.7 | | |
| Gain Variation Over Temperature | $f_{RF} = 1950MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$, $T_A = T_{MIN}$ to T_{MAX} (Note 1) | | 1.7 | 2.3 | dB |
| Input Third-Order Intercept Point (Note 3) | $f_{RF} = 900MHz$, $901MHz$, $f_{LO} = 970MHz$, $f_{IF} = 70MHz$ | | -6.1 | | dBm |
| | $f_{RF} = 1950MHz$, $1951MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$ | | +0.5 | | |
| | $f_{RF} = 2450MHz$, $2451MHz$, $f_{LO} = 2210MHz$, $f_{IF} = 240MHz$ | | +1.0 | | |
| Noise Figure (Single Sideband) | $f_{RF} = 900MHz$, $f_{LO} = 970MHz$, $f_{IF} = 70MHz$ | | 7.0 | | dB |
| | $f_{RF} = 1950MHz$, $f_{LO} = 2020MHz$, $f_{IF} = 70MHz$ | | 11.1 | | |
| | $f_{RF} = 2450MHz$, $f_{LO} = 2210MHz$, $f_{IF} = 240MHz$ | | 12.7 | | |
| LO Input VSWR | 50Ω source impedance | | 1.5:1 | | |
| LO Leakage at IFOUT Port | $f_{LO} = 1880MHz$ | | -23 | | dBm |
| LO Leakage at RFIN Port | $f_{LO} = 1880MHz$ | | -27 | | dBm |
| IF/2 Spurious Response | $f_{RF} = 1915MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$ (Note 4) | | -65 | | dBm |
| MAX2682 | | | | | |
| RF Frequency Range | (Notes 1, 2) | 400 | | 2500 | MHz |
| LO Frequency Range | (Notes 1, 2) | 400 | | 2500 | MHz |
| IF Frequency Range | (Notes 1, 2) | 10 | | 500 | MHz |
| Conversion Power Gain | $f_{RF} = 400MHz$, $f_{LO} = 445MHz$, $f_{IF} = 45MHz$ | | 13.4 | | dB |
| | $f_{RF} = 900MHz$, $f_{LO} = 970MHz$, $f_{IF} = 70MHz$ | | 14.7 | | |
| | $f_{RF} = 1950MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$ (Note 1) | 8.7 | 10.4 | 11.7 | |
| | $f_{RF} = 2450MHz$, $f_{LO} = 2210MHz$, $f_{IF} = 240MHz$ | | 7.9 | | |
| Gain Variation Over Temperature | $f_{RF} = 1950MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$, $T_A = T_{MIN}$ to T_{MAX} (Note 1) | | 2.1 | 3.2 | dB |
| Input Third-Order Intercept Point (Note 3) | $f_{RF} = 900MHz$, $901MHz$, $f_{LO} = 970MHz$, $f_{IF} = 70MHz$ | | -1.8 | | dBm |
| | $f_{RF} = 1950MHz$, $1951MHz$, $f_{LO} = 1880MHz$, $f_{IF} = 70MHz$ | | +4.4 | | |
| | $f_{RF} = 2450MHz$, $2451MHz$, $f_{LO} = 2210MHz$, $f_{IF} = 240MHz$ | | +3.2 | | |
| Noise Figure (Single Sideband) | $f_{RF} = 900MHz$, $f_{LO} = 970MHz$, $f_{IF} = 70MHz$ | | 6.5 | | dB |
| | $f_{RF} = 1950MHz$, $f_{LO} = 2020MHz$, $f_{IF} = 70MHz$ | | 10.2 | | |
| | $f_{RF} = 2450MHz$, $f_{LO} = 2210MHz$, $f_{IF} = 240MHz$ | | 13.4 | | |

AC Electrical Characteristics (continued)

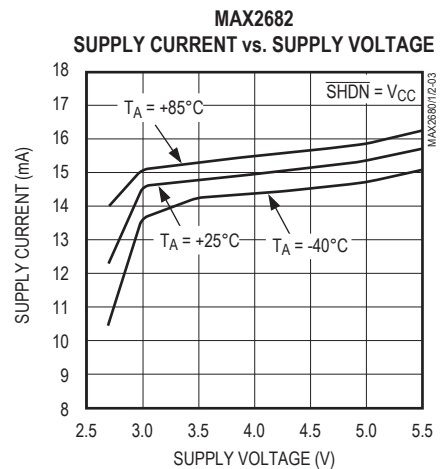
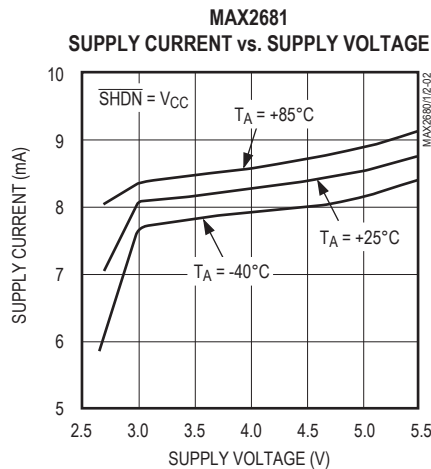
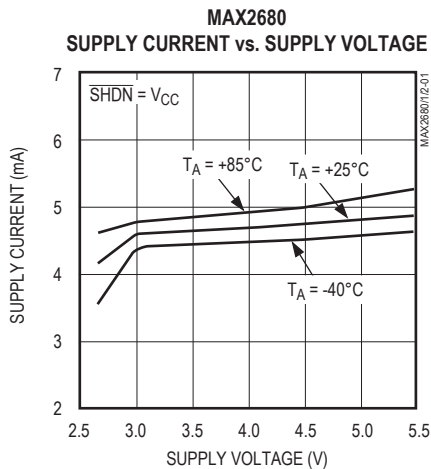
(MAX2680/1/2 EV Kit, $V_{CC} = \overline{\text{SHDN}} = +3.0\text{V}$, $T_A = +25^\circ\text{C}$, unless otherwise noted. RFIN and IFOUT matched to 50Ω . $P_{LO} = -5\text{dBm}$, $P_{\text{RFIN}} = -25\text{dBm}$.)

