

## Low Power Universal SLIC Family

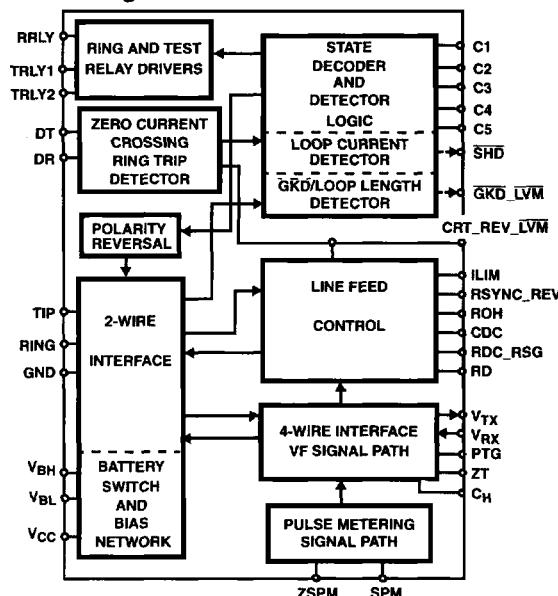
The UniSLIC14 is a family of Ultra Low Power SLICs. The feature set and common pinouts of the UniSLIC14 family positions it as a universal solution for: Plain Old Telephone Service (POTS), PBX, Central Office, Loop Carrier, Fiber in the Loop, ISDN-TA and NT1+, Pairgain and Wireless Local Loop.

The UniSLIC14 family achieves its ultra low power operation through: It's automatic single and dual battery operation (based on line length), low power standby state and battery tracking saturation guard to ensure the maximum loop coverage on the lowest battery voltage. This architecture is ideal for power critical applications such as ISDN NT1+, Pairgain and Wireless local loop products.

The UniSLIC14 family has many user programmable features. This family of SLICs delivers a low noise, low component count solution for Central Office and Loop Carrier universal voice grade designs. The product family integrates advanced pulse metering, test and signaling capabilities, and zero crossing ring control.

The UniSLIC14 family is designed in the Harris "Latch" free Bonded Wafer process. This process dielectrically isolates the active circuitry to eliminate any leakage paths as found in our competitor's JI process. This makes the UniSLIC14 family compliant with "hot plug" requirements and operation in harsh outdoor environments.

## Block Diagram



## Features

- Ultra Low Active Power (OHT) < 60mW
- Low Standby Power < 25mW
- Single/Dual Battery Operation
- Automatic Silent Battery Switching
- Thermal Management/Shutdown
- Battery Tracking Saturation Guard
- Single 5V Supply
- Zero Crossing Ring Control
  - Zero Voltage On/Zero Current Off
- Tip/Ring Disconnect
- Pulse Metering Capability
- 4 Wire Loopback
- Programmable Constant Current Feed
- Programmable Resistive Feed
- Programmable Loop Detect Threshold
- Programmable On-Hook and Off-Hook Overheads
- Programmable Overhead for Pulse Metering
- Programmable Polarity Reversal Time
- Selectable Transmit Gain 0dB/-6dB
- 2 Wire Impedance Set by Single Network
- Loop and Ground Key Detectors
- On-Hook Transmission
- Common Pinout
- HC55121
  - Polarity Reversal
- HC55130
  - -63dB Longitudinal Balance
- HC55140
  - Polarity Reversal
  - Ground Start
  - Line Voltage Measurement
  - 2 Wire Loopback
  - -63dB Longitudinal Balance
- HC55142
  - Polarity Reversal
  - Ground Start
  - Line Voltage Measurement
  - 2.2V<sub>RMS</sub> Pulse Metering
  - 2 Wire Loopback
- HC55150
  - Polarity Reversal
  - Line Voltage Measurement
  - 2.2V<sub>RMS</sub> Pulse Metering
  - 2 Wire Loopback

## Related Literature

- AN9832, User's Guide for Development Board

***Ordering Information (PLCC Package Only)***

PART NUMBER	MAX LOOP CURRENT (mA)	POLARITY REVERSAL	GROUND START	GROUND KEY	LINE VOLTAGE MEASUREMENT †	PULSE METERING	2 TEST RELAY DRIVERS	2 WIRE LOOPBACK †	LONGITUDINAL BALANCE	TEMP RANGE (°C)	PKG. NO.
HC55120CM	30			•					53dB	0 to 70	M28.3 N28.45
HC55121IM	30	•		•		•		•	53dB	-40 to 85	M28.3 N28.45
HC55130IM	45						HC55131IM		63dB	-40 to 85	M28.3 N28.45
HC55140IM	45	•	•	•	•		HC55141IM	•	63dB	-40 to 85	M28.3 N28.45
HC55142IM	45	•	•	•	•	•	HC55143IM	•	53dB	-40 to 85	M28.3 N28.45
HC55150CM	45	•			•	•	HO5515ICM	•	55dB	0 to 70	M28.3 N28.45
HC5514XEV1	Evaluation board										

† Available by placing SLIC in Test mode.

***Device Operating Modes***

C3	C2	C1	DESCRIPTION	HC55120	HC55121	HC55130/1	HC55140/1	HC55142/3	HC55150/1
0	0	0	Disconnect Tip and Ring.	•	•	•	•	•	•
0	0	1	Ringing	•	•	•	•	•	•
0	1	0	Forward Active	•	•	•	•	•	•
0	1	1	Test Forward Active 2 Wire Loopback and Line Voltage Measurement		•		•	•	•
1	0	0	Tip Open Ground Start				•	•	
1	0	1	Low Power Standby	•	•	•	•	•	•
1	1	0	Reverse Active		•		•	•	•
1	1	1	Test Reverse Active Line Voltage Measurement				•	•	•

**Absolute Maximum Ratings  $T_A = 25^\circ\text{C}$** 

Temperature, Humidity	
Storage Temperature Range	-65°C to 150°C
Operating Temperature Range	-40°C to 110°C
Operating Junction Temperature Range	-40°C to 150°C
Power Supply (-40°C $\leq T_A \leq +85^\circ\text{C}$ )	
Supply Voltage V <sub>CC</sub> to GND	.04V to 7V
Supply Voltage V <sub>BL</sub> to GND	-V <sub>BL</sub> to 0.4V
Supply Voltage V <sub>BH</sub> to GND, Continuous	-75V to 0.4V
Supply Voltage V <sub>BH</sub> to GND, 10ms	-80V to 0.4V
Relay Driver	
Ring Relay Supply Voltage	.0V to 14V
Ring Relay Current	.50mA
Digital Inputs, Outputs (C1, C2, C3, C4, C5, SHD, GKD_LVM)	
Input Voltage	-0.4V to V <sub>CC</sub>
Output Voltage (SHD, GKD_LVM Not Active)	-0.4V to V <sub>CC</sub>
Output Current (SHD, GKD_LVM)	.5mA
ESD Rating	.500V
Gate Count	.543 Transistors, 51 Diodes
Tipx and Ringx Terminals (-40°C $\leq T_A \leq 85^\circ\text{C}$ )	
Tipx or Ringx Current	-100mA to 100mA

**CAUTION:** Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

**NOTE:**

1.  $\theta_{JA}$  is measured with the component mounted on an evaluation PC board in free air.

**Typical Operating Conditions**

These represent the conditions under which the part was developed and are suggested as guidelines.

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Ambient Temperature	HC55120, HC55150/1	0	-	70	°C
	HC55121, HC55130/1, HC55140/1, HC55142/3	-40	-	85	°C
V <sub>BH</sub> with Respect to GND		-58	-	-8	V
V <sub>BL</sub> with Respect to GND		V <sub>BH</sub>	-	-8	V
V <sub>CC</sub> with Respect to GND		4.75	-	5.25	V

**Electrical Specifications**

$T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $V_{CC} = +5V \pm 5\%$ ,  $V_{BH} = -48V$ ,  $V_{BL} = \text{No connect}$ ,  $PTG = \text{Open}$ ,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_T = 120\text{k}\Omega$ ,  $R_{IM} = 38.3\text{k}\Omega$ ,  $R_D = 58.8\text{k}\Omega$ ,  $R_{SG} = 20\text{k}\Omega$ ,  $R_{OH} = 50\text{k}\Omega$ ,  $C_{dI} = 0.47\mu\text{F}$ ,  $C_{DC} = 1.0\mu\text{F}$ ,  $C_{TR/REV} = 0.47\mu\text{F}$ ,  $GND = 0V$ ,  $RL = 600\Omega$ . Unless Otherwise Specified. (•) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part.

PARAMETER	TEST CONDITIONS	MIN	MAX	UNITS	HC55120	HC55121	HC55130/1	HC55140/1	HC55142/3	HC55150/1
<b>2-WIRE PORT</b>										
Overload Level, Off Hook Forward and Reverse	1% THD, $ IdCMET  \geq 18\text{mA}$ (Note 2, Figure 1)	3.2	-	$V_{PEAK}$	Forward Only	•	Forward Only	•	•	•
Overload Level, On Hook Forward and Reverse	1% THD, $ IdCMET  \leq 5\text{mA}$ (Note 3, Figure 1)	1.3	-	$V_{PEAK}$	Forward Only	•	Forward Only	•	•	•
Input Impedance (into Tip and Ring)	-	$Z_T/200$	-	$\Omega$	•	•	•	•	•	•
Longitudinal Impedance (Tip, Ring) Forward and Reverse	$0 < f < 100\text{Hz}$ (Note 4, Figure 2)	-	0	$\Omega/\text{Wire}$	Forward Only	•	Forward Only	•	•	•
<b>LONGITUDINAL CURRENT LIMIT (TIP, RING)</b>										
On-Hook, Off-Hook (Active), $RL = 736\Omega$ Forward and Reverse	No False Detections, (Loop Current, LB > 45dB (Note 5, Figure 3A))	28	-	$\text{mA RMS/Wire}$	Forward Only	•	Forward Only	•	•	•
On-Hook (Low Power Standby), $RL = \infty$ , $R_D = 58.8\text{k}\Omega$	No False Detections (Loop Current) (Note 6, Figure 3B)	-	8.5	$\text{mA PEAK/Wire}$	•	•	•	•	•	•

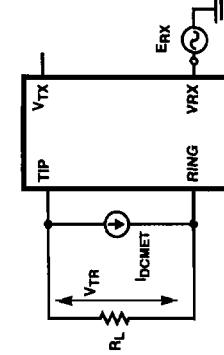


FIGURE 1. OVERLOAD LEVEL (OFF HOOK, ON HOOK)

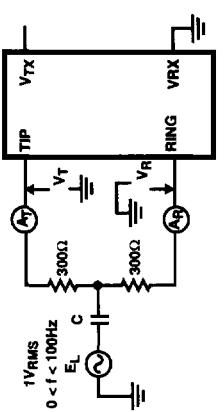


FIGURE 2. LONGITUDINAL IMPEDANCE

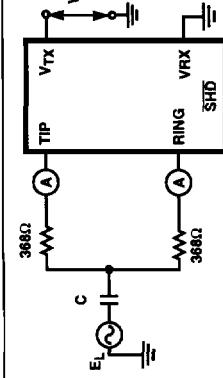


FIGURE 3A. LONGITUDINAL CURRENT LIMIT OFF-HOOK (ACTIVE)

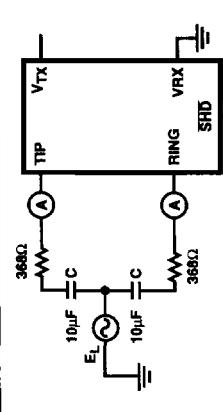


FIGURE 3B. LONGITUDINAL CURRENT LIMIT ON-HOOK (STANDBY)