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|--------------------------|--|-----|-------|-----|-------|
| LO Input VSWR | 50Ω source impedance | | 1.5:1 | | |
| LO Leakage at IFOUT Port | $f_{LO} = 1880\text{MHz}$ | | -23 | | dBm |
| LO Leakage at RFIN Port | $f_{LO} = 1880\text{MHz}$ | | -27 | | dBm |
| IF/2 Spurious Response | $f_{RF} = 1915\text{MHz}$, $f_{LO} = 1880\text{MHz}$, $f_{IF} = 70\text{MHz}$ (Note 4) | | -61 | | dBm |

- Note 1:** Guaranteed by design and characterization.
- Note 2:** Operation outside of this specification is possible, but performance is not characterized and is not guaranteed.
- Note 3:** Two input tones at -25dBm per tone.
- Note 4:** This spurious response is caused by a higher-order mixing product (2×2). Specified RF frequency is applied and IF output power is observed at the desired IF frequency (70MHz).

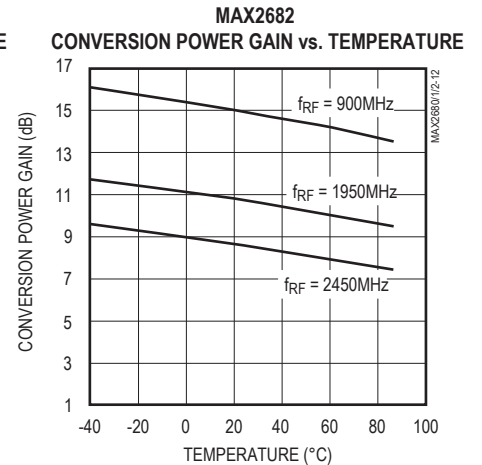
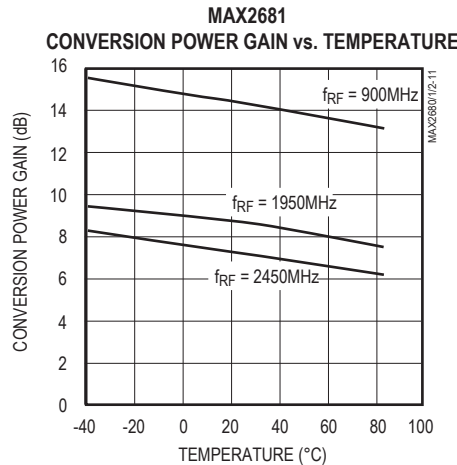
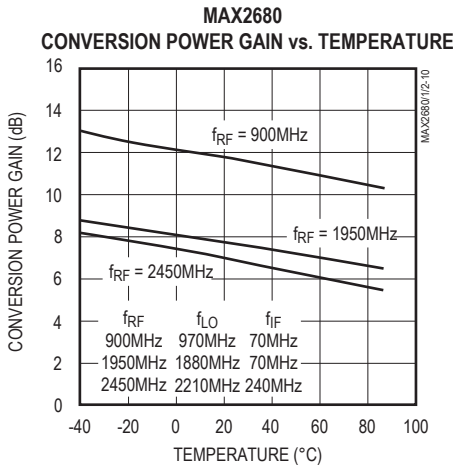
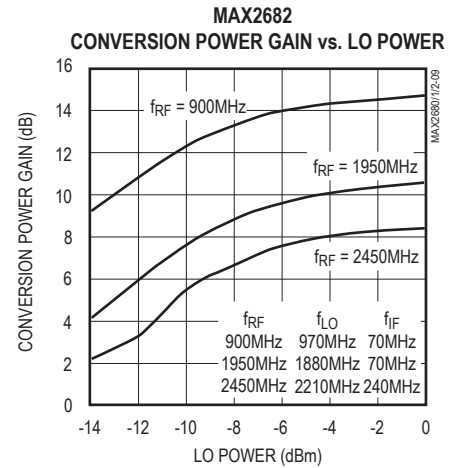
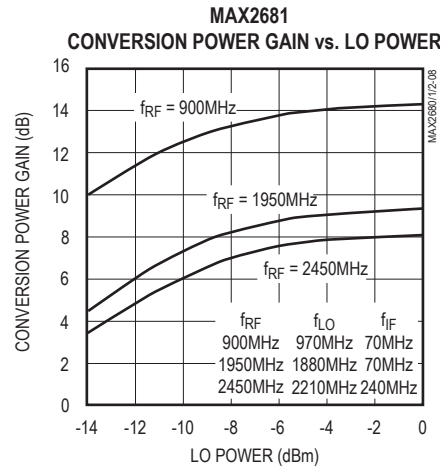
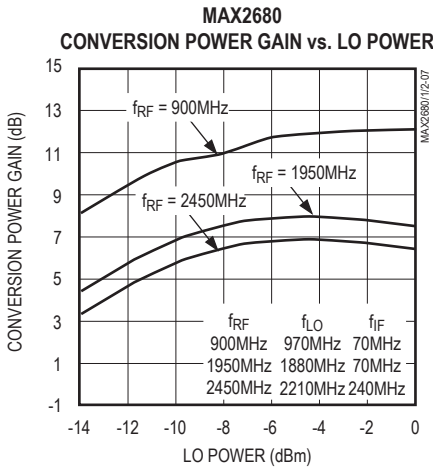
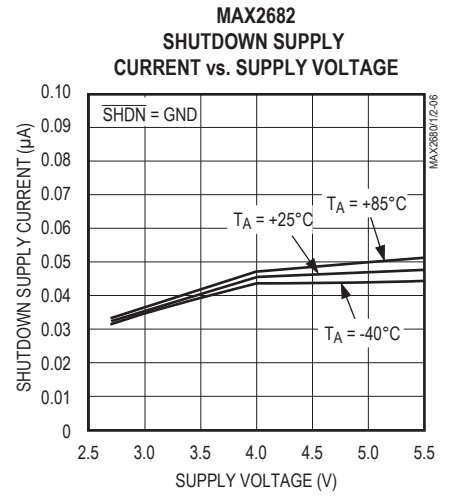
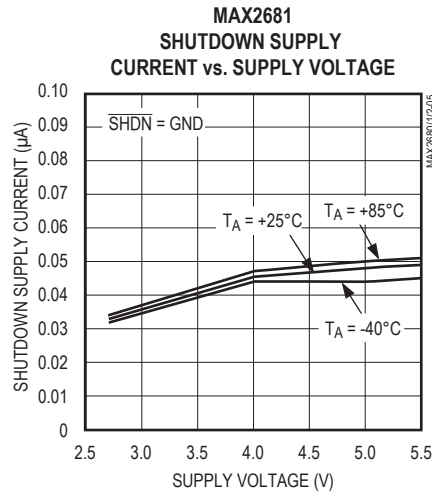
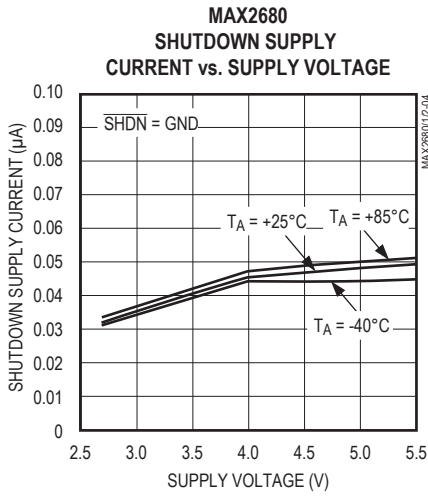
Typical Operating Characteristics

(*Typical Operating Circuit*, $V_{CC} = \overline{\text{SHDN}} = +3.0\text{V}$, $P_{\text{RFIN}} = -25\text{dBm}$, $P_{LO} = -5\text{dBm}$, $T_A = +25^\circ\text{C}$, unless otherwise noted.)



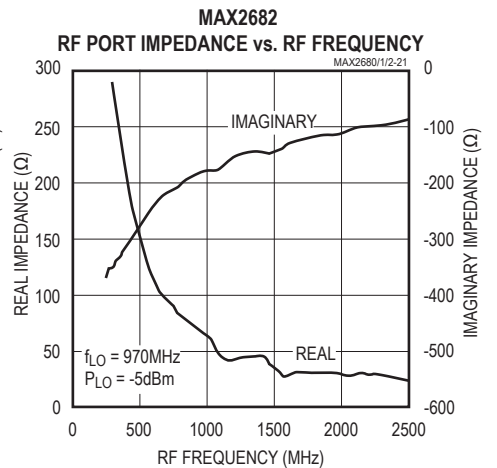
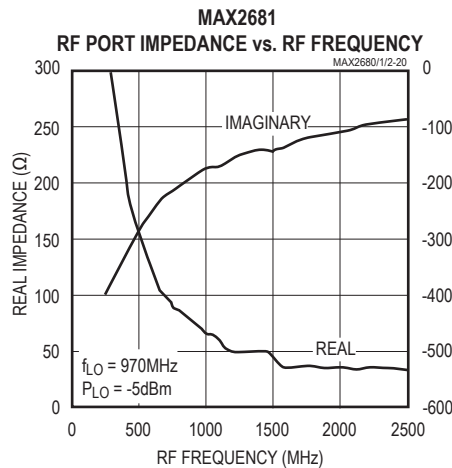
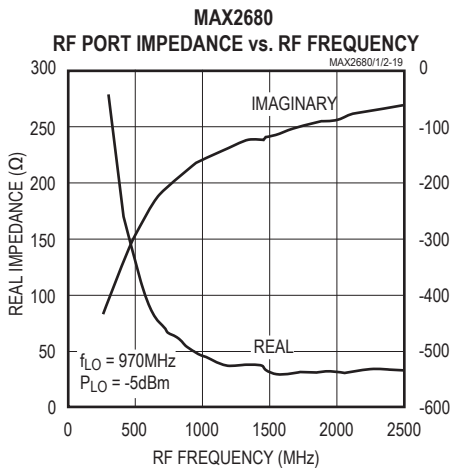
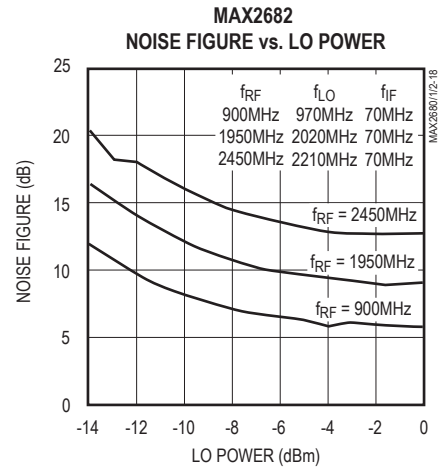
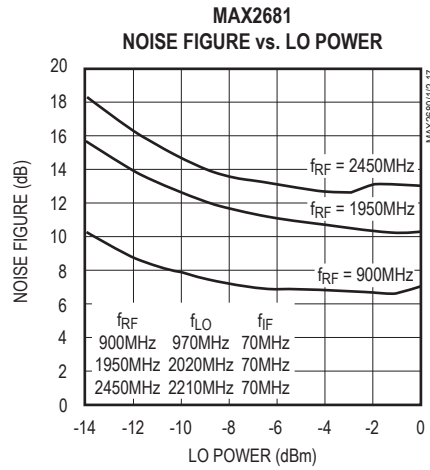
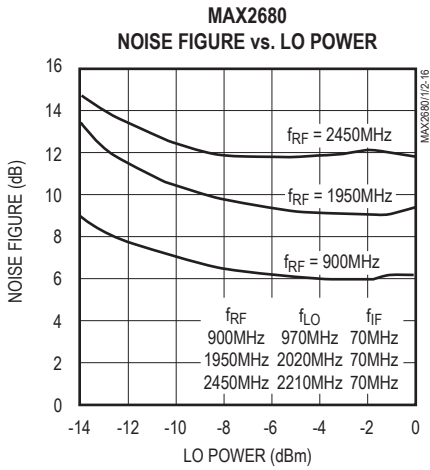
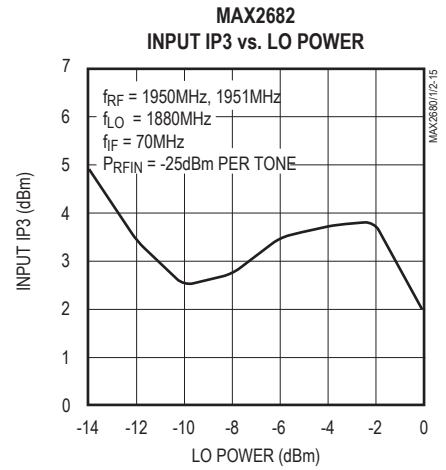
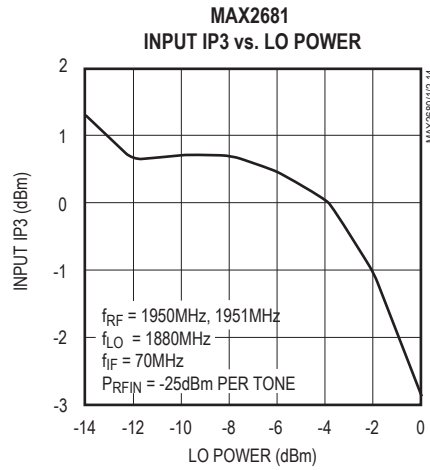
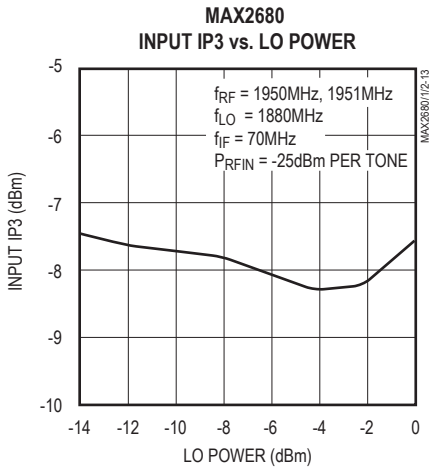
Typical Operating Characteristics (continued)