**Electrical Specifications**  $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ ,  $V_{CC} = +5\text{V} \pm 5\%$ ,  $V_{BH} = -48\text{V}$ ,  $V_{BL} = \text{No connect}$ ,  $\text{PTG} = \text{Open}$ ,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_T = 120\text{k}\Omega$ ,  $R_{IM} = 38.3\text{k}\Omega$ ,  $R_D = 58.8\text{k}\Omega$ ,  $R_{SG} = 20\text{k}\Omega$ ,  $R_{OH} = 50\text{k}\Omega$ ,  $C_H = 0.47\mu\text{F}$ ,  $CDC = 1.0\text{pF}$ ,  $\text{CR/REV} = 0.47\mu\text{fF}$ ,  $\text{GND} = 0\text{V}$ ,  $R_L = 600\Omega$  Unless Otherwise Specified. (\*) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part. (Continued)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNITS	HC55120		HC55121		HC55130/1		HC55140/1		HC55142/3		HC55150/1	
					HC55120	HC55121	HC55130/1	HC55140/1	HC55142/3	HC55150/1						
<b>OFF-HOOK LONGITUDINAL BALANCE</b>																
Longitudinal to Metallic (Note 7) Forward and Reverse	IEEE 455 - 1985, $R_{LR}, R_{LT} = 368\Omega$ , Normal Polarity: 0.2kHz < f < 1.0kHz, 0°C to 70°C 1.0kHz < f < 3.4kHz, 0°C to 70°C 0.2kHz < f < 1.0kHz, -40°C to 85°C 1.0kHz < f < 3.4kHz, -40°C to 85°C Reverse Polarity 0.2kHz < f < 3.4kHz, (Figure 4)	-	-	dB	53	NA	NA	NA	NA	NA	NA	NA	NA	NA	NA	NA
Longitudinal to Metallic (Note 7) Forward and Reverse	$R_{LR}, R_{LT} = 300\Omega$ , Normal Polarity: 0.2kHz < f < 1.0kHz, 0°C to 70°C 1.0kHz < f < 3.4kHz, 0°C to 70°C 0.2kHz < f < 1.0kHz, -40°C to 85°C 1.0kHz < f < 3.4kHz, -40°C to 85°C Reverse Polarity 0.2kHz < f < 3.4kHz, (Figure 4)	-	-	dB	53	NA	NA	NA	NA	NA	NA	NA	NA	NA	NA	NA
Longitudinal to 4-Wire (Note 9) (Forward and Reverse)	Normal Polarity: 0.2kHz < f < 1.0kHz, 0°C to 70°C 1.0kHz < f < 3.4kHz, 0°C to 70°C 0.2kHz < f < 1.0kHz, -40°C to 85°C 1.0kHz < f < 3.4kHz, -40°C to 85°C Reverse Polarity 0.2kHz < f < 3.4kHz, (Figure 4)	-	-	dB	53	NA	NA	NA	NA	NA	NA	NA	NA	NA	NA	NA
Metallic to Longitudinal (Note 10) Forward and Reverse	FCC Part 68, Para 6B.310 (Note 8) 0.2kHz < f < 3.4kHz, (Figure 5)	40	50	dB	Forward Only	•	Forward Only	•	Forward Only	•	Forward Only	•	Forward Only	•	Forward Only	•
4-Wire to Longitudinal (Note 11) Forward and Reverse	0.2kHz < f < 3.4kHz, (Figure 5)	40	-	dB	Forward Only	•	Forward Only	•	Forward Only	•	Forward Only	•	Forward Only	•	Forward Only	•

**Electrical Specifications**  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $V_{CC} = +5\text{V} \pm 5\%$ ,  $V_{BH} = -48\text{V}$ ,  $V_{BL} = \text{No connect}$ ,  $\text{PTG} = \text{Open}$ ,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_T = 120\text{k}\Omega$ ,  $R_{LM} = 38.3\text{k}\Omega$ ,  $R_D = 58.8\text{k}\Omega$ ,  $R_{DC\_RSR} = 20\text{k}\Omega$ ,  $R_{OH} = 50\text{k}\Omega$ ,  $C_{TH} = 0.47\mu\text{F}$ ,  $C_{DC} = 1.0\mu\text{F}$ ,  $C_{RT/REV} = 0.47\mu\text{F}$ ,  $\text{GND} = 0\text{V}$ ,  $R_L = 600\Omega$ . Unless Otherwise Specified, ( $\bullet$ ) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part. (Continued)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNITS	HC55120	HC55121	HC55130/1	HC55140/1	HC55142/3	HC55150/1
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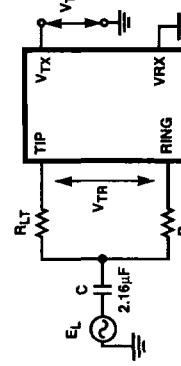


FIGURE 4. LONGITUDINAL TO METALLIC AND LONGITUDINAL TO 4-WIRE BALANCE

FIGURE 5. METALLIC TO LONGITUDINAL AND 4-WIRE TO LONGITUDINAL BALANCE

2-Wire Return Loss Forward and Reverse	0.2kHz to 1.0kHz (Note 12, Figure 6) 1.0kHz to 3.4kHz (Note 12, Figure 6)	30	35	-	dB	Forward Only	•	Forward Only	•	•
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#### TIP IDLE VOLTAGE (User Programmable)

TIP Idle Voltage Forward and Reverse	Active, $I_L < 5\text{mA}$	-	-2.0	-	V	Forward Only	•	Forward Only	•	•
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#### RING IDLE VOLTAGE (User Programmable)

RINGX Idle Voltage Forward and Reverse	Active, $I_L < 5\text{mA}$ Tip open, $I_L < 5\text{mA}$	-	$V_{BH}+2.5$	-	V	Forward Only	•	Forward Only	•	•
$V_{TR}$ Forward and Reverse	Active, $I_L < 5\text{mA}$	-	$V_{BH}+4.5$	-	V	Forward Only	•	Forward Only	•	•
$V_{TR(OH)}$ Pulse Metering Forward and Reverse	Active, $I_L \geq 8.5\text{mA}$ , $R_{OH} = 50\text{k}\Omega$	-	$V_{BH}+8.9$	-	V	NA	•	NA	NA	•

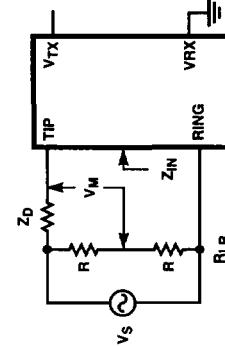


FIGURE 6. TWO-WIRE RETURN LOSS

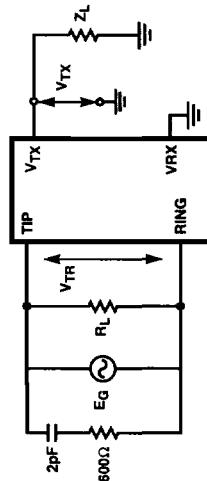


FIGURE 7. OVERLOAD LEVEL (4-WIRE TRANSMIT PORT), OUTPUT OFFSET VOLTAGE AND HARMONIC DISTORTION

**Electrical Specifications**

$T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $V_{CC} = +5\text{V} \pm 5\%$ ,  $V_{BH} = -48\text{V}$ ,  $V_{BL} = \text{No connect}$ ,  $PTG = \text{Open}$ ,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_T = 120\text{k}\Omega$ ,  $R_{LIM} = 38.3\text{k}\Omega$ ,  $R_D = 58.8\text{k}\Omega$ ,  $RDC\_RSG = 20\text{k}\Omega$ ,  $R_{OH} = 50\text{k}\Omega$ ,  $C_H = 0.47\mu\text{F}$ ,  $C_{DC} = 1.0\mu\text{F}$ ,  $C_{TR/REV} = 0.47\mu\text{F}$ ,  $GND = 0\text{V}$ ,  $R_L = 600\Omega$ . Unless Otherwise Specified. (\*) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part. (Continued)

PARAMETER	TEST CONDITIONS	MIN	MAX	TYP	UNITS	HC55120	HC55121	HC55130/1	HC55140/1	HC55142/3	HC55150/1
<b>4-WIRE TRANSMIT PORT (<math>V_{TX}</math>)</b>											
Overload Level, Off-Hook ( $I_L \geq 18\text{mA}$ )	$ Z_L > 20\text{k}\Omega$ , IL 1% THD	3.2	-	-	VPEAK	Forward Only	•	Forward Only	•	•	•
Forward and Reverse	$ Z_L > 20\text{k}\Omega$ , 1% THD	1.3	-	-	VPEAK	Forward Only	•	Forward Only	•	•	•
Overload Level, On Hook ( $I_L \leq 5\text{mA}$ )	$ Z_L > 20\text{k}\Omega$ , 1% THD	-	-	-	VPEAK	Forward Only	•	Forward Only	•	•	•
Forward and Reverse	$ Z_L = \infty$ , (Note 14, Figure 7)	-	-	-	mV	Forward Only	•	Forward Only	•	•	•
Output Offset Voltage	$E_G = 0$ , $Z_L = \infty$ , (Note 15, Figure 7)	-100	-100	100	mV	Forward Only	•	Forward Only	•	•	•
Forward and Reverse	$0.2\text{kHz} < f < 0.3\text{kHz}$	-	0.1	1	Ω	•	•	•	•	•	•
Output Impedance (Guaranteed by Design)	$0.2\text{kHz} < f < 3.4\text{kHz}$	-	500	600	kΩ	•	•	•	•	•	•
<b>4-WIRE RECEIVE PORT (RX)</b>											
VRX Input Impedance (Guaranteed by Design)	$0.2\text{kHz} < f < 3.4\text{kHz}$	-	500	600	kΩ	•	•	•	•	•	•
<b>FREQUENCY RESPONSE (OFF-HOOK)</b>											
2-Wire to 4-Wire Forward and Reverse	Relative to 0dBm at 1.0kHz, $E_{RX} = 0\text{V}$ $0.3\text{kHz} < f < 3.4\text{kHz}$ $f = 8.0\text{kHz}, 12\text{kHz}, 16\text{kHz}$ (Note 16, Figure 8)	-0.2	-	0.1	dB	Forward Only	•	Forward Only	•	•	•
4-Wire to 2-Wire Forward and Reverse	Relative to 0dBm at 1.0kHz, $E_G = 0\text{V}$ $0.3\text{kHz} < f < 3.4\text{kHz}$ $f = 8.0\text{kHz}, 12\text{kHz}$ $f = 16\text{kHz}$ (Note 17, Figure 8)	-0.2	-	0.1	dB	Forward Only	•	Forward Only	•	•	•
4-Wire to 4-Wire Forward and Reverse	Relative to 0dBm at 1.0kHz, $E_G = 0\text{V}$ $0.3\text{kHz} < f < 3.4\text{kHz}$ (Note 18, Figure 8)	-0.2	-	0.1	dB	Forward Only	•	Forward Only	•	•	•

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PARAMETER	TEST CONDITIONS	MIN	MAX	TYP	UNITS	HC55120	HC55121	HC55130/1	HC55140/1	HC55142/3	HC55150/1
<b>4-WIRE TRANSMIT PORT (<math>V_{TX}</math>)</b>											
Overload Level, Off-Hook ( $I_L \geq 18\text{mA}$ )	$ Z_L > 20\text{k}\Omega$ , IL 1% THD	3.2	-	-	VPEAK	Forward Only	•	Forward Only	•	•	•
Forward and Reverse	$ Z_L > 20\text{k}\Omega$ , 1% THD	1.3	-	-	VPEAK	Forward Only	•	Forward Only	•	•	•
Overload Level, On Hook ( $I_L \leq 5\text{mA}$ )	$ Z_L > 20\text{k}\Omega$ , 1% THD	-	-	-	VPEAK	Forward Only	•	Forward Only	•	•	•
Forward and Reverse	$E_G = 0$ , $Z_L = \infty$ , (Note 15, Figure 7)	-100	-100	100	mV	Forward Only	•	Forward Only	•	•	•
Output Offset Voltage	$0.2\text{kHz} < f < 0.3\text{kHz}$	-	0.1	1	Ω	•	•	•	•	•	•
Forward and Reverse	$0.2\text{kHz} < f < 3.4\text{kHz}$	-	500	600	kΩ	•	•	•	•	•	•
Output Impedance (Guaranteed by Design)	$0.2\text{kHz} < f < 3.4\text{kHz}$	-	500	600	kΩ	•	•	•	•	•	•
<b>4-WIRE RECEIVE PORT (RX)</b>											
VRX Input Impedance (Guaranteed by Design)	$0.2\text{kHz} < f < 3.4\text{kHz}$	-	500	600	kΩ	•	•	•	•	•	•
<b>FREQUENCY RESPONSE (OFF-HOOK)</b>											
2-Wire to 4-Wire Forward and Reverse	Relative to 0dBm at 1.0kHz, $E_{RX} = 0\text{V}$ $0.3\text{kHz} < f < 3.4\text{kHz}$ $f = 8.0\text{kHz}, 12\text{kHz}, 16\text{kHz}$ (Note 16, Figure 8)	-0.2	-	0.1	dB	Forward Only	•	Forward Only	•	•	•
4-Wire to 2-Wire Forward and Reverse	Relative to 0dBm at 1.0kHz, $E_G = 0\text{V}$ $0.3\text{kHz} < f < 3.4\text{kHz}$ $f = 8.0\text{kHz}, 12\text{kHz}$ $f = 16\text{kHz}$ (Note 17, Figure 8)	-0.2	-	0.1	dB	Forward Only	•	Forward Only	•	•	•
4-Wire to 4-Wire Forward and Reverse	Relative to 0dBm at 1.0kHz, $E_G = 0\text{V}$ $0.3\text{kHz} < f < 3.4\text{kHz}$ (Note 18, Figure 8)	-0.2	-	0.1	dB	Forward Only	•	Forward Only	•	•	•