(Typical Operating Circuit, $V_{CC} = \overline{\text{SHDN}} = +3.0\text{V}$, $P_{\text{RFIN}} = -25\text{dBm}$, $P_{\text{LO}} = -5\text{dBm}$, $T_A = +25^\circ\text{C}$, unless otherwise noted.)



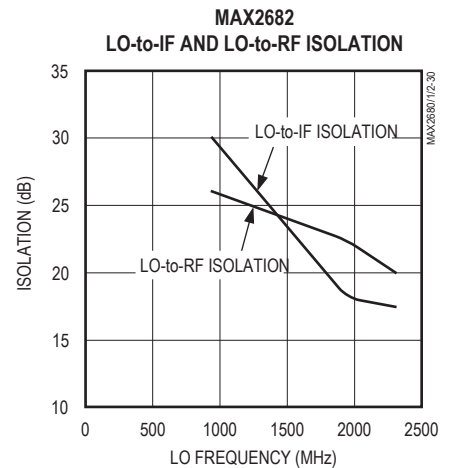
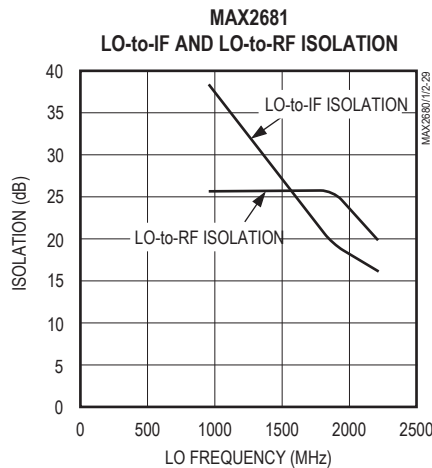
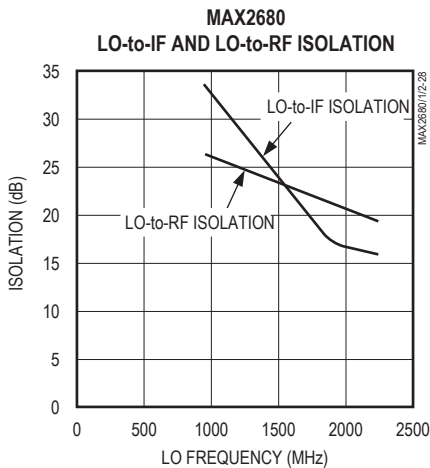
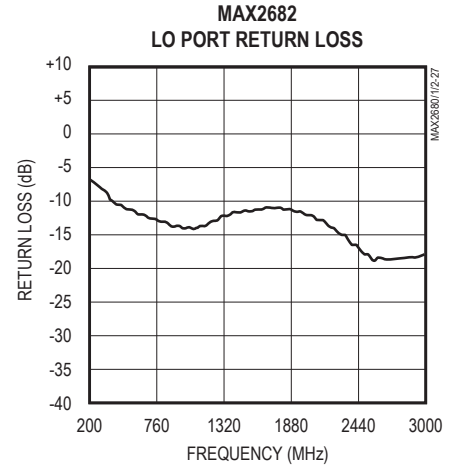
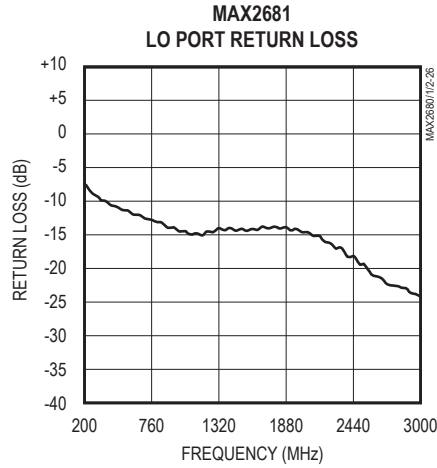
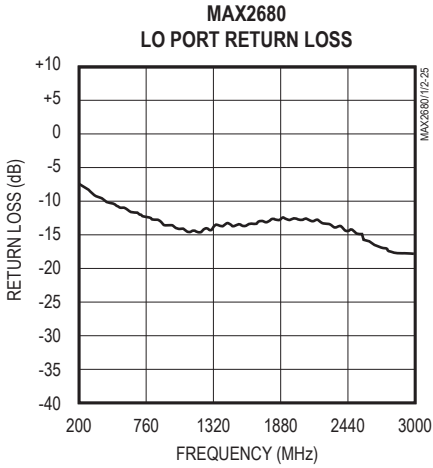
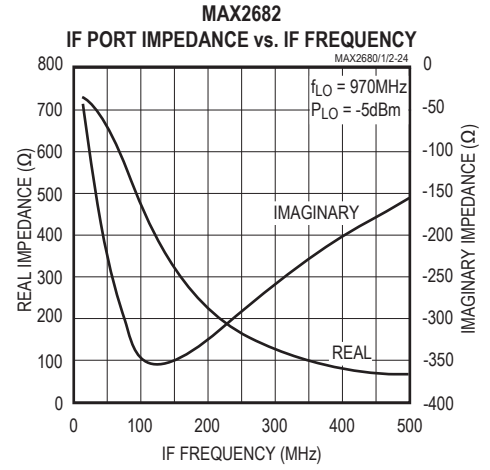
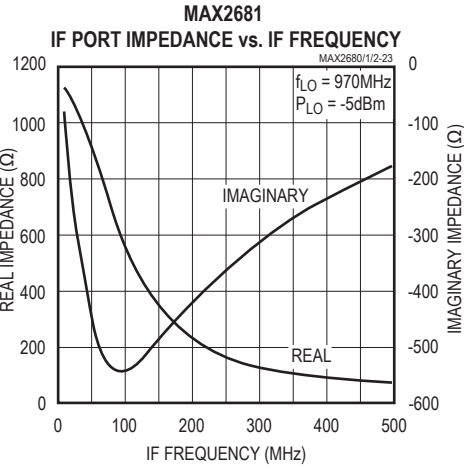
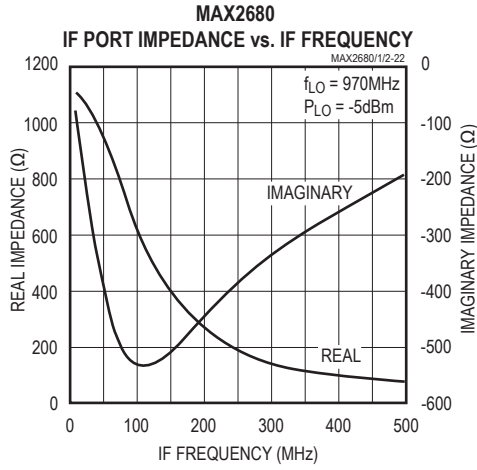
Typical Operating Characteristics (continued)