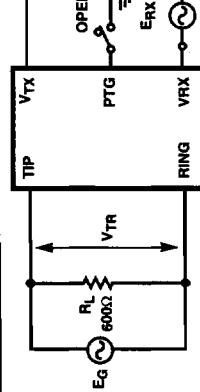


FIGURE 8. FREQUENCY RESPONSE, INSERTION LOSS, GAIN TRACKING AND HARMONIC DISTORTION

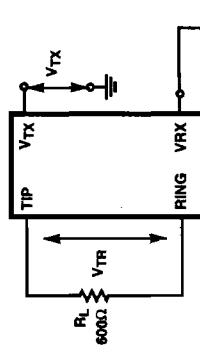
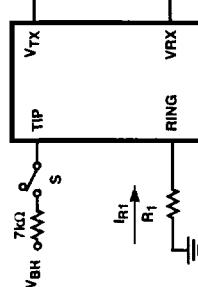
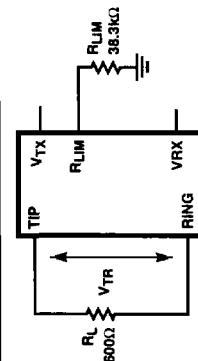


FIGURE 9. IDLE CHANNEL NOISE

**Electrical Specifications**  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $V_{CC} = +5V \pm 5\%$ ,  $V_{BH} = -48V$ ,  $V_{BL} = \text{No connect}$ ,  $\text{PTG} = \text{Open}$ ,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_T = 0\Omega$ ,  $R_{DC\_RSG} = 20\text{k}\Omega$ ,  $R_{OH} = 50\text{k}\Omega$ ,  $C_H = 0.47\mu\text{F}$ ,  $C_D = 1.0\mu\text{F}$ ,  $C_{RT/REV} = 0.47\mu\text{F}$ ,  $\text{GND} = 0V$ ,  $RL = 600\Omega$ . Unless Otherwise Specified. (•) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part. (Continued)

PARAMETER	TEST CONDITIONS	MIN	MAX	TYPE	UNITS	HC55120	HC55121	HC55120/1	HC55121/1	HC55142/3	HC55150/1
<b>INSERTION LOSS</b>											
2-Wire to 4-Wire Forward and Reverse	0dBm, 1kHz PTG = Open (Note 19, Figure 8)	-0.2	-	0.2	dB	Forward Only	•	Forward Only	•	•	•
4-Wire to 2-Wire Forward and Reverse	PTG = GND (Note 20, (Figure 8)) 0dBm, 1kHz (Note 21, Figure 8)	-6.22	-6.02	-5.82	dB	Forward Only	•	Forward Only	•	•	•
<b>GAIN TRACKING (Ref = -10dBm, at 1.0kHz)</b>											
2-Wire to 4-Wire Forward and Reverse	-40dBm to -3dBm (Note 22, Figure 8) -55dBm to -40dBm (Note 22, Figure 8)	-0.1	-	0.1	dB	Forward Only	•	Forward Only	•	•	•
4-Wire to 2-Wire Forward and Reverse	-40dBm to -3dBm (Note 23, Figure 8) -55dBm to -40dBm (Note 23, Figure 8)	-0.1	-	0.1	dB	Forward Only	•	Forward Only	•	•	•
<b>NOISE</b>											
Idle Channel Noise at 2-Wire Forward and Reverse	C-Message Weighting Psophometric Weighting (Note 24, Figure 9)	-	-	12	dBrnC	Forward Only	•	Forward Only	•	•	•
Idle Channel Noise at 4-Wire Forward and Reverse	C-Message Weighting Psophometrical Weighting (Note 25, Figure 9)	-	-	-78	dBrnC	Forward Only	•	Forward Only	•	•	•
<b>HARMONIC DISTORTION</b>											
2-Wire to 4-Wire Forward and Reverse	0dBm, 0.3kHz to 3.4kHz (Note 26, Figure 7)	-	-67	-50	dB	Forward Only	•	Forward Only	•	•	•
4-Wire to 2-Wire Forward and Reverse	0dBm, 0.3kHz to 3.4kHz (Note 27, Figure 8)	-	-67	-50	dB	Forward Only	•	Forward Only	•	•	•
<b>FIGURE 10. CONSTANT LOOP CURRENT TOLERANCE</b>											
<b>FIGURE 11. TIPX VOLTAGE</b>											



**Electrical Specifications**  $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ ,  $V_{CC} = +5\text{V} \pm 5\%$ ,  $V_{BH} = -48\text{V}$ ,  $V_{BL} = \text{No connect}$ ,  $\text{PTG} = \text{Open}$ ,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_1 = 120\text{k}\Omega$ ,  $R_{DC} = 58.8\text{k}\Omega$ ,  $R_D = 38.3\text{k}\Omega$ ,  $R_{LIM} = 600\Omega$ , Unless Otherwise Specified. (\*) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part. (Continued)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNITS	HC55120	HC55121	HC55140/1	HC55142/3	HC55150/1
<b>BATTERY FEED CHARACTERISTICS</b>									
Constant Loop Current Tolerance $I_L = 26.5\text{mA}$ , $R_{LIM} = 38.3\text{k}\Omega$ Forward and Reverse	$18\text{mA} \leq I_L \leq 45\text{mA}$ , (Note 27, Figure 10)	0.92 <i>I<sub>L</sub></i>	<i>I<sub>L</sub></i>	$1.08I_L$	mA	mA	Forward Only	Forward Only	•
Tip Open State TipX Leakage Current	S = Closed (Figure 11)	-	-	-100	µA	•	•	•	•
Tip Open State RINGX Leakage Current	$R_1 = 0\Omega$ , $V_{BH} = -48\text{V}$ $R_1 = 2.5\text{k}\Omega$ , $V_{BH} = 48\text{V}$ (Figure 11)	-	26.8	-	mA	•	•	•	•
Tip Open State RINGX Voltage	$5\text{mA} < R_1 < 26\text{mA}$ (Figure 11)	-	$V_{BH} + 4.5$	-	V	•	•	•	•
Tip Voltage (Ground Start)	Active State, (S Open), $R_1 = 150\Omega$ (Figure 11)	-5	.4	-	V	NA	NA	•	NA
Tip Voltage (Ground Start)	Active State, (S Closed), Tip Lead to -48V Through 7kΩ, Ring Lead to Ground Through 150Ω (Figure 11)	-5	.4	-	V	NA	NA	•	NA
Open Circuit State Loop Current	(Active) $R_L = 0\Omega$	-100	0	100	µA	•	•	•	•
<b>LOOP CURRENT DETECTOR</b>									
Programmable Threshold Forward and Reverse	$I_{LT,TH} = (500/\text{R}_D) \geq 5\text{mA}$ , $I_{LT,TH} = 8.5\text{mA}$ $R_D = 58.8\text{k}\Omega$	$0.9I_{LT,TH}$	$I_{LT,TH}$	$1.1I_{LT,TH}$	mA	Forward Only	Forward Only	•	•
<b>GROUND KEY DETECTOR</b>									
Ground Key Detector Threshold Tip/Ring Current Difference	Tip Open Active (Note 29, $R_1 = 2.5\text{k}\Omega$ , Figure 12)	5	8	11	mA	•	NA	•	NA
<b>LINE VOLTAGE MEASUREMENT</b>									
Pulse Width (GKD, $ V_m $ )	-	0.36	-	ms/V	NA	NA	NA	•	•
<b>RING TRIP DETECTOR (DT, DR)</b>									
Ring Trip Comparator Current	Source Res = $2\text{M}\Omega$	-	2	-	µA	•	•	•	•
Input Common-Mode Range	Source Res = $2\text{M}\Omega$	-	-	$\pm 200$	V	•	•	•	•

**Electrical Specifications**  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $V_{CC} = +5V \pm 5\%$ ,  $V_{BH} = -48V$ ,  $V_{BL} = \text{No connect}$ ,  $\text{PTG} = \text{Open}$ ,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_T = 120\text{k}\Omega$ ,  $R_{LIM} = 38.3\text{k}\Omega$ ,  $R_D = 58.8\text{k}\Omega$ ,  $R_{DC\_RSR} = 20\text{k}\Omega$ ,  $R_{OH} = 50\text{k}\Omega$ ,  $C_H = 0.47\mu\text{F}$ ,  $C_{DC} = 1.0\mu\text{F}$ ,  $C_{T/REV} = 0.47\mu\text{F}$ ,  $GND = 0V$ ,  $R_L = 600\Omega$ . Unless Otherwise Specified. (•) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part. (Continued)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNITS	HC55120	HC55121	HC55130/1	HC55140/1	HC55142/3	HC55150/1
<b>RING RELAY DRIVER</b>										
$V_{SAT}$ at 50mA	$I_{OL} = 50\text{mA}$	-	0.3	0.5	V	•	•	•	•	•
Off-State Leakage Current	$V_{OH} = 13.2V$	-	-	10	$\mu\text{A}$	•	•	•	•	•
<b>TEST RELAY DRIVER (TRLY1, TRLY2)</b>										
$V_{SAT}$ at 50mA	$I_{OL} = 50\text{mA}$	-	0.3	0.5	V	NA	NA	NA	NA	NA
Off-State Leakage Current	$V_{OH} = 13.2V$	-	-	10	$\mu\text{A}$	NA	NA	NA	NA	NA

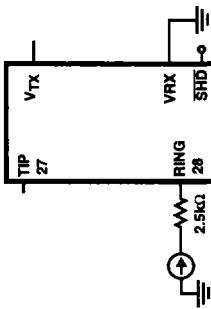


FIGURE 12. GROUND KEY DETECT

**DIGITAL INPUTS (C1, C2, C3)**

Input Low Voltage, $V_{IL}$		0	-	0.8	V	•	•	•	•	•
Input High Voltage, $V_{IH}$		2.0	-	$V_{CC}$	V	•	•	•	•	•
Input Low Current, $I_{IL}$		-	-	-50	$\mu\text{A}$	•	•	•	•	•
Input High Current, $I_{IH}$		-	25	50	$\mu\text{A}$	•	•	•	•	•

**DETECTOR OUTPUTS (SHD, GKD, LVM)**

SHD Output Low Voltage, $V_{OL}$ Forward, Reverse and Low Power Standby	$I_{OL} = 1\text{mA}$	-	-	0.5	V	Forward Only	Forward Only	•	•	•
SHD Output High Voltage, $V_{OH}$ Forward, Reverse and Low Power Standby	$I_{OH} = 100\mu\text{A}$	2.7	-	V	Forward Only	•	Forward Only	•	•	•
GKD_LVM Output Low Voltage, $V_{OL}$ Forward and Tip Open	$I_{OL} = 1\text{mA}$ $R_1 = 2.5\text{k}\Omega$ (Figure 11)	-	-	0.5	V	GKD	GKD	NA	GKD_LVM	GKD_LVM
GKD_LVM Output High Voltage, $V_{OH}$ Forward and Tip Open	$I_{OH} = 100\mu\text{A}$	2.7	-	V	•GKD	GKD	NA	GKD_LVM	GKD_LVM	GKD_LVM
Internal Pull-Up Resistor			15	-	kΩ	•	•	•	•	•

**Electrical Specifications**

$T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $V_{CC} = +5\text{V} \pm 5\%$ ,  $V_{BH} = -48\text{V}$ ,  $V_{BL} = \text{No connect}$ , PTG = Open,  $R_{F1} = R_{F2} = 0\Omega$ ,  $Z_T = 120\text{k}\Omega$ ,  $R_{LIM} = 38.3\text{k}\Omega$ ,  $R_D = 58.8\text{k}\Omega$ , RSG =  $20\text{k}\Omega$ ,  $R_{OH} = 50\text{k}\Omega$ ,  $C_H = 0.47\mu\text{F}$ ,  $CDC = 1.0\mu\text{F}$ ,  $CRT/REV = 0.47\mu\text{F}$ , GND = 0V,  $R_L = 600\text{Q}$  Unless Otherwise Specified. (\*) symbol used to indicate the test applies to the part. (NA) symbol used to indicate the test does not apply to the part. (Continued)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS	HC55120	HC55121	HC55140/1	HC55142/3	HC55150/1
<b>POWER DISSIPATION (<math>V_{BH} = -48\text{V}</math>, <math>V_{BL} = -24\text{V}</math>)</b>										
Open Circuit State	$C_1, C_2, C_3 = 0, 0, 0$	-	10	-	mW	Forward Only	•	Forward Only	•	•
Forward and Reverse						Forward Only	•	Forward Only	•	•
On-Hook, Low Power Standby	$C_1, C_2, C_3 = 1, 0, 1$	-	15	-	mW	Forward Only	•	Forward Only	•	•
Forward and Reverse						Forward Only	•	Forward Only	•	•
On-Hook, Active	$C_1, C_2, C_3 = 0, 1, 0$ $I_L = 0\text{mA}$ , Longitudinal Current = $0\text{mA}$	-	52	-	mW	Forward Only	•	Forward Only	•	•
Forward and Reverse						Forward Only	•	Forward Only	•	•
<b>POWER SUPPLY CURRENTS (<math>V_{BH} = -48\text{V}</math>, <math>V_{BL} = -24\text{V}</math>)</b>										
$V_{CC}$ Current, $ I_{CC} $ , Forward and Reverse	Open Circuit State	-	0.95	-	mA	Forward Only	•	Forward Only	•	•
$V_{BH}$ Current, $ I_{BH} $ , Forward and Reverse		-	-0.1	-	mA	Forward Only	•	Forward Only	•	•
$V_{BL}$ Current, $ I_{BL} $ , Forward and Reverse		-	-0.001	-	mA	Forward Only	•	Forward Only	•	•
$V_{CC}$ Current, $ I_{CC} $ , Forward and Reverse	Low Power Standby State $I_L = 0\text{mA}$ , Longitudinal Current = $0\text{mA}$	-	0.95	-	mA	Forward Only	•	Forward Only	•	•
$V_{BH}$ Current, $ I_{BH} $ , Forward and Reverse		-	-0.1	-	mA	Forward Only	•	Forward Only	•	•
$V_{BL}$ Current, $ I_{BL} $ , Forward and Reverse		-	-0.001	-	mA	Forward Only	•	Forward Only	•	•
$V_{CC}$ Current, $ I_{CC} $ , Forward and Reverse	Active State $I_L = 0\text{mA}$ , Longitudinal Current = $0\text{mA}$	-	2.7	-	mA	Forward Only	•	Forward Only	•	•
$V_{BH}$ Current, $ I_{BH} $ , Forward and Reverse		-	-0.8	-	mA	Forward Only	•	Forward Only	•	•
$V_{BL}$ Current, $ I_{BL} $ , Forward and Reverse		-	-0.001	-	mA	Forward Only	•	Forward Only	•	•
<b>POWER SUPPLY REJECTION RATIOS</b>										
$V_{CC}$ to 2 or 4 Wire Port	Active State $R_L = 600\Omega$ 50Hz < f < 3400Hz, $V_{IN} = 100\text{mV}$	-	40	-	dB	Forward Only	•	Forward Only	•	•
Forward and Reverse						Forward Only	•	Forward Only	•	•
$V_{BH}$ to 2 or 4 Wire Port		-40	-	dB		Forward Only	•	Forward Only	•	•
Forward and Reverse						Forward Only	•	Forward Only	•	•
$V_{BL}$ to 2 or 4 Wire Port		40	-	dB		Forward Only	•	Forward Only	•	•
Forward and Reverse						Forward Only	•	Forward Only	•	•
RFI Rejection	Figure 13, $50\text{kHz} \leq f \leq 100\text{MHz}$ High	-	TBD	-						
<b>TEMPERATURE GUARD</b>										
Junction Threshold Temperature		-	160	-	°C		•	•	•	•

## Notes

- 2. Overload Level (Two-Wire Port, Off Hook)** - The overload level is specified at the 2-wire port ( $V_{TR}$ ) with the signal source at the 4-wire receive port ( $E_{RX}$ ).  $R_L = 600\Omega$ ,  $I_{DCMET} \geq 18mA$ . Increase the amplitude of  $E_{RX}$  until 1% THD is measured at  $V_{TR}$ . Reference Figure 1.
- 3. Overload Level (Two-Wire port, On Hook)** - The overload level is specified at the 2-wire port ( $V_{TR}$ ) with the signal source at the 4-wire receive port ( $E_{RX}$ ).  $R_L = \infty$ ,  $I_{DCMET} = 0mA$ . Increase the amplitude of  $E_{RX}$  until 1% THD is measured at  $V_{TR}$ . Reference Figure 1.
- 4. Longitudinal Impedance** - The longitudinal impedance is computed using the following equations, where TIP and RING voltages are referenced to ground.  $L_{ZT}$ ,  $L_{ZR}$ ,  $V_T$ ,  $V_R$ ,  $A_T$  and  $A_R$  are defined in Figure 2.  
 $(TIP) L_{ZT} = V_T/A_T$   
 $(RING) L_{ZR} = V_R/A_R$   
where:  $E_L = 1V_{RMS}$  (0Hz to 100Hz)
- 5. Longitudinal Current Limit (On/Off-Hook Active)** - Off-Hook longitudinal current limit is determined by increasing the (60Hz) amplitude of  $E_L$  (Figure 3A) until the 2-wire longitudinal current is greater than 28mA RMS/Wire. Under this condition,  $\overline{SHD}$  pin remains low (no false detection) and the 2-wire to 4-wire longitudinal balance is verified to be greater than 45dB ( $LB_{2-4} = 20\log(V_{TX}/E_L)$ ).
- 6. Longitudinal Current Limit (On-Hook Standby)** - On-Hook longitudinal current limit is determined by increasing the (60Hz) amplitude of  $E_L$  (Figure 3B) until the 2-wire longitudinal current is greater than 8.5mA RMS/Wire. Under this condition,  $\overline{SHD}$  pin remains high (no false detection).
- 7. Longitudinal to Metallic Balance** - The longitudinal to metallic balance is computed using the following equation:  
 $BLME = 20 \cdot \log(E_L/V_{TR})$ , where:  $E_L$  and  $V_{TR}$  are defined in Figure 4.
- 8. Metallic to Longitudinal FCC Part 68, Para 68.310** - The metallic to longitudinal balance is defined in this spec.
- 9. Longitudinal to Four-Wire Balance** - The longitudinal to 4-wire balance is computed using the following equation:  
 $BLFE = 20 \cdot \log(E_L/V_{TX})$ ;  $E_L$  and  $V_{TX}$  are defined in Figure 4.
- 10. Metallic to Longitudinal Balance** - The metallic to longitudinal balance is computed using the following equation:  
 $BMLE = 20 \cdot \log(E_{TR}/V_L)$ ,  $E_{RX} = 0$   
where:  $E_{TR}$ ,  $V_L$  and  $E_{RX}$  are defined in Figure 5.
- 11. Four-Wire to Longitudinal Balance** - The 4-wire to longitudinal balance is computed using the following equation:  
 $BFLE = 20 \cdot \log(E_{RX}/V_L)$ ,  $E_{TR}$  = source is removed.  
where:  $E_{RX}$ ,  $V_L$  and  $E_{TR}$  are defined in Figure 5.
- 12. Two-Wire Return Loss** - The 2-wire return loss is computed using the following equation:  
 $r = -20 \cdot \log(2V_M/V_S)$   
where:  $Z_D$  = The desired impedance; e.g., the characteristic impedance of the line, nominally  $600\Omega$ . (Reference Figure 6).
- 13. Overload Level (4-Wire port Off-Hook)** - The overload level is specified at the 4-wire transmit port ( $V_{TX}$ ) with the signal source ( $E_G$ ) at the 2-wire port,  $Z_L = 20k\Omega$ ,  $R_L = 600\Omega$  (Reference Figure 7). Increase the amplitude of  $E_G$  until 1% THD is measured at  $V_{TX}$ . Note the PTG pin is open, and the gain from the 2-wire port to the 4-wire port is equal to 1.
- 14. Overload Level (4-Wire port On-Hook)** - The overload level is specified at the 4-wire transmit port ( $V_{TX}$ ) with the signal source ( $E_G$ ) at the 2-wire port,  $Z_L = 20k\Omega$ ,  $R_L = \infty$  (Reference Figure 7). Increase the amplitude of  $E_G$  until 1% THD is measured at  $V_{TX}$ . Note the PTG pin is open, and the gain from the 2-wire port to the 4-wire port is equal to 1.
- 15. Output Offset Voltage** - The output offset voltage is specified with the following conditions:  $E_G = 0$ ,  $R_L = 600\Omega$ ,  $Z_L = \infty$  and is measured at  $V_{TX}$ .  $E_G$ ,  $R_L$ ,  $V_{TX}$  and  $Z_L$  are defined in Figure 7.
- 16. Two-Wire to Four-Wire Frequency Response** - The 2-wire to 4-wire frequency response is measured with respect to  $E_G = 0dBm$  at 1.0kHz,  $E_{RX} = 0V$  (VRX input floating),  $R_L = 600\Omega$ . The frequency response is computed using the following equation:  
 $F_{2-4} = 20 \cdot \log(V_{TX}/V_{TR})$ , vary frequency from 300Hz to 3.4kHz and compare to 1kHz reading.  
 $V_{TX}$ ,  $V_{TR}$ ,  $R_L$  and  $E_G$  are defined in Figure 8.
- 17. Four-Wire to Two-Wire Frequency Response** - The 4-wire to 2-wire frequency response is measured with respect to  $E_{RX} = 0dBm$  at 1.0kHz,  $E_G$  source removed from circuit,  $R_L = 600\Omega$ . The frequency response is computed using the following equation:  
 $F_{4-2} = 20 \cdot \log(V_{TR}/E_{RX})$ , vary frequency from 300Hz to 3.4kHz and compare to 1kHz reading.  
 $V_{TR}$ ,  $R_L$  and  $E_{RX}$  are defined in Figure 8.
- 18. Four-Wire to Four-Wire Frequency Response** - The 4-wire to 4-wire frequency response is measured with respect to  $E_{RX} = 0dBm$  at 1.0kHz,  $E_G$  source removed from circuit,  $R_L = 600\Omega$ . The frequency response is computed using the following equation:  
 $F_{4-4} = 20 \cdot \log(V_{TX}/E_{RX})$ , vary frequency from 300Hz to 3.4kHz and compare to 1kHz reading.  
 $V_{TX}$ ,  $R_L$  and  $E_{RX}$  are defined in Figure 8.
- 19. Two-Wire to Four-Wire Insertion Loss (PTG = Open)** - The 2-wire to 4-wire insertion loss is measured with respect to  $E_G = 0dBm$  at 1.0kHz input signal,  $E_{RX} = 0$  (VRX input floating),  $R_L = 600\Omega$  and is computed using the following equation:  
 $L_{2-4} = 20 \cdot \log(V_{TX}/V_{TR})$   
where:  $V_{TX}$ ,  $V_{TR}$ ,  $R_L$  and  $E_G$  are defined in Figure 8. (Note: The fuse resistors,  $R_F$ , impact the insertion loss. The specified insertion loss is for  $R_{F1} = R_{F2} = 0$ ).
- 20. Two-Wire to Four-Wire Insertion Loss (PTG = AGND)** - The 2-wire to 4-wire insertion loss is measured with respect to  $E_G = 0dBm$  at 1.0kHz input signal,  $E_{RX} = 0$  (VRX input floating),  $R_L = 600\Omega$  and is computed using the following equation:  
 $L_{2-4} = 20 \cdot \log(V_{TX}/V_{TR})$   
where:  $V_{TX}$ ,  $V_{TR}$ ,  $R_L$  and  $E_G$  are defined in Figure 8. (Note: The fuse resistors,  $R_F$ , impact the insertion loss. The specified insertion loss is for  $R_{F1} = R_{F2} = 0$ ).
- 21. Four-Wire to Two-Wire Insertion Loss** - The 4-wire to 2-wire insertion loss is measured based upon  $E_{RX} = 0dBm$ , 1.0kHz input signal,  $E_G$  source removed from circuit,  $R_L = 600\Omega$  and is computed using the following equation:  
 $L_{4-2} = 20 \cdot \log(V_{TR}/E_{RX})$   
where:  $V_{TR}$ ,  $R_L$  and  $E_{RX}$  are defined in Figure 8.