(Typical Operating Circuit, $V_{CC} = \overline{SHDN} = +3.0V$, $P_{RFIN} = -25dBm$, $P_{LO} = -5dBm$, $T_A = +25^\circ C$, unless otherwise noted.)



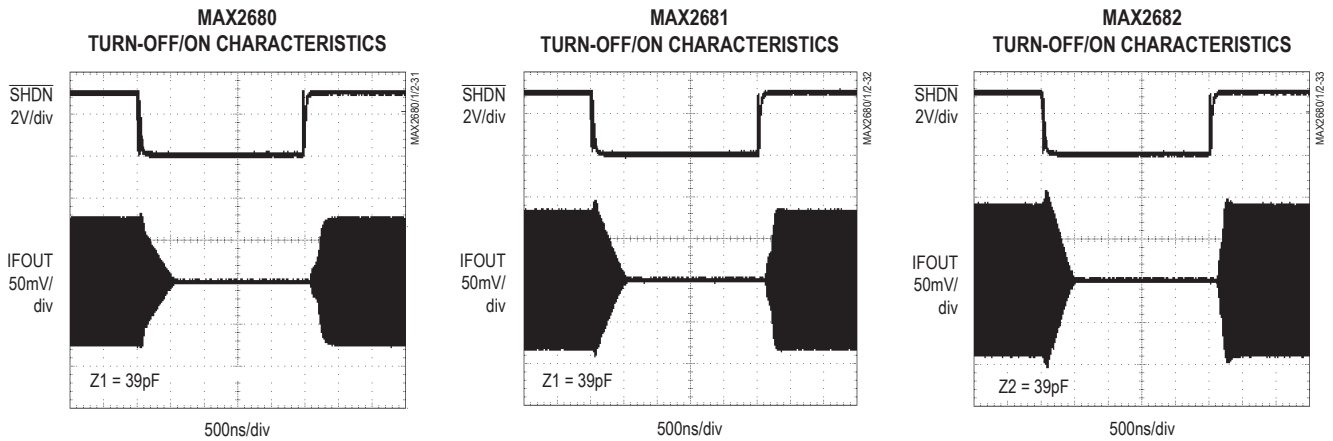
Typical Operating Characteristics (continued)

(Typical Operating Circuit, $V_{CC} = \overline{\text{SHDN}} = +3.0\text{V}$, $P_{\text{RFIN}} = -25\text{dBm}$, $P_{\text{LO}} = -5\text{dBm}$, $T_A = +25^\circ\text{C}$, unless otherwise noted.)