- 22. Two-Wire to Four-Wire Gain Tracking** - The 2-wire to 4-wire gain tracking is referenced to measurements taken for  $E_G = -10\text{dBm}$ , 1.0kHz signal,  $E_{RX} = 0$  (VRX output floating),  $R_L = 600\Omega$  and is computed using the following equation.
- $$G_{2-4} = 20 \cdot \log (V_{TX}/V_{TR}) \text{ vary amplitude } -40\text{dBm} \text{ to } +3\text{dBm}, \text{ or } -55\text{dBm} \text{ to } -40\text{dBm} \text{ and compare to } -10\text{dBm} \text{ reading.}$$
- $V_{TX}$ ,  $R_L$  and  $V_{TR}$  are defined in Figure 8.
- 23. Four-Wire to Two-Wire Gain Tracking** - The 4-wire to 2-wire gain tracking is referenced to measurements taken for  $E_{RX} = -10\text{dBm}$ , 1.0kHz signal,  $E_G$  source removed from circuit,  $R_L = 600\Omega$  and is computed using the following equation:
- $$G_{4-2} = 20 \cdot \log (V_{TR}/E_{RX}) \text{ vary amplitude } -40\text{dBm} \text{ to } +3\text{dBm}, \text{ or } -55\text{dBm} \text{ to } -40\text{dBm} \text{ and compare to } -10\text{dBm} \text{ reading.}$$
- $V_{TR}$ ,  $R_L$  and  $E_{RX}$  are defined in Figure 8. The level is specified at the 4-wire receive port and referenced to a  $600\Omega$  impedance level.
- 24. Two-Wire Idle Channel Noise** - The 2-wire idle channel noise at  $V_{TR}$  is specified with the 2-wire port terminated in  $600\Omega$  ( $R_L$ ) and with the 4-wire receive port (VTX) floating (Reference Figure 9).
- 25. Four-Wire Idle Channel Noise** - The 4-wire idle channel noise at  $V_{TX}$  is specified with the 2-wire port terminated in  $600\Omega$  ( $R_L$ ).

The noise specification is with respect to a  $600\Omega$  impedance level at  $V_{TX}$ . The 4-wire receive port (VTX) floating (Reference Figure 9).

- 26. Harmonic Distortion (2-Wire to 4-Wire)** - The harmonic distortion is measured with the following conditions.  $E_G = 0\text{dBm}$  at 1kHz,  $R_L = 600\Omega$ . Measurement taken at  $V_{TX}$ . (Reference Figure 7).
- 27. Harmonic Distortion (4-Wire to 2-Wire)** - The harmonic distortion is measured with the following conditions.  $E_{RX} = 0\text{dBm}$ . Vary frequency between 300Hz and 3.4kHz,  $R_L = 600\Omega$ . Measurement taken at  $V_{TR}$ . (Reference Figure 8).
- 28. Constant Loop Current** - The constant loop current is calculated using the following equation:
- $$I_L = 1000/R_{LIM} = V_{TP}/600 \text{ (Reference Figure 10).}$$
- 29. Ground Key Detector** - (TRIGGER) Ground the Ring pin through a  $2.5k\Omega$  resistor and verify that  $\overline{GKD}$  goes low. (RESET) Disconnect the Ring pin and verify that  $\overline{GKD}$  goes high. (Hysteresis) Compare difference between trigger and reset.

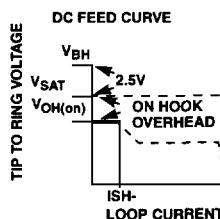
## Circuit Operation and Design Information

The UniSLIC14 family of SLIC's operate as a voltage feed current sense subscriber line interface circuit on the 2-wire side, yet appear as a current feed voltage sense SLIC on the 4-wire side. This architecture allows easy implementation of thermal management on the 2-wire side and easy connection to DSP CODEC's on the 4-wire side. Thus, for long loop applications the SLIC provides a programmed constant voltage to the tip and ring terminals while sensing the tip to ring current.

The following discussion separates the SLIC's operation into it's DC and AC paths, then follows up with additional circuit and design information.

### DC Feed Curve

The DC feed curve for the UniSLIC14 family is user programmable. The user has complete control over: The on hook and off hook overhead voltages, resistance of the feed curve, minimum open circuit voltage, the value of the current limit and the saturation guard.



The on hook overhead voltage  $V_{OH(on)}$  is independent of the  $V_{BH}$  battery voltage. So once set, the on hook voltage remains constant as the  $V_{BH}$  battery voltage changes.  $V_{OH(on)}$  also remains constant over temperature and line leakages up to ISH-. ISH- is equal to 0.6 times the Switch Hook Detect threshold current  $I_{SHD}$ . The on hook overhead

voltage, required for a given signal level, is calculated with Equation 1,

$$V_{OH(on)} = V_{sp(on)} \times \left( 1 + \frac{2R_P + 50}{Z_0} \right) \quad (\text{EQ. 1})$$

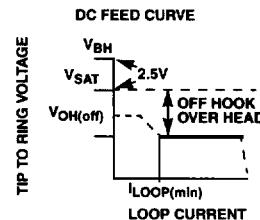
where  $V_{OH(on)}$  = On hook overhead voltage

$V_{sp(on)}$  = Required OHT (Vpeak) on hook voltage

$R_P$  = Protection Resistors (Typically  $30\Omega$ )

$Z_0$  = Line impedance for (voice)

Reference Figure 13 for  $R_P$  and  $Z_0$



The off hook overhead voltage  $V_{OH(off)}$  is also independent of the  $V_{BH}$  battery voltage and remains constant over temperature. The required off hook overhead voltage is the sum of the voltage drop across the internal sense resistors, the protection resistors, the speech signal and pulse

metering signal. The off hook overhead voltage is calculated with Equations 2 and 3.

$$V_{OH(off)} = V_{OH(Rsense)} + V_{OH(sp)} + V_{OH(pm)} \quad (\text{EQ. 2})$$

where  $V_{OH(off)}$  = Off hook overhead voltage

$V_{OH(Rsense)}$  = Required overhead for voltage drop across sense resistors as a result of the DC loop current (reference Figure 13).

$V_{OH(sp)}$  = Required (peak) off hook voltage for speech.

$V_{OH(pm)}$  = Required (peak) off hook voltage for pulse metering.

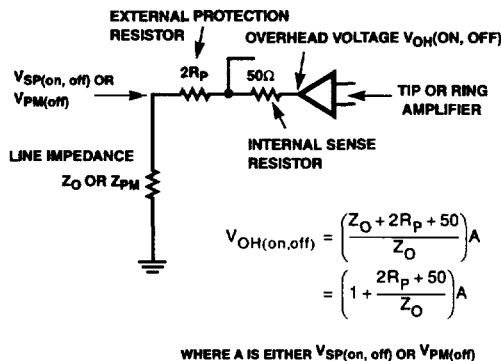
$$V_{OH(off)} = 50 \times I_{LOOP(max)} + V_{sp(off)} \times \left(1 + \frac{2R_p + 50}{Z_o}\right) + V_{pm(off)} \times \left(1 + \frac{2R_p + 50}{Z_{pm}}\right) \quad (\text{EQ. 3})$$

where  $I_{LOOP(max)}$  = Desired loop current limit.

$V_{sp(off)}$  = Required (peak) off hook voltage for speech.

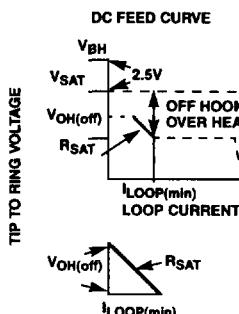
$V_{pm(off)}$  = Required (peak) off hook voltage for pulse metering.

$Z_{pm}$  = Line impedance (pulse metering, typically 200Ω).



**FIGURE 13. OVERHEAD VOLTAGE OF THE TIP AND RING AMPLIFIERS**

The Overhead voltage is defined at the output of the tip or ring amplifiers. The off hook overhead voltage requirement is the voltage drop across the internal sense resistors at maximum loop current, plus the required voltage for both speech and pulse metering. The two protection resistors ( $R_p$ ) are considered as part of the load.

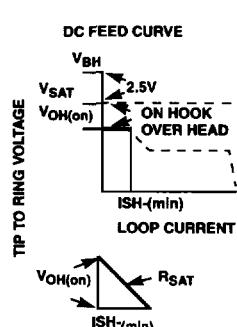


For a constant current design, the  $R_{SAT}$  resistance of the DC feed curve is equal to the  $V_{OH(off)}$  voltage divided by the minimum loop current requirement (Equation 4).

The external saturation guard resistor  $RDC\_RSG$  can now be calculated and is given in Equation 5.

$$R_{SAT} = \frac{V_{OH(off)}}{I_{LOOP(min)}} \quad (\text{EQ. 4})$$

$$RDC\_RGS = 50 \times R_{SAT} \quad (\text{EQ. 5})$$



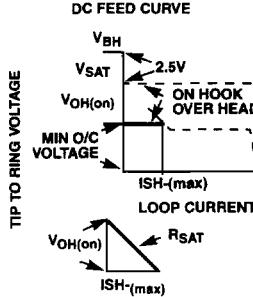
The minimum ISH- is equal to  $V_{OH(on)}$  divided by  $R_{SAT}$  as shown in Equation 6.

ISH- is used to find the minimum and maximum Switch Hook Detect thresholds ( $I_{SHD}$ ) for a given design. The minimum  $I_{SHD(min)}$  is established by the on hook overhead voltage requirement. The maximum  $I_{SHD(max)}$  is established by the minimum open circuit voltage requirement.

$$ISH_{-min} = \frac{V_{OH(on)}}{R_{SAT}} = \frac{V_{SP(on)} \times \left(1 + \frac{2R_p + 50}{Z_o}\right)}{R_{SAT}} \quad (\text{EQ. 6})$$

The minimum Switch Hook Detect threshold current ( $I_{SHD(min)}$ ) is defined by Equation 7.

$$I_{SHD(min)} = 1.7 \times ISH_{-min} \quad (\text{EQ. 7})$$



$$ISH\text{-(max)} = \frac{V_{BH} \cdot V_{SAT} - V_{O/C}}{R_{SAT}} \quad (\text{EQ. 8})$$

The maximum Switch Hook Detect Current threshold is defined by Equation 9.

$$I_{SHD(\text{MAX})} = 1.7 \times ISH\text{-(max)} \quad (\text{EQ. 9})$$

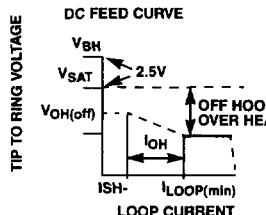
The above analysis establishes a range for the Switch Hook Detect threshold current to meet the selected requirements. After selecting an  $I_{SHD}$  threshold between the  $I_{SHD(\text{min})}$  and  $I_{SHD(\text{max})}$ , the programming resistor is determined by Equation 10.

$$R_D = \frac{500}{I_{SHD}} \quad (\text{EQ. 10})$$

The true value of  $ISH\text{-}$  for the selected value of  $I_{SHD}$  is given by Equation 11:

$$ISH\text{-} = I_{SHD} (0.6) \quad (\text{EQ. 11})$$

Verify that the value of  $ISH\text{-}$  is above the suspected line leakage of the application. The UniSLIC family will provide a constant on hook voltage level for leakage currents up to this value of line leakage.



$$R_{OH(\text{max})} = \frac{500}{IOH} = \frac{500}{I_{LOOP(\text{min})} \cdot ISH\text{-}} \quad (\text{EQ. 12})$$

The maximum  $ISH\text{-}$  is determined by the minimum open circuit voltage requirement (Typical value is -43 volts).

Equation 8 determines the maximum  $ISH\text{-}$  allowed for a given minimum open circuit voltage.

The current limit is set by a single resistor and is calculated using Equation 13:

$$R_{LIM} = \frac{1000}{I_{LOOP(\text{max})}} \quad (\text{EQ. 13})$$

The maximum loop resistance for the given conditions is calculated by Equation 14.

$$R_{LOOP(\text{max})} = \frac{V_{BH} - V_{SAT}}{I_{LOOP(\text{min})}} - 50 - \frac{V_{SP(\text{off})}}{I_{LOOP(\text{min})}} \times \left( 1 + \frac{2R_P + 50}{Z_0} \right) - \frac{V_{PM(\text{off})}}{I_{LOOP(\text{min})}} \times \left( 1 + \frac{2R_P + 50}{Z_{PM}} \right) \cdot 2R_P \quad (\text{EQ. 14})$$

### SLIC in the Standby Mode

Overall system power is saved by configuring the SLIC in the standby state when not in use. In the standby state the tip and ring amplifiers are disabled and internal resistors are connected between tip to ground and ring to  $V_{BH}$ . This connection enables a loop current to flow when the phone goes off hook. The loop current detector then detects this current and the  $\overline{SHD}$  pin goes low.

### SLIC in the Active Mode

Figure 14 shows a simplified AC transmission model. Circuit analysis yields the following design equations:

Node Equation

$$\frac{V_{RX}}{500K} - \frac{V_A}{1000K} = I_X \quad (\text{EQ. 15})$$

Loop Equation

$$I_X 500k - V_{TX} = I_X 500k = 0 \quad (\text{EQ. 16})$$

Loop Equation

$$V_{TR} - I_M 2R_P + V_{TX} = 0 \quad (\text{EQ. 17})$$

where:

$V_{RX}$  = Is the input voltage at the VRX pin.

$V_A$  = Is an internal node voltage that is a function of the loop current detector and the impedance matching networks

$I_X$  = Internal current in the SLIC that is the result between the input receive current and the feedback current.

$I_M$  = Is the AC metallic current.

$R_P$  = Is a fuse resistor.

$Z_T$  = Is used to set the SLIC's 2-wire impedance.

$V_{TX}$  = Is the tip to ring voltage at the output pins of the SLIC.

$V_{TR}$  = Is the tip to ring voltage including the voltage across the protection resistors.

$Z_L$  = Is the line impedance.

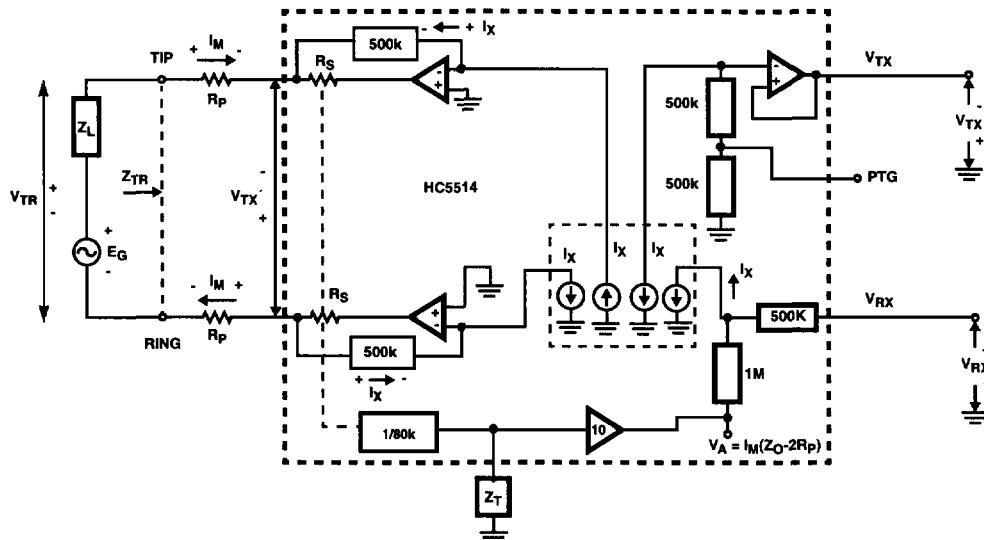


FIGURE 14. SIMPLIFIED AC TRANSMISSION CIRCUIT

#### (AC) 4-Wire to 2-Wire Gain

The 4-wire to 2-wire gain is equal to  $V_{TR}/V_{RX}$ .

From Equations 15, 16 and 17 with  $E_G = 0$ .

$$A_{4-2} = -\frac{Z_L}{Z_O + Z_L} \quad (\text{EQ. 18})$$

#### (AC) 2-Wire to 4-Wire Gain

The 2-wire to 4-wire gain is equal to  $V_{TX}/E_G$  with  $V_{RX} = 0$ .

$$A_{2-4} = \frac{V_{TX}}{E_G} = -\frac{Z_O - 2R_P}{Z_O + Z_L} \quad (\text{EQ. 19})$$

#### (AC) 4-Wire to 4-Wire Gain

The 4-wire to 4-wire gain is equal to  $V_{TX}/V_{RX}$ .

From Equations 15, 16 and 17 with  $E_G = 0$ .

$$A_{4-4} = \frac{V_{TX}}{V_{RX}} = 2 \left( \frac{Z_L - 2R_P}{Z_L + Z_O} \right) \quad (\text{EQ. 20})$$

#### (AC) 2-Wire Impedance

The AC 2-wire impedance ( $Z_{TR}$ ) is the impedance looking into the SLIC, including the fuse resistors. The formula to calculate the proper  $Z_T$  for matching the 2-wire impedance is shown in Equation 21.

$$Z_T = 200 \cdot (Z_{TR} - 2R_P) \quad (\text{EQ. 21})$$

Equation 21 can now be used to match the SLIC's impedance to any known line impedance ( $Z_{TR}$ ).

#### EXAMPLE:

Calculate  $Z_T$  to make  $Z_{TR} = 600\Omega$  in series with  $2.16\mu F$ .  $R_P = 30\Omega$ .

$$Z_T = 200 \cdot \left( 600 + \frac{1}{j\omega \cdot 2.16 \cdot 10^{-6}} - 2 \cdot 30 \right) \quad (\text{EQ. 22})$$

$Z_T = 114k\Omega$  in series with  $0.0108\mu F$ .

#### Layout Considerations

##### Floating the PTG Pin:

The PTG pin is a high impedance pin that is used to reduce the 2-wire to 4-wire gain to 0dB. If 0dB is required, it is necessary to float the PTG pin. The PC board interconnect should be as short as possible to minimize stray capacitance on this pin. Stray capacitance on this pin forms a low pass filter and will cause the 2-wire to 4-wire gain to roll off at the higher frequencies.

If a 2-wire to 4-wire gain of -6dB is required in the design, the PTG pin should be grounded at the pin.

##### Layout of the 2-Wire Impedance Matching Resistor $Z_T$ :

Proper connection to this pin is to have the external  $Z_T$  network as close to the pin as possible.

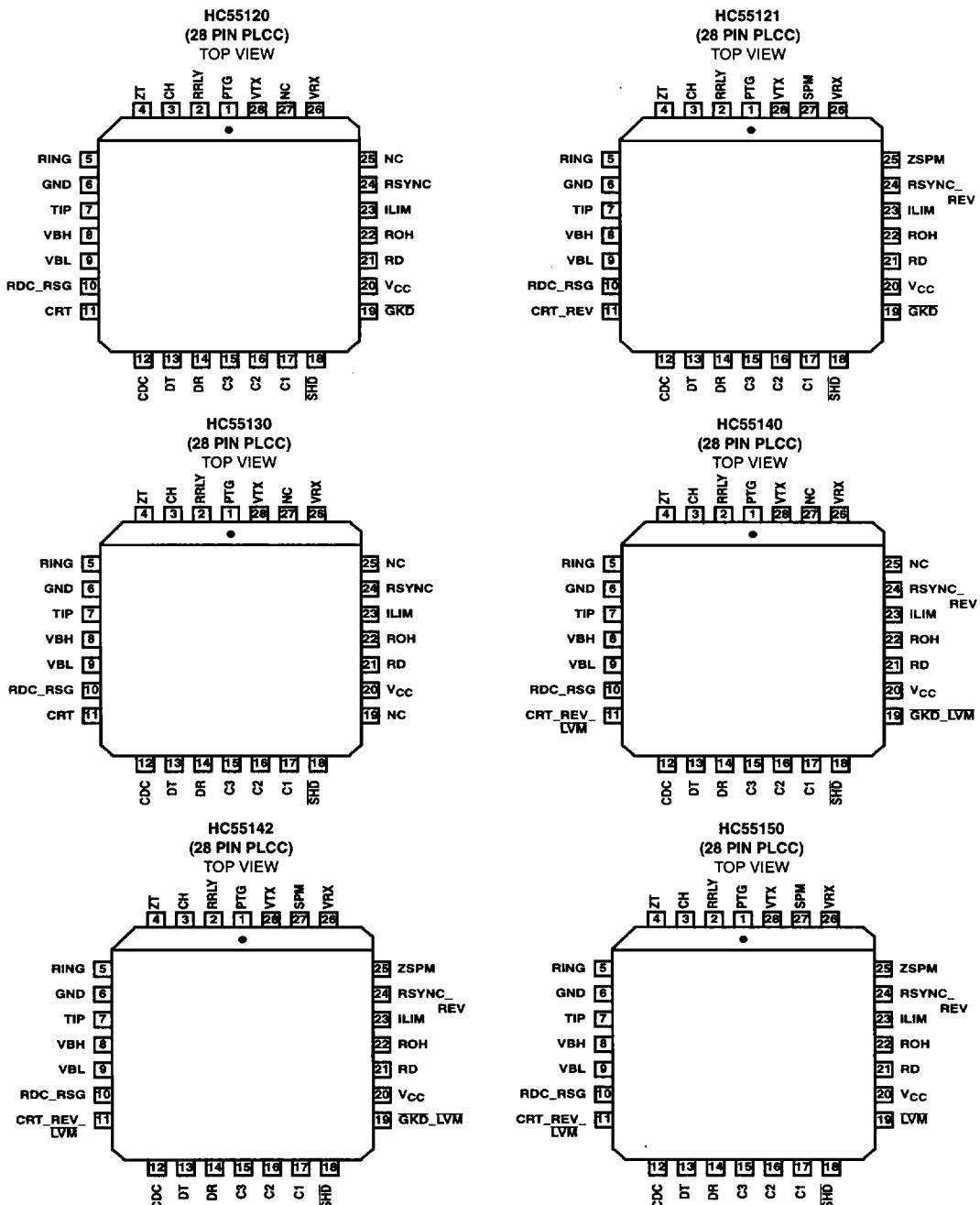
The  $Z_T$  pin is a high impedance pin that is used to set the proper feedback for matching the impedance of the 2-wire side. This will eliminate circuit board capacitance on this pin to maintain the 2-wire return loss across frequency.

##### SPM Pin:

For optimum performance, the PC board interconnect connected to the SPM pin should be as short as possible.