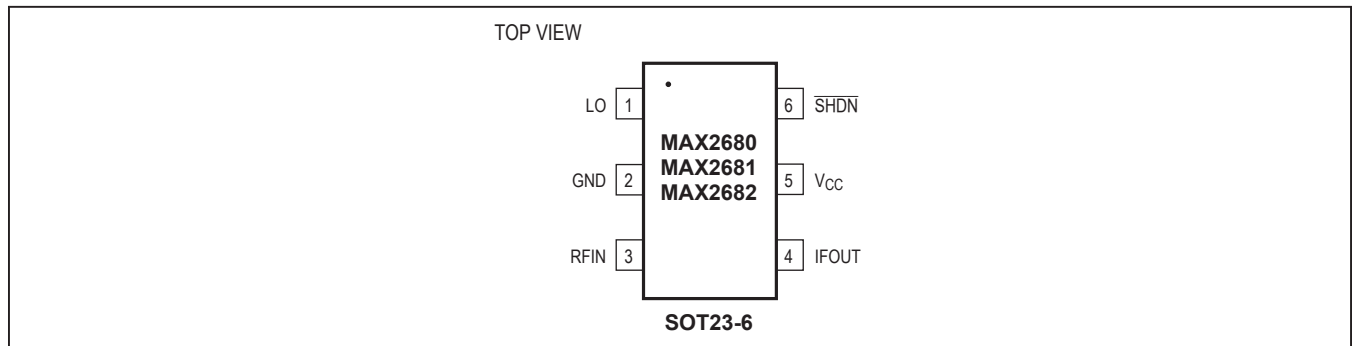


Typical Operating Characteristics (continued)

(Typical Operating Circuit, $V_{CC} = \overline{\text{SHDN}} = +3.0\text{V}$, $P_{\text{RFIN}} = -25\text{dBm}$, $P_{\text{LO}} = -5\text{dBm}$, $T_A = +25^\circ\text{C}$, unless otherwise noted.)



Pin Configuration



Pin Description

| PIN | NAME | FUNCTION |
|-----|--------------------------|--|
| 1 | LO | Local-Oscillator Input. Apply a local-oscillator signal with an amplitude of -10dBm to 0 (50Ω source). AC-couple this pin to the oscillator with a DC-blocking capacitor. Nominal DC voltage is $V_{CC} - 0.4\text{V}$. |
| 2 | GND | Mixer Ground. Connect to the ground plane with a low-inductance connection. |
| 3 | RFIN | Radio Frequency Input. AC-couple to this pin with a DC-blocking capacitor. Nominal DC voltage is 1.5V. See the <i>Applications Information</i> section for details on impedance matching. |
| 4 | IFOUT | Intermediate Frequency Output. Open-collector output requires an inductor to V_{CC} . AC-couple to this pin with a DC-blocking capacitor. See the <i>Applications Information</i> section for details on impedance matching. |
| 5 | V_{CC} | Supply Voltage Input, +2.7V to +5.5V. Bypass with a capacitor to the ground plane. Capacitor value depends upon desired operating frequency. |
| 6 | $\overline{\text{SHDN}}$ | Active-Low Shutdown. Drive low to disable all device functions and reduce the supply current to less than 5μA. For normal operation, drive high or connect to V_{CC} . |

Detailed Description

The MAX2680/MAX2681/MAX2682 are 400MHz to 2.5GHz, silicon-germanium, double-balanced downconverter mixers. They are designed to provide optimum linearity performance for a specified supply current. They consist of a double-balanced Gilbert-cell mixer with single-ended RF, LO, and IF port connections. An on-chip bias cell provides a low-power shutdown feature. Consult the [Selector Guide](#) for device features and comparison.

Applications Information

Local-Oscillator (LO) Input

The LO input is a single-ended broadband port with a typical input VSWR of better than 2.0:1 from 400MHz to 2.5GHz. The LO signal is mixed with the RF input signal, and the resulting downconverted output appears at IFOUT. AC-couple LO with a capacitor. Drive the LO port with a signal ranging from -10dBm to 0 (50Ω source).

RF Input

The RF input frequency range is 400MHz to 2.5GHz. The RF input requires an impedance-matching network as well as a DC-blocking capacitor that can be part of the matching network. Consult Tables 1 and 2, as well as the RF Port Impedance vs. RF Frequency graph in the [Typical Operating Characteristics](#) section for information on matching.

Table 1. RFIN Port Impedance

| PART | FREQUENCY | | | |
|---------|-----------|---------|---------|---------|
| | 400MHz | 900MHz | 1950MHz | 2450MHz |
| MAX2680 | 179-j356 | 54-j179 | 32-j94 | 33-j73 |
| MAX2681 | 209-j332 | 75-j188 | 34-j108 | 33-j86 |
| MAX2682 | 206-j306 | 78-j182 | 34-j106 | 29-j86 |

Table 2. RF Input Impedance-Matching Component Values

| MATCHING COMPONENTS | FREQUENCY | | | | | | | | | | | |
|---------------------|-----------|---------|----------|----------|---------|---------|----------|----------|---------|---------|----------|----------|
| | MAX2680 | | | | MAX2681 | | | | MAX2682 | | | |
| | 400 MHz | 900 MHz | 1950 MHz | 2450 MHz | 400 MHz | 900 MHz | 1950 MHz | 2450 MHz | 400 MHz | 900 MHz | 1950 MHz | 2450 MHz |
| Z1 | 86nH | 270pF | 1.5pF | Short | 68nH | 270pF | 1.5pF | Short | 68nH | 1.5pF | Short | Short |
| Z2 | 270pF | 22nH | 270pF | 270pF | 270pF | 18nH | 270pF | 270pF | 270pF | 270pF | 270pF | 270pF |
| Z3 | Open | Open | 1.8nH | 1.8nH | 0.5pF | Open | 1.8nH | 2.2nH | 0.5pF | 10nH | 2.2nH | 1.2nH |

Note: Z1, Z2, and Z3 are found in the *Typical Operating Circuit*.

IF Output

The IF output frequency range extends from 10MHz to 500MHz. IFOUT is a high-impedance, open-collector output that requires an external inductor to V_{CC} for proper biasing. For optimum performance, the IF port requires an impedance-matching network. The configuration and values for the matching network is dependent upon the frequency and desired output impedance. For assistance in choosing components for optimal performance, see [Table 3](#) and [Table 4](#) as well as the IF Port Impedance vs. IF Frequency graph in the [Typical Operating Characteristics](#) section.