TABLE 1. DETECTOR STATES

STATE	C3	C2	C1	SLIC OPERATING STATE	ACTIVE DETECTOR	OUTPUT	
						SHD	GKD_LVM
0	0	0	0	Open Circuit State		HIGH	HIGH
1	0	0	1	Ringing State	Ring Trip Detector		HIGH
2	0	1	0	Forward Active State	Loop Current Detector		
					Ground Key Detector		
3	0	1	1	Test Active State Requires previous state to be in the Forward Active state to determine the On hook or Off hook status of the line.	On Hook Loopback Detector	LOW	
					Ground Key Detector		
					On Hook Loop Current Detector	LOW	
					Line Voltage Detector		
4	1	0	0	Tip Open State	Ground Key Detector		
5	1	0	1	Low Power Standby State	Loop Current Detector		
					Ground Key Detector		
6	1	1	0	Reverse Active State	Loop Current Detector		
					Ground Key Detector		
7	1	1	1	Test Reversal Active State Requires previous state to be in the Reverse Active state to determine the On hook or Off hook status of the line.	On Hook Loop Current Detector		HIGH
					On Hook Loop Current Detector	LOW	
					Line Voltage Detector		
8	X	X	X	Thermal Shutdown		LOW	LOW



**Pinouts SOIC Packages**

HC55120 (28 PIN SOIC) TOP VIEW		HC55121 (28 PIN SOIC) TOP VIEW	
ZT	28 NC	ZT	28 ZSPM
PTG	27 VTX	PTG	27 VTX
RRLY	26 NC	RRLY	26 SPM
CH	25 VRX	CH	25 VRX
RING	24 RSYNC	RING	24 RSYNC_REV
GND	23 ILIM	GND	23 ILIM
TIP	22 ROH	TIP	22 ROH
VBH	21 RD	VBH	21 RD
VBL	20 VCC	VBL	20 VCC
RDC_RSG	19 SHD	RDC_RSG	19 SHD
CDC	18 C1	CDC	18 C1
DT	17 C2	DT	17 C2
DR	16 C3	DR	16 C3
CRT	15 GKD	CRT_REV	15 GKD

HC55130 (28 PIN SOIC) TOP VIEW		HC55140 (28 PIN SOIC) TOP VIEW	
ZT	28 NC	ZT	28 NC
PTG	27 VTX	PTG	27 VTX
RRLY	26 NC	RRLY	26 NC
CH	25 VRX	CH	25 VRX
RING	24 RSYNC	RING	24 RSYNC_REV
GND	23 ILIM	GND	23 ILIM
TIP	22 ROH	TIP	22 ROH
VBH	21 RD	VBH	21 RD
VBL	20 VCC	VBL	20 VCC
RDC_RSG	19 SHD	RDC_RSG	19 SHD
CDC	18 C1	CDC	18 C1
DT	17 C2	DT	17 C2
DR	16 C3	DR	16 C3
CRT	15 NC	CRT_REV_LVM	15 GKD_LVM

HC55142 (28 PIN SOIC) TOP VIEW		HC55150 (28 PIN SOIC) TOP VIEW	
ZT	28 ZSPM	ZT	28 ZSPM
PTG	27 VTX	PTG	27 VTX
RRLY	26 SPM	RRLY	26 SPM
CH	25 VRX	CH	25 VRX
RING	24 RSYNC_REV	RING	24 RSYNC_REV
GND	23 ILIM	GND	23 ILIM
TIP	22 ROH	TIP	22 ROH
VBH	21 RD	VBH	21 RD
VBL	20 VCC	VBL	20 VCC
RDC_RSG	19 SHD	RDC_RSG	19 SHD
CDC	18 C1	CDC	18 C1
DT	17 C2	DT	17 C2
DR	16 C3	DR	16 C3
CRT	15 GKD_LVM	CRT_REV_LVM	15 LVM

**Pin Descriptions**

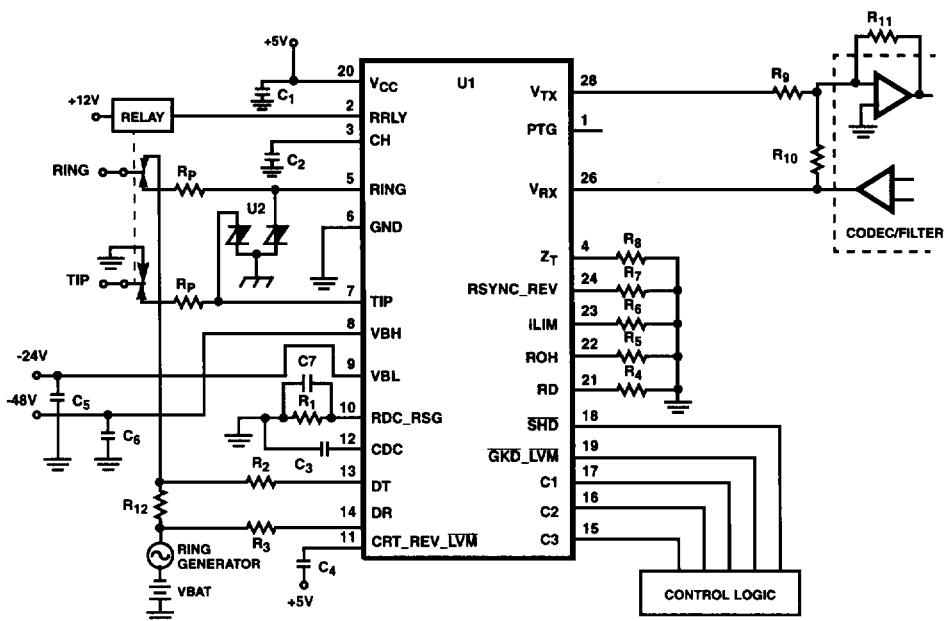
28 PIN PLCC	32 PIN PLCC	28 PIN SOIC	SYMBOL	DESCRIPTION
1	1	2	PTG	Programmable Transmit Gain - The 2-wire to 4-wire transmission gain is 0dB if this pin is left floating and -6.02dB if tied to ground. The -6.02dB gain option is useful in systems where Pulse Metering is used. See Figure 14.
2	2	3	RRLY	Ring Relay Driver Output - The relay coil may be connected to a maximum of 14 volts.
3	3	4	CH	AC/DC Separation Capacitor - CH is required to properly process the AC current from the DC loop current. Recommended value 0.47μF.
4	4	1	ZT	2-Wire Impedance Matching Pin - Impedance matching of the 2-wire side is accomplished by placing an impedance between the Z <sub>T</sub> pin and ground. See Equation 21.
5	5	5	RING	Connects via protection resistors (R <sub>P</sub> reference Figure 15 - 18) to ring wire of subscriber pair.
6	6	6	GND	Analog and Digital ground.
7	7	7	TIP	Connects via protection resistors (R <sub>P</sub> reference Figure 19) to tip wire of subscriber pair.
8	8	8	V <sub>BH</sub>	High Battery Supply (negative with respect to GND pin 6).
9	9	9	V <sub>BL</sub>	Low Battery Supply (negative with respect to GND pin 6, magnitude ≤ V <sub>BH</sub> ).
10	10	10	RDC_RSG	Resistive Feed/Saturation Guard - Performs the saturation guard function on constant current designs and sets the slope of the resistive feed curve for constant voltage designs.
11	11	14	CRT_REV_LVM	Ring Trip, Soft Polarity Reversal and Line Voltage Measurement - A capacitor when placed between the CRT_REV_LVM pin and ground performs 3 mutually exclusive functions. When the SLIC is configured in the Ringing mode it provides filtering of the ringing signal to prevent false detect. When the SLIC is transitioning between the Forward Active State and Reverse Active State it provides Soft Polarity Reversal and performs charge storage in the Line Voltage Measurement State. Recommended value 0.47μF.
12	12	11	CDC	Filter Capacitor- The CDC Capacitor removes the VF signals from the battery feed control loop.
13	13	12	DT	Tip side of Ring Trip Detector - Ring trip detection is accomplished by connecting an external network to a detector in the SLIC with inputs DT and DR. Ring trip occurs when the voltage on DT is more negative than the voltage on DR.
14	14	13	DR	Ring Side of Ring Trip Detector - Ring trip detection is accomplished by connecting an external network to a detector in the SLIC with inputs DT and DR. Ring trip occurs when the voltage on DR is more positive than the voltage on DT.
-	15	-	C5	Activates Test Relay TRLY1.
-	16	-	C4	Activates Test Relay TRLY2.
15	17	16	C3	TTL Compatible Logic Input. The logic states of C1, C2 and C3 determine the operating states of the SLIC. Reference Table 1 for details.
16	18	17	C2	TTL Compatible Logic Input. The logic states of C1, C2 and C3 determine the operating states of the SLIC. Reference Table 1 for details.
17	19	18	C1	TTL Compatible Logic Input. The logic states of C1, C2 and C3 determine the operating states of the SLIC. Reference Table 1 for details.
18	20	19	SHD	Switch Hook Detect - Active during off hook, ground key and loopback. Reference Table 1 for details.
19	21	15	GKD_LVM	Ground Key Detector and Line Voltage Measurement - Reference Table 1 for details.
20	22	20	VCC	5V Supply.

**Pin Descriptions (Continued)**

28 PIN PLCC	32 PIN PLCC	28 PIN SOIC	SYMBOL	DESCRIPTION
21	23	21	RD	Loop Current Threshold Programming Pin - A resistor between this pin and ground will determine the trigger level for the loop current detect circuit. See Equation 10.
22	24	22	ROH	Off Hook Overload Setting Resistor - Used to set combined overhead for voice and pulse metering signals. See Equation 12.
23	25	23	ILIM	Current Limit Programming Pin - A resistor between this pin and ground will determine the constant current limit of the feed curve. See Equation 13.
24	26	24	RSYNC_REV	Ring Synchronization Input and Reversal Time Setting. A resistor between this pin and GND (pin 6) determines the polarity reversal time. Synchronization of the closing of the relay to connect the ringing signal to the subscriber pair is achieved via the grounding of this pin.
25	27	28	ZSPM	Pulse Metering Signal Impedance Pin - A resistor on the input of this pin will allow programming of the source impedance of the pulse metering signal for maximum signal on the 2-wire loop.
26	28	25	VRX	Receive Input - Ground referenced 4-wire side.
27	29	26	SPM	Pulse Metering Signal Input.
28	30	27	VTX	Transmit Output - Ground referenced 4-wire side.
-	31	-	TRLY1	Test Relay Driver 1.
-	32	-	TRLY2	Test Relay Driver 2.

**Basis Application Circuit**

**Voice Only 28 Pin PLCC Package**



**FIGURE 15. UniSLIC14 VOICE ONLY BASIC APPLICATION CIRCUIT**

**TABLE 2. BASIC APPLICATION CIRCUIT COMPONENT LIST**

COMPONENT	VALUE	TOLERANCE	RATING
U1 - SLIC	UniSLIC14 Family	N/A	N/A
U2 - Dual Asymmetrical Transient Voltage Suppressor	TISP1072F3	N/A	N/A
R <sub>P</sub> (Line Feed Resistors)	30Ω	Matched 1%	2.0W
R1 (RDC_RSG Resistor)	13kΩ	1%	1/4W
R2, R3	2 MegΩ	1%	1/4W
R4 (ROH Resistor)	43.2kΩ	1%	1/4W
R5 (RILIM Resistor)	37.4kΩ	1%	1/4W
R6 (RBLIM Resistor)	33.2kΩ	1%	1/4W
R7 (RSYNC_REV Resistor)	34.8kΩ	1%	1/4W
R8 (RZT Resistor)	107kΩ	1%	1/4W
R9, R10, R11	20kΩ	1%	1/4W
R12	400Ω	5%	2W
C1	0.1µF + 1.0µF	20%	10V
C5	0.1µF + 1.0µF	20%	50V
C6	0.1µF + 1.0µF	20%	100V
C2, C4, C7	0.47µF	20%	10V
C3	2.2µF	20%	50V

**Design Parameters:** Maximum onhook voltage = 0.775V<sub>RMS</sub>. Maximum Offhook Voice = 3.1Vpeak, Switch Hook Threshold = 11.6mA, Loop Current Limit = 29.6mA, Synthesize Device Impedance = 600 -60 = 540Ω, with 30Ω protection resistors, impedance across Tip and Ring terminals = 600Ω. Where applicable, these component values apply to the Basic Application Circuits for the HC55120, HC55121, HC55130, HC55140, HC55142 and HC55150. Pins not shown in the Basic Application Circuit are no connect (NC) pins.

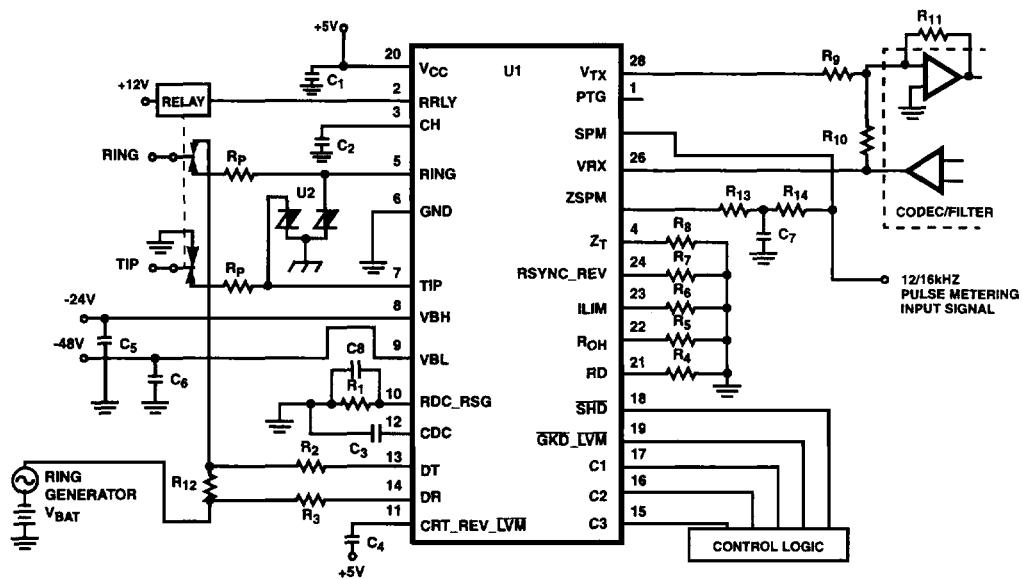
**Basis Application Circuit****Pulse Metering 28 Pin PLCC Package**

FIGURE 16. UniSLIC14 PULSE METERING BASIC APPLICATION CIRCUIT

TABLE 3. BASIC APPLICATION CIRCUIT COMPONENT LIST

COMPONENT	VALUE	TOLERANCE	RATING
U1 - SLIC	UniSLIC14 Family	N/A	N/A
U2 - Dual Asymmetrical Transient Voltage Suppressor	TISP1072F3	N/A	N/A
RP (Line Feed Resistors)	30Ω	Matched 1%	2.0W
R1 (RDC_RSG Resistor)	19.1kΩ	1%	1/4W
R2, R3	2 MegΩ	1%	1/4W
R4 (RD Resistor)	51.1kΩ	1%	1/4W
R5 (ROH Resistor)	34.8kΩ	1%	1/4W
R6 (RILIM Resistor)	33.2kΩ	1%	1/4W
R7 (RSYNC_REV Resistor)	34.8kΩ	1%	1/4W
R8 (RZT Resistor)	107kΩ	1%	1/4W
R9, R10, R11	20kΩ	1%	1/4W
R12	400Ω	5%	2W
R13, R14	10.7kΩ	1%	1/4W
C1	0.1μF + 1.0μF	20%	10V
C5	0.1μF + 1.0μF	20%	50V
C6	0.1μF + 1.0μF	20%	100V
C2, C4, C8	0.47μF	20%	10V
C3	2.2μF	20%	50V
C7	680pF	10%	10V

**Design Parameters:** Maximum onhook voltage = 0.775VRMS, Maximum offhook voice = 1.1Vpeak, Maximum simultaneous pulse metering signal = 3.1Vpeak, Switch Hook Threshold = 9.8mA, Loop Current Limit = 29.6mA, Synthesize Device Impedance = 600-60 = 540Ω, with 30Ω protection resistors, impedance across Tip and Ring terminals = 600Ω. Where applicable, these component values apply to the Basic Application Circuits for the HC55120, HC55121, HC55130, HC55140, HC55142 and HC55150. Pins not shown in the Basic Application Circuit are no connect (NC) pins.

**Basis Application Circuit**

**Voice Only 28 Pin SOIC Package**

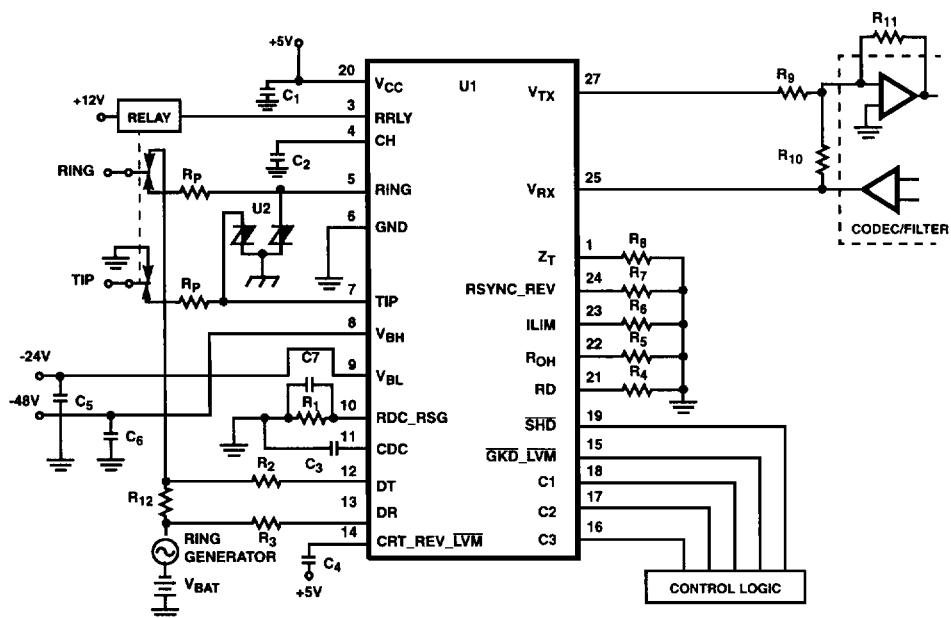


FIGURE 17. UniSLIC14 VOICE ONLY BASIC APPLICATION CIRCUIT

TABLE 4. BASIC APPLICATION CIRCUIT COMPONENT LIST

COMPONENT	VALUE	TOLERANCE	RATING
U1 - SLIC	UniSLIC14 Family	N/A	N/A
U2 - Dual Asymmetrical Transient Voltage Suppressor	TISP1072F3	N/A	N/A
RP (Line Feed Resistors)	30Ω	Matched 1%	2.0W
R1 (RDC_RSG Resistor)	13kΩ	1%	1/4W
R2, R3	2 MegΩ	1%	1/4W
R4 (RD Resistor)	43.2kΩ	1%	1/4W
R5 (ROH Resistor)	37.4kΩ	1%	1/4W
R6 (RILIM Resistor)	33.2kΩ	1%	1/4W
R7 (RSYNC_REV Resistor)	34.8kΩ	1%	1/4W
R8 (RZT Resistor)	107kΩ	1%	1/4W
R9, R10, R11	20kΩ	1%	1/4W
R12	400Ω	5%	2W
C1	0.1μF + 1.0μF	20%	10V
C5	0.1μF + 1.0μF	20%	50V
C6	0.1μF + 1.0μF	20%	100V
C2, C4, C7	0.47μF	20%	10V
C3	2.2μF	20%	50V

**Design Parameters:** Maximum onhook voltage = 0.775V RMS, Maximum offhook voice = 3.1V peak, Switch Hook Threshold = 11.6mA, Loop Current Limit = 29.6mA, Synthesize Device Impedance = 600-60 = 540Ω, with 30Ω protection resistors, Impedance across Tip and Ring terminals = 600Ω. Where applicable, these component values apply to the Basic Application Circuits for the HC55120, HC55121, HC55130, HC55140, HC55142 and HC55150. Pins not shown in the Basic Application Circuit are no connect (NC) pins.

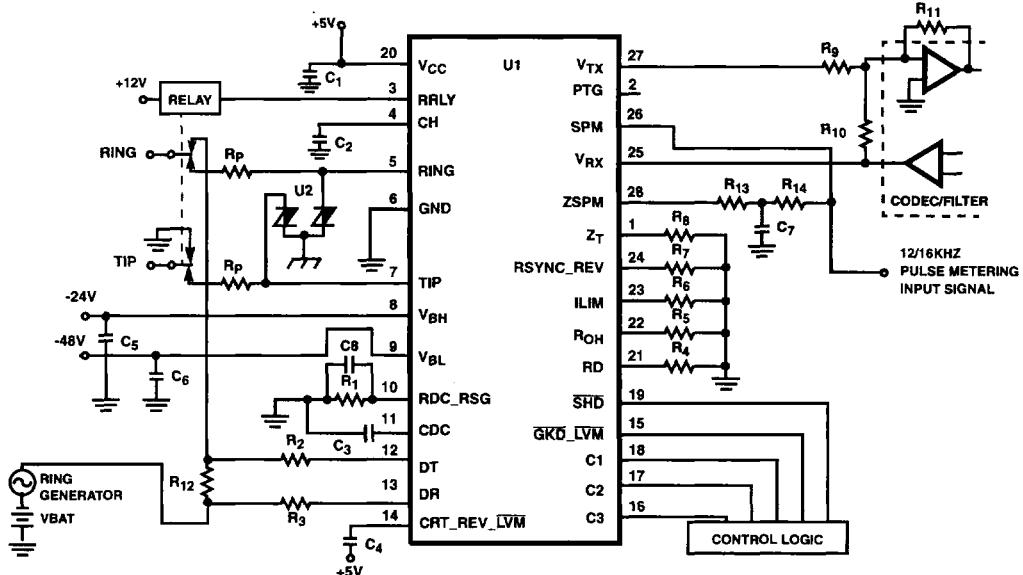
**Basis Application Circuit****Pulse Metering 28 Pin SOIC Package**

FIGURE 18. UniSLIC14 PULSE METERING BASIC APPLICATION CIRCUIT

TABLE 5. BASIC APPLICATION CIRCUIT COMPONENT LIST

COMPONENT	VALUE	TOLERANCE	RATING
U1 - SLIC	UniSLIC14 Family	N/A	N/A
U2 - Dual Asymmetrical Transient Voltage Suppressor	TISP1072F3	N/A	N/A
RP (Line Feed Resistors)	30Ω	Matched 1%	2.0W
R1 (RDC_RSG Resistor)	19.1kΩ	1%	1/4W
R2, R3	2 MegΩ	1%	1/4W
R4 (RD Resistor)	51.1kΩ	1%	1/4W
R5 (ROH Resistor)	34.8kΩ	1%	1/4W
R6 (RILIM Resistor)	33.2kΩ	1%	1/4W
R7 (RSYNC_REV Resistor)	34.8kΩ	1%	1/4W
R8 (RZT Resistor)	107kΩ	1%	1/4W
R9, R10, R11	20kΩ	1%	1/4W
R12	400Ω	5%	2W
R13, R14	10.7kΩ	1%	1/4W
C1	0.1μF + 1.0μF	20%	10V
C5	0.1μF + 1.0μF	20%	50V
C6	0.1μF + 1.0μF	20%	100V
C2, C4, C8	0.47μF	20%	10V
C3	2.2μF	20%	50V
C7	680pF	10%	10V

**Design Parameters:** Maximum onhook voltage = 0.775V<sub>RMS</sub>, Maximum offhook voice = 1.1V<sub>peak</sub>, Maximum simultaneous pulse metering signal = 3.1V<sub>peak</sub>, Switch Hook Threshold = 9.8mA, Loop Current Limit = 29.6mA, Synthesize Device Impedance = 600-60 = 540Ω, with 30Ω protection resistors, impedance across Tip and Ring terminals = 600Ω. Where applicable, these component values apply to the Basic Application Circuits for the HC55120, HC55121, HC55130, HC55140, HC55142 and HC55150. Pins not shown in the Basic Application Circuit are no connect (NC) pins.