Power-Supply and SHDN Bypassing

Proper attention to voltage supply bypassing is essential for high-frequency RF circuit stability. Bypass V_{CC} with a 10μF capacitor in parallel with a 1000pF capacitor. Use separate vias to the ground plane for each of the bypass capacitors and minimize trace length to reduce inductance. Use separate vias to the ground plane for each ground pin. Use low-inductance ground connections.

Decouple SHDN with a 1000pF capacitor to ground to minimize noise on the internal bias cell. Use a series resistor (typically 100Ω) to reduce coupling of high-frequency signals into the SHDN pin.

Layout Issues

A well-designed PC board is an essential part of an RF circuit. For best performance, pay attention to power-supply issues as well as to the layout of the RFIN and IFOUT impedance-matching network.

Table 3. IFOUT Port Impedance

| PART | FREQUENCY | | |
|---------|-----------|----------|----------|
| | 45MHz | 70MHz | 240MHz |
| MAX2680 | 960-j372 | 803-j785 | 186-j397 |
| MAX2681 | 934-j373 | 746-j526 | 161-j375 |
| MAX2682 | 670-j216 | 578-j299 | 175-j296 |

Table 4. IF Output Impedance-Matching Components

| MATCHING COMPONENT | FREQUENCY | | |
|--------------------|-----------|-------|--------|
| | 45MHz | 70MHz | 240MHz |
| L1 | 390nH | 330nH | 82nH |
| C2 | 39pF | 15pF | 3pF |
| R1 | 250Ω | Open | Open |

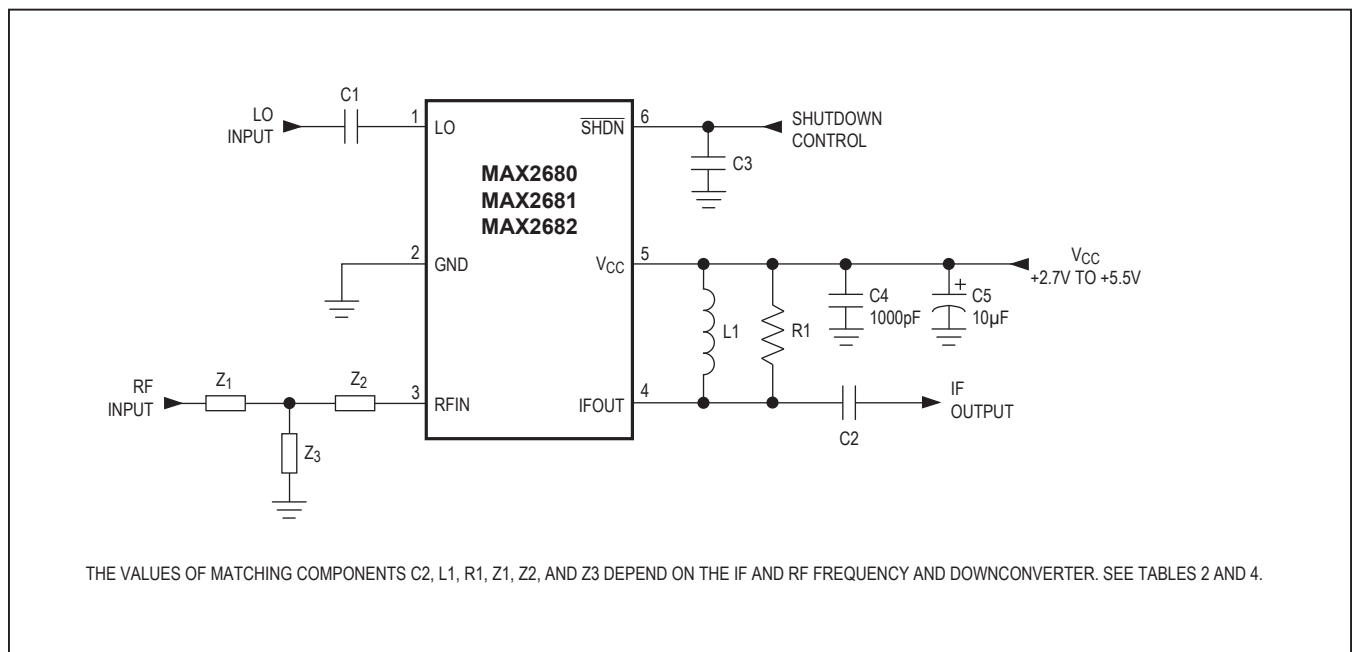
Power-Supply Layout

To minimize coupling between different sections of the IC, the ideal power-supply layout is a star configuration with a large decoupling capacitor at a central V_{CC} node. The V_{CC} traces branch out from this central node, each going to a separate V_{CC} node on the PC board. At the end of each trace is a bypass capacitor that has low ESR at the RF frequency of operation. This arrangement provides local decoupling at the V_{CC} pin. At high frequencies, any signal leaking out of one supply pin sees a relatively high impedance (formed by the V_{CC} trace inductance) to the central V_{CC} node, and an even higher impedance to any other supply pin, as well as a low impedance to ground through the bypass capacitor.

Impedance-Matching Network Layout

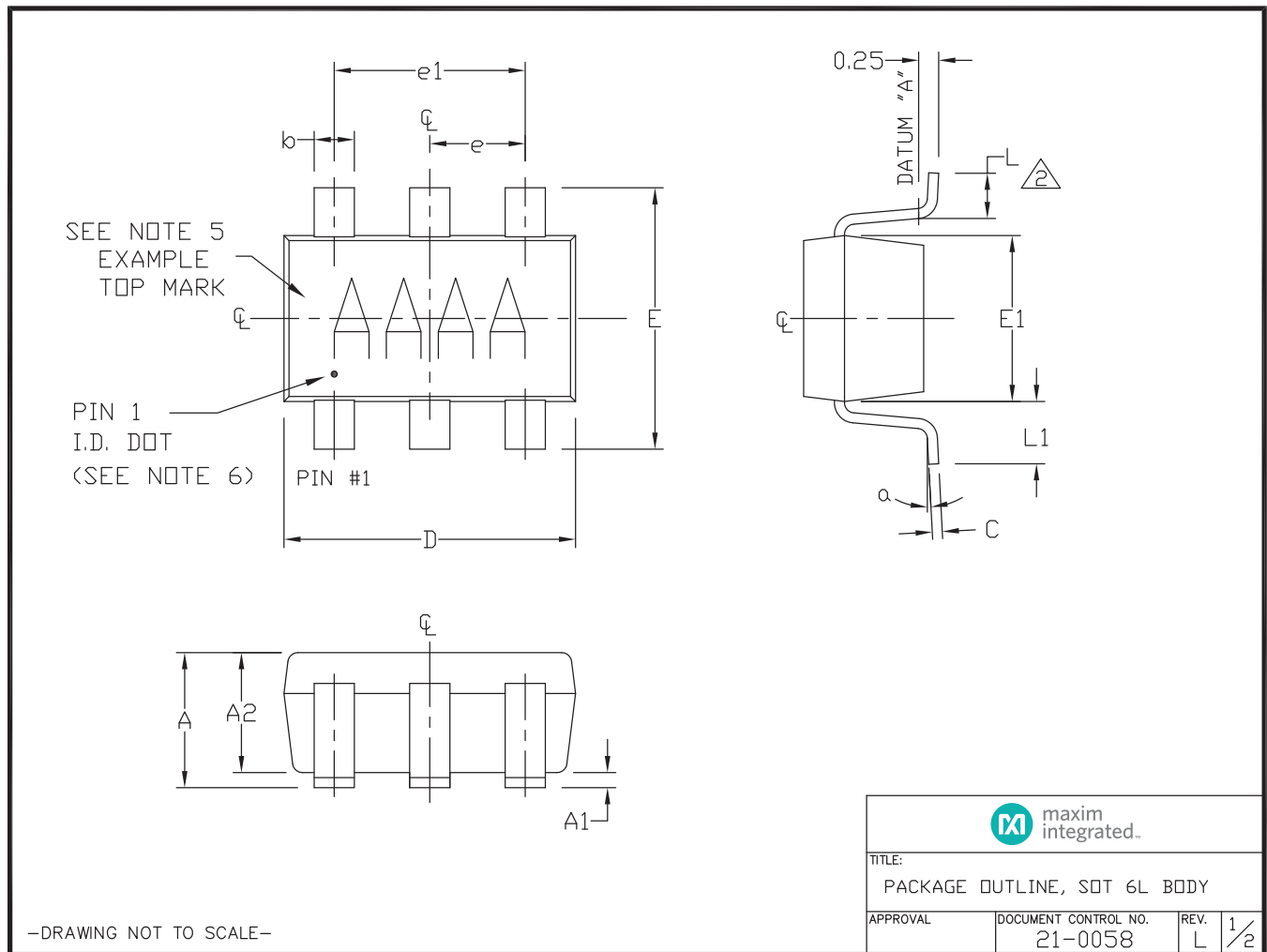
The RFIN and IFOUT impedance-matching networks are very sensitive to layout-related parasitics. To minimize parasitic inductance, keep all traces short and place components as close as possible to the chip. To minimize parasitic capacitance, use cutouts in the ground plane (and any other plane) below the matching network components. However, avoid cutouts that are larger than necessary since they act as aperture antennas.

Typical Operating Circuit



Package Information


For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.



Package Information (continued)

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

NOTES:


1. ALL DIMENSIONS ARE IN MILLIMETERS.
2.  FOOT LENGTH MEASURED AT INTERCEPT POINT BETWEEN DATUM A & LEAD SURFACE.
3. PACKAGE OUTLINE EXCLUSIVE OF MOLD FLASH & METAL BURR. MOLD FLASH, PROTRUSION OR METAL BURR SHOULD NOT EXCEED 0.25mm.
4. PACKAGE OUTLINE INCLUSIVE OF SOLDER PLATING.
5. PIN 1 IS LOWER LEFT PIN WHEN READING TOP MARK FROM LEFT TO RIGHT. (SEE EXAMPLE TOP MARK)
6. PIN 1 I.D. DOT IS 0.3mm ϕ MIN. LOCATED ABOVE PIN 1.
7. MEETS JEDEC MO178, VARIATION AB.
8. SOLDER THICKNESS MEASURED AT FLAT SECTION OF LEAD BETWEEN 0.08mm AND 0.15mm FROM LEAD TIP.
9. LEAD TO BE COPLANAR WITHIN 0.1mm.
10. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.
11. MARKING IS FOR PACKAGE ORIENTATION REFERENCE ONLY.
12. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PbFREE (+) PKG. CODES.

| SYMBOL | MIN | NOMINAL | MAX |
|----------|-----------|---------|------|
| A | 0.90 | 1.25 | 1.45 |
| A1 | 0.00 | 0.05 | 0.15 |
| A2 | 0.90 | 1.10 | 1.30 |
| b | 0.35 | 0.40 | 0.50 |
| C | 0.08 | 0.15 | 0.20 |
| D | 2.80 | 2.90 | 3.00 |
| E | 2.60 | 2.80 | 3.00 |
| E1 | 1.50 | 1.625 | 1.75 |
| L | 0.35 | 0.45 | 0.60 |
| L1 | 0.60 REF. | | |
| e1 | 1.90 BSC. | | |
| e | 0.95 BSC. | | |
| α | 0° | 2.5° | 10° |

PKG CODES:
U6-1, U6-2, U6-4, U6CN-2,
U6SN-1, U6F-6, U6FH-6; U6FH-7

** U6FH-7 TO BE USED FOR NP42 PARTS ONLY.

-DRAWING NOT TO SCALE-

| | | | |
|---|---------------------------------|-----------|-----|
|  maxim integrated. | | | |
| TITLE: PACKAGE OUTLINE, SOT 6L BODY | | | |
| APPROVAL | DOCUMENT CONTROL NO. 21-0058 | REV. L | 2/2 |

